# Features

- ARM7TDMI<sup>®</sup> ARM<sup>®</sup> Thumb<sup>®</sup> Processor Core
  - High Performance 32-bit RISC
  - High-density 16-bit Instruction set (Thumb)
  - Leader in MIPS/Watt
  - Embedded ICE (In Circuit Emulation)
- 16 Kbytes Internal SRAM
- Fully Programmable External Bus Interface (EBI)
  - Maximum External Address Space of 6 Mbytes, Up to Four Chip Select Lines
- 8-level Priority, Vectored Interrupt Controller
  - Three External Interrupts Including One Fast Interrupt Line
- Ten-channel Peripheral Data Controller (PDC)
- 57 Programmable I/O Lines
- Four 16-bit General Purpose Timers (GPT)
  - Three Configurable Modes: Counter, PWM, Capture
  - Four External Clock Inputs, Three Multi-purpose I/O Pins per Timer
- Four 16-bit Simple Timers (ST)
- Four Channel 16-bit Pulse Width Modulation (PWM)
- Four CAN Controllers 2.0A and 2.0B Full CAN
- One with 32 Buffers, Three with 16 Buffers
- Two USARTs
  - Support for J1587 and LIN Protocols
- One Master/Slave SPI Interface
  - 8 to 16-bit Programmable Data Length
  - Four External Serial Peripheral Chip Selects
- Two 8-channel 10-bit Analog to Digital Converters (ADC)
- Two 16-bit Capture Modules (CAPT)
- Programmable Watch Timer (WT)
- Programmable Watchdog (WD)
- Power Management Controller (PMC)
   32 kHz Oscillator, Main Oscillator and PLL
- IEEE 1149.1 JTAG Boundary-scan on all Digital Pins
- Fully Static Operation: 0 Hz to 30 MHz at VDDCORE = 3.3V, 85°C
- 3.0V to 5.5V Operating Voltage Range
- 3.0V to 3.6V Core, Memory and Analog Voltage Range
- -40° to +85°C Operating Temperature Range
- Available in a 176-lead LQFP Package

# 1. Description

The AT91SAM7A2 is based on the ARM7TDMI embedded processor. This processor has a high-performance 32-bit RISC architecture with a high-density 16-bit instruction set and very low power consumption.

In addition, a large number of internally banked registers result in very fast exception handling, making the device ideal for real-time control applications.

The AT91SAM7A2 has a direct connection to off-chip memory, including Flash, through the fully programmable External Bus Interface.



AT91 ARM Thumb Microcontrollers

# AT91SAM7A2

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An 8-level priority vectored Interrupt Controller in conjunction with the Peripheral Data Controller significantly improves the real time performance of the device. The device is manufactured using high-density CMOS technology.

By combining the ARM7TDMI processor with an on-chip SRAM, and a wide range of peripheral functions, including USART, SPI, CAN Controllers, Timer Counter and Analog-to-Digital Converters, on a monolithic chip, the AT91SAM7A2 is a powerful device that provides a flexible, cost-effective solution to many compute-intensive embedded control applications in the automotive and industrial world.

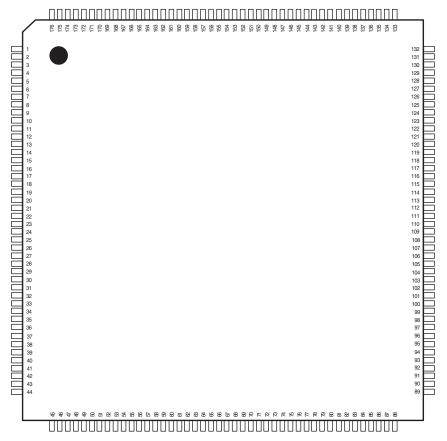
# 2. Pin Configuration Table 2-1. Pinout

Pin	Name	Pin	Name	Pin	Name	Pin	Name
1	VDDIO	45	GND	89	VDDIO	133	NOE/NRD
2	IRQ0	46	VDDIO	90	VDDANA	134	NCS0
3	IRQ1	47	UPIO5	91	VREFP0	135	ADD1
4	FIQ	48	UPIO6	92	ANA0IN0	136	D9
5	SCK0/MPIO	49	GND	93	ANA0IN1	137	D2
6	TXD0/MPIO	50	VDDIO	94	ANA0IN2	138	VDDCORE
7	RXD0/MPIO	51	UPIO7	95	ANA0IN3	139	D10
8	SCK1/MPIO	52	UPIO 8	96	ANA0IN4	140	D3
9	TXD1/MPIO	53	UPIO 9	97	ANA0IN5	141	D11
10	RXD1/MPIO	54	UPIO 10	98	ANA0IN6	142	D4
11	VDDCORE	55	UPIO 11	99	GND	143	D12
12	CANTX3	56	UPIO 12	100	VDDANA	144	D5
13	CANRX3	57	UPIO 13	101	ANA0IN7	145	D13
14	CAPT0	58	UPIO 14	102	VREFP1	146	D6
15	CAPT1	59	UPIO 15	103	ANA1IN0	147	D14
16	SPCK/MPIO	60	UPIO 16	104	ANA1IN1	148	D7
17	MISO/MPIO	61	UPIO 17	105	ANA1IN2	149	D15
18	MOSI/MPIO	62	UPIO 18	106	ANA1IN3	150	GND
19	NPCS0/MPIO	63	GND	107	ANA1IN4	151	ADD0/NLB
20	VDDIO	64	VDDIO	108	ANA1IN5	152	ADD17
21	GND	65	UPIO19	109	ANA1IN6	153	ADD16
22	NPCS1/MPIO	66	UPIO20	110	ANA1IN7	154	ADD15
23	NPCS2/MPIO	67	UPIO21	111	GND	155	ADD14
24	NPCS3/MPIO	68	UPIO22	112	VDDCORE	156	ADD13
25	T0TIOA0/MPIO	69	UPIO23	113	RTCKI	157	ADD12
26	T0TIOB0/MPIO	70	UPIO24	114	RTCKO	158	ADD11
27	T0TCLK0/MPIO	71	UPIO25	115	GND	159	ADD10
28	T0TIOA1/MPIO	72	UPIO26	116	VDDCORE	160	ADD9
29	T0TIOB1/MPIO	73	UPIO27	117	SCANEN	161	ADD20/CS3
30	T0TCLK1/MPIO	74	UPIO28	118	TEST	162	VDDCORE
31	T0TIOA2/MPIO	75	UPIO29	119	TMS	163	NWR0/NWE
32	T0TIOB2/MPIO	76	UPIO30/NWAIT	120	TDO	164	NCS2
33	VDDIO	77	UPIO31/CORECLK	121	TDI	165	NCS1
34	GND	78	CANTX0	122	TCK	166	ADD19
35	T0TCLK2/MPIO	79	CANRX0	123	GND	167	ADD18
36	T1TIOA0/MPIO	80	CANTX1	124	PLLRC	168	ADD8
37	T1TIOB0/MPIO	81	CANRX1	125	VDDCORE	169	ADD7
38	T1TCLK0/MPIO	82	CANTX2	126	MCKI	170	ADD6
39	NRESET	83	CANRX2	127	МСКО	171	ADD2
40	UPIO0	84	PWM0	128	GND	172	ADD3
41	UPIO1	85	PWM1	120	NWR1/NUB	172	ADD3
42	UPIO2	86	PWM2	130	D8	173	ADD4 ADD5
43	UPIO3	87	PWM3	130	D1	174	GND
43	UPIO4	88	GND	132	D1	175	GND





### Figure 2-1. Pin Configuration



# 4 **AT91SAM7A2**

# 3. Signal Description

# Table 3-1.Signal Description

Module	Name	Function	Туре	Active Level	Comments
	ADD[19:1]	External address bus	0	(Z) <sup>(1)</sup>	
	ADD0/NLB	External address line line/ Lower byte enable	0	L (Z)	
	ADD20/CS3	External address line/ Chip select	0	H (Z)	The EBI is tri-stated when NRESET is
	D[15:0] External da	External data bus	I/O	(Z)	at a logical low level.
	NOE	Output enable	0	L (Z)	Internal pull-downs on data bus bits
EBI	NWR0/NWE	Write enable	0	L (Z)	
	NCS[2:0]	Chip select lines	0	L (Z)	
	NWR1/NUB	Upper byte enable	0	L (Z)	
	NWAIT	External Wait	I	L	Disable at reset, multiplexed with UPIO30
	CORECLK	Core CLock	0		Disable at reset, multiplexed with UPIO31
GIC	IRQ[1:0]	External interrupt lines	Ι		
GIC	FIQ	Fast interrupt line	I		
Power-on Reset	NRESET	Hardware reset input	I	L	Schmitt input with internal filter
	МСКІ	Master clock input	I		Connected to external crystal (4 to 6
Master Clock	МСКО	Master clock output	0		Mhz)
	PLLRC	PLL RC network input	I		
32.768 kHz clock	RTCKI	32.768 KHz clock input	I		Connected to external 32.768 Khz crystal
CIUCK	RTCKO	32.768 KHz clock output	0		
PIO	UPIO[31:0]	General purpose I/O	I/O	(Z)	
	SCK0/MPIO	USART0 clock line	I/O	(Z)	Multiplexed with general purpose I/O
USART0	RXD0/MPIO	USART0 receive line	I/O	(Z)	Multiplexed with general purpose I/O
	TXD0/MPIO	USART0 transmit line	I/O	(Z)	Multiplexed with general purpose I/O
	SCK1/MPIO	USART1 clock line	I/O	(Z)	Multiplexed with general purpose I/O
USART1	RXD1/MPIO	USART1 receive line	I/O	(Z)	Multiplexed with general purpose I/O
	TXD1/MPIO	USART1 transmit line	I/O	(Z)	Multiplexed with general purpose I/O
Capture0	CAPT0	Capture input	Ι		
Capture1	CAPT1	Capture input	Ι		
PWM	PWM[3:0]	Pulse Width Modulation output	0	(L)	





# Table 3-1. Signal Description (Continued)

Module	Name	Function	Туре	Active Level	Comments
	T0TIOA[2:0]/MPIO	Capture/waveform I/O	I/O	(Z)	Multiplexed with a general purpose I/O
Timer T0	T0TIOB[2:0]/MPIO	Trigger/waveform I/O	I/O	(Z)	Multiplexed with a general purpose I/O
	T0TIOCLK[2:0]/MPI O	External clock/trigger/input	I/O	(Z)	Multiplexed with a general purpose I/O
	T1TIOA/MPIO	Capture/waveform I/O	I/O	(Z)	Multiplexed with a general purpose I/O
Timer T1	T1TIOB/MPIO	Trigger/waveform I/O	I/O	(Z)	Multiplexed with a general purpose I/O
	T0TIOCLK/MPIO	External clock/trigger/input	I/O	(Z)	Multiplexed with a general purpose I/O
	ANA0IN[7:0]	Analog input	I		
ADC0	VREFP0	Positive voltage reference	I		
ADC1	ANA1IN[7:0]	Analog input	I		
ADCT	VREFP1	Positive voltage reference	I		
	SPCK/MPIO	SPI clock line	I/O	(Z)	Multiplexed with a general purpose I/O
	MISO/MPIO	SPI master in slave out	I/O	(Z)	Multiplexed with a general purpose I/O
SPI	MOSI/MPIO	SPI master out slave in	I/O	(Z)	Multiplexed with a general purpose I/O
	NPCS[3:1]/MPIO	SPI chip select	I/O	(Z)	Multiplexed with a general purpose I/O
	NPCS0/NSS/MPIO	SPI chip select (slave input)	I/O	(Z)	Multiplexed with a general purpose I/O
0.4.1.0	CANRX0	CAN0 receive line	I	L	
CAN0	CANTX0	CAN0 transmit line	0	L (H)	
CAN1	CANRX1	CAN1 receive line	I	L	
CANT	CANTX1	CAN1 transmit line	0	L (H)	
CAN2	CANRX2	CAN2 receive line	I	L	
CANZ	CANTX2	CAN2 transmit line	0	L (H)	
CAN3	CANRX3	CAN3 receive line	I	L	
CANS	CANTX3	CAN3 transmit line	0	L (H)	
	SCANEN	Scan enable	I	Н	Internal pull-down (connected GND or leave unconnected)
	TDI	Test Data In	I		Schmitt trigger, internal pull-up
	TDO	Test Data Out	0		
JTAG	TMS	Test Mode Select	I		Schmitt trigger, internal pull-up
	ТСК	Test Clock	I		Schmitt trigger, internal pull-up
	TEST	Factory Test	I	Н	Internal pull-down (connected GND or leave unconnected)
Core Power	VDDCORE	3.3V	-		
Supply	GND	Ground	-		

# AT91SAM7A2

Table 3-1.	Signal Description	(Continued)
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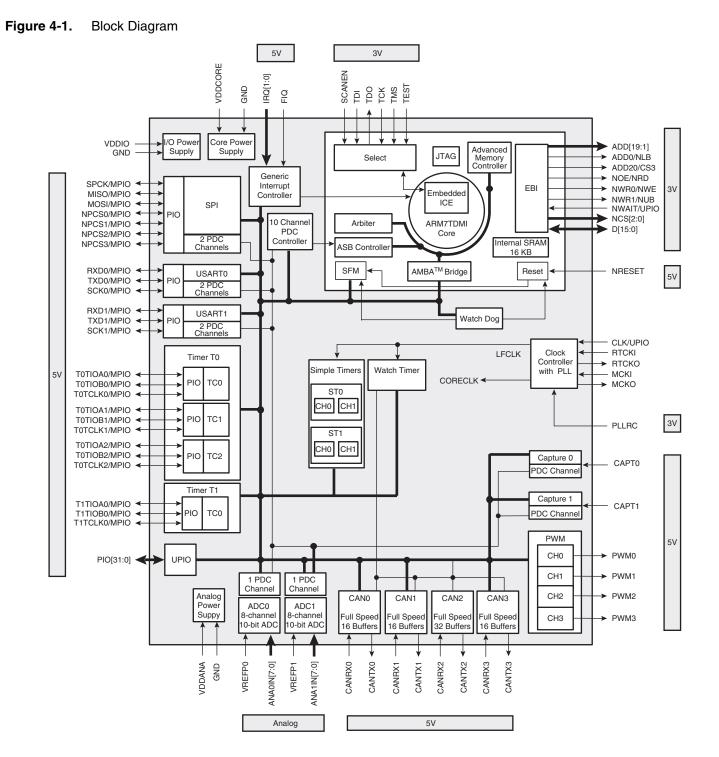
Module	Name	Function	Туре	Active Level	Comments
Analog	VDDANA	3.3V	-		
Power Supply	GND	Analog Ground	-		
I/O power	VDDIO	3.3V to 5V	-		
supply	GND	Ground	-		

Note: 1. Values in brackets are the values at reset (H = High, L = Low, Z = High Impedance State).





# 4. Block Diagram



# 5. Product Overview

#### 5.1 Register Considerations

#### 5.1.1 Enable/Disable/Status Registers

In order to reduce code size and subsequently increase speed when accessing internal peripherals, most of the registers have been split into three address locations:

- The first address location (Enable or Set Register) is used to set a bit to a logical 1.
- The second address location (Disable or Clear Register) is used to set a bit to a logical 0.
- The third address location (Status register or Mask Register) gives the current state of the bit.

To set a bit to a logical 1 in the Status or Mask Register, a write command in the Enable or Set Register must be performed with the corresponding bit at a logical 1.

To set a bit to a logical 0 in the Status or Mask Register, a write command in the Disable or Clear Register must be performed with the corresponding bit at a logical 1.

#### 5.1.2 Example

Supposing that the US0\_PSR register value is 0x00000000. To enable the RXD and SCK pins as PIOs in the USART0 block, 0x00050000 must be written in the US0\_PER register. The value read in the US0\_PSR register will be 0x00050000.

Now if the software wants to disable the RXD pin as a PIO (i.e. enable it for USART0 use), a write access to the US0\_PDR register with the value 0x00040000 must be performed. The new value read in the US0\_PSR register will be 0x00010000.

#### 5.1.3 Key Access to Registers

Some bits in registers can be set to a value (0 or 1) only if the right key is written at the same time.

#### 5.1.3.1 Example

The TESTEN bit in the SFM\_TM register can be set to a logical 0 or 1 only if the KEY[15:0] bits are equal to 0xD64A.

To enable test mode, 0xD64A0002 must be written in the SFM\_TM register.

To disable test mode, 0xD64A0000 must be written in the SFM\_TM register.

#### 5.1.4 Ghost Registers

The AT91SAM7A2 microcontroller integrates an ICE (In-Circuit Emulation) interface that is associated with a JTAG connection and a software debugger that provides powerful debug possibility.

Effectively,

- A running program can be stopped.
- Internal registers and internal/external memories can be monitored.
- Instructions can be added when the core is stopped.
- The program can be resumed.

However, some AT91SAM7A2 registers are "read-active", meaning that reading such registers can affect the state of other registers. This is usual and wanted register behavior. For example,





in the ADC module, the bit EOC (End Of Conversion) is automatically cleared when the DR (numerical value of the input converted) is read. The aim is to cut off the amount of code needed in an application.

Meanwhile, when debugging software, users can monitor the value of a register, without modifying the state of another register. For this purpose, for each module, a ghost register field has been implemented in the design. Users reading in this ghost field will not affect the value of any other register.

Ghost registers are not "read-active", and are mirrors of original registers. They are located in memory by inversing the 13th bit in the module base address. For example, base address of ADC module is 0xFFFC0000, so the ghost register's base address of ADC is 0xFFFC2000. By reading this ghost field, users do not disturb the behavior of the ADC module.

Ghost registers exist for all modules.

#### 5.1.5 Read Active Registers

The following table demonstrates the effects of the read active registers.

Module	Read-active registers	Effect
010	GIC_IVR	Clears IRQ interrupt if present at the GIC.
GIC	GIC_FVR	Clears FIQ interrupt if present at the GIC.
ADC	ADC_DR	Clears EOC bit in ADC_SR register if set.
USART	US_SR	CLears the following bits in the US_SR register if set. IDLE ENDRX ENDTX SCK TXD RXD
	US_RHR	Clears RXRDY bit in the US_SR register if set.
CAPT	CAPT_DR	Clears DATACAPT bit in the CAPT_SR register if set.
SPI	SPI_SR	Clears the following bits in the SPI_SR register if set. MODF SPIOVRE REND TEND SPCK MISO MOSI NPCS0 NPCS1 NPCS2 NPCS3
	SPI_RDR	Clears RDRF bit in the SPI_SR register if set.
PIO	UPIO_SR	Clears all bits in the UPIO_SR register if set.

Table 5-1.	Read-active Registers
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Module	Read-active registers	Effect
GPT	GPT_SR	When in capture mode, it clears the following bits in the GPT_SR register if set. COVFS LOVRS CPCS LDRAS LDRBS ETRGS TIOBS TIOAS TCLKS
	GPT_SR	When in waveform mode, it clears the following bits in the GPT_SR register if set. COVFS CPAS CPBS CPCS ETRGS TIOBS TIOAS TCLKS

#### Table 5-1. Read-active Registers

#### 5.2 Power Consumption

The power consumption is described in the specific modes of the AT91SAM7A2.

#### 5.2.1 Working Modes

The AT91SAM7A2 microcontroller provides different working modes as outlined in Table 5-2 below.

Table 5-2.	Working Modes
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Mode		Note		
Low Power Mode	LPM	The master clock oscillator. The PLL and the internal divider are switched off. The real time oscillator is enabled. The low frequency clock is selected from the real time oscillator and used as system clock (i.e. 32.768 kHz used for GIC, WD, WT, ST and any peripheral needed for interrupt generation). CORECLK = RTCK, LFCLK = RTCK.		
Slow Mode	SLM	The PLL is switched off. The system clock is the master clock (CORECLK = MCK) or the master clock divided by $\beta$ (CORECLK = MCK/ $\beta$ , $\beta$ is in the range of [2:256]).		
Operational	OPE	Master oscillator and PLL are enabled. The system clock is the clock from the PLL, CORECLK = $\alpha x$ MCK ( $\alpha$ is in the range of [x2:x20]).		

#### 5.2.2 Low Power Mode

Low power mode is defined as the state in which:

- Master clock oscillator and PLL are halted.
- Low frequency oscillator (32.768 kHz) used as internal system clock for core and all the peripherals (CORECLK = RTCK, LFCLK = RTCK)





 $V_{VDDCORE}$  = 3.3 V,  $V_{VDDIO}$  = 5 V. No loads on outputs, ground level on all inputs, 25°C, fetch out of internal RAM in ARM mode.

 Table 5-3.
 Low Power Mode Consumption.

Mode	Parameters	Typical	Max	Unit
LPM	All peripheral clocks disable, ARM clock enable		240	μA
LPM	All peripheral clocks disable, ARM clock disable		240	μA

#### 5.2.3 Slow Power Mode

Slow mode is defined as the state in which:

- Master clock oscillator is enabled, divided by β (β is in the range of [2:256]) and used as the system clock (CORECLK = MCK or MCK/β).
- The low frequency clock can still be used as low frequency clock for peripherals (LFCLK = RTCK or MCK/ $\beta$ ).

 $V_{VDDCORE}$  = 3.3 V,  $V_{VDDIO}$  = 5 V. No loads on outputs, ground level on all inputs, 25°C, oscillator 4 MHz,  $\beta$  = 256.

**Table 5-4.**Slow Mode Consumption.

Mode	Parameters	Typical	Max	Unit
SLM	All peripheral clocks disable, ARM clock enable		1140	μA
SLM	All peripheral clocks disable, ARM clock disable		1140	μA

#### 5.2.4 Operational Mode

Operational mode is defined as the state in which:

- Master clock oscillator and PLL are enabled, system clock is taken from the PLL output (CORECLK =  $\alpha$  x MCK, where  $\alpha$  is in the range of [2:20]).
- The low frequency clock can still be used as low frequency clock for peripherals (LFCLK = RTCK or MCK/ $\beta$ ,  $\beta$  is in the range of [2:256]).

The total power dissipation of the AT91SAM7A2 embedded system, when in operational mode, is estimated to be 200 mW<sup>(1)</sup> maximum, at an operating voltage of 3.3 V, over the operating temperature range.

Table 5-5.	Operational Mode Consumption.

Table F F

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Mode	Parameters	Тур	Unit
OPE	Core	900	µA/MHz
SLM	PLL (frequency independent)	1.5	mA

#### 5.2.5 Module Consumption

Initial condition: 3.3 V, 25°C, All inputs grounded, low level and no load on all outputs, ARM clock enable.

Table 5-0.					
Symbol	Parameters	Тур	Unit		
PDC	Peripheral Data controller	160	µA/MHz		
UPIO	Unified Parallel Input Output	40	µA/MHz		
USART	Universal Sync/Async Receiver Transceiver	110	µA/MHz		
SPI	Serial Peripheral Interface	60	µA/MHz		
GPT3CH	General Purpose Timer 3 Channels	150	µA/MHz		
GPT1CH	General Purpose Timer 1 Channel	40	µA/MHz		
ADC	Analog to Digital Converter	20	µA/MHz		
CAN16	CAN 16 Channels	210	µA/MHz		
CAN32	CAN 32 Channels	280	µA/MHz		
ST	Simple Timer	40	µA/MHz		
CAPT	Capture	20	µA/MHz		
PWM4C	PWM 4 Channels	60	µA/MHz		
ALL	All Modules	1650	µA/MHz		

Table 5-6. Operational Mode Consumption.

#### 5.3 Reset

To properly reset the chip, users must maintain a reset of at least 1µs.

After a reset, the program starts executing after the PLL stabilization time (11.3 ms for an oscillator of 4 MHz).

Note: 1. ARM core and modules working at CORECLK frequency = 30 MHz (i.e. MCK = 6 MHz, PLL multiplier = 5).





# 5.4 Electrical Characteristics

#### Table 5-7.Pin Connection

Pin	Name	Pad	Pin	Name	Pad	Pin	Name	Pad	Pin	Name	Pad
1	VDDIO		45	GND		89	VDDIO		133	NOE/NRD	PC3T03
2	IRQ0	MC5D00	46	VDDIO		90	VDDANA		134	NCS0	PC3T01
3	IRQ1	MC5D00	47	UPIO5	MC5B03	91	VREFP0	ANAIN	135	ADD1	PC3T01
4	FIQ	MC5D00	48	UPIO6	MC5B03	92	ANA0IN0	AIMUX1	136	D9	PC3B01D
5	SCK0/MPIO	MC5B01	49	GND		93	ANA0IN1	AIMUX1	137	D2	PC3B01D
6	TXD0/MPIO	MC5B01	50	VDDIO		94	ANA0IN2	AIMUX1	138	VDDCORE	
7	RXD0/MPIO	MC5B01	51	UPIO7	MC5B03	95	ANA0IN3	AIMUX1	139	D10	PC3B01D
8	SCK1/MPIO	MC5B01	52	UPIO8	MC5B02	96	ANA0IN4	AIMUX1	140	D3	PC3B01D
9	TXD1/MPIO	MC5B01	53	UPIO9	MC5B02	97	ANA0IN5	AIMUX1	141	D11	PC3B01D
10	RXD1/MPIO	MC5B01	54	UPIO10	MC5B02	98	ANA0IN6	AIMUX1	142	D4	PC3B01D
11	VDDCORE		55	UPIO11	MC5B02	99	GND		143	D12	PC3B01D
12	CANTX3	MC5O01	56	UPIO12	MC5B02	100	VDDANA		144	D5	PC3B01D
13	CANRX3	MC5D00	57	UPIO13	MC5B02	101	ANA01N7	AIMUX1	145	D13	PC3B01D
14	CAPT0	MC5D00	58	UPIO14	MC5B02	102	VREFP1	ANAIN	146	D6	PC3B01D
15	CAPT1	MC5D00	59	UPIO15	MC5B02	103	ANA1IN0	AIMUX1	147	D14	PC3B01D
16	SPCK/MPIO	MC5B01	60	UPIO16	MC5B01	104	ANA1IN1	AIMUX1	148	D7	PC3B01D
17	MISO/MPIO	MC5B01	61	UPIO17	MC5B01	105	ANA1IN2	AIMUX1	149	D15	PC3B01D
18	MOSI/MPIO	MC5B01	62	UPIO18	MC5B01	106	ANA1IN3	AIMUX1	150	GND	
19	NPCS0/MPIO	MC5B01	63	GND		107	ANA1IN4	AIMUX1	151	ADD0/NLB	PC3T01
20	VDDIO		64	VDDIO		108	ANA1IN5	AIMUX1	152	ADD17	PC3T01
21	GND		65	UPIO19	MC5B01	109	ANA1IN6	AIMUX1	153	ADD16	PC3T01
22	NPCS1/MPIO	MC5B01	66	UPIO20	MC5B01	110	ANA1IN7	AIMUX1	154	ADD15	PC3T01
23	NPCS2/MPIO	MC5B01	67	UPIO21	MC5B01	111	GND		155	ADD14	PC3T01
24	NPCS3/MPIO	MC5B01	68	UPIO22	MC5B01	112	VDDCORE		156	ADD13	PC3T01
25	T0TIOA0/MPIO	MC5B01	69	UPIO23	MC5B01	113	RTCKI	32.768 kHz crystal oscillator pad	157	ADD12	PC3T01
26	T0TIOB0/MPIO	MC5B01	70	UPIO24	MC5B01	114	RTCKO	32.768 kHz crystal oscillator pad	158	ADD11	PC3T01
27	T0TCLK0/MPIO	MC5B01	71	UPIO25	MC5B01	115	GND		159	ADD10	PC3T01
28	T0TIOA1/MPIO	MC5B01	72	UPIO26	MC5B01	116	VDDCORE		160	ADD9	PC3T01
29	T0TIOB1/MPIO	MC5B01	73	UPIO27	MC5B01	117	SCANEN	PC3D01D	161	ADD20/CS3	PC3T01
30	T0TCLK1/MPIO	MC5B01	74	UPIO28	MC5B01	118	TEST	PC3D01D	162	VDDCORE	
31	T0TIOA2/MPIO	MC5B01	75	UPIO29	MC5B01	119	TMS	PC3D21U	163	NWR0/NWE	PC3B01
32	T0TIOB2/MPIO	MC5B01	76	UPIO3/ NWAIT	MC5B01	120	TDO	PC3T03	164	NCS2	PC3T01
33	VDDIO		77	UPIO3/ CORECLK	MC5B01	121	TDI	PC3D21U	165	NCS1	PC3T01
34	GND		78	CANTX0	MC5O01	122	ТСК	PC3D21U	166	ADD19	PC3T01
35	T0TCLK2/MPIO	MC5B01	79	CANRX0	MC5D00	123	GND		167	ADD18	PC3T01

14 AT91SAM7A2

# AT91SAM7A2

Table 5-7. Pin Connection

Pin	Name	Pad		
36	T1TIOA0/MPIO	MC5B01		
37	T1TIOB0/MPIO	MC5B01		
38	T1TCLK0/MPIO	MC5B01		
39	NRESET	MC5D20		
40	UPIO0	MC5B04		
41	UPIO1	MC5B04		
42	UPIO2	MC5B04		
43	UPIO3	MC5B04		
44	UPIO4	MC5B03		

Pin	Name	Pad
80	CANTX1	MC5O01
81	CANRX1	MC5D00
82	CANTX2	MC5O01
83	CANRX2	MC5D00
84	PWM0	MC5O01
85	PWM1	MC5O01
86	PWM2	MC5O01
87	PWM3	MC5O01
88	GND	

			_			
Pin	Name	Pad	I	Pin	Name	Pad
124	PLLRC	PLL080M1		168	ADD8	PC3T01
125	VDDCORE			169	ADD7	PC3T01
126	MCKI	OSC16M		170	ADD6	PC3T01
127	МСКО	OSC16M		171	ADD2	PC3T01
128	GND			172	ADD3	PC3T01
129	NWR1/NUB	PC3B01		173	ADD4	PC3T01
130	D8	PC3B01D		174	ADD5	PC3T01
131	D1	PC3B01D		175	GND	
132	D0	PC3B01D		176	GND	

Pad types are given in Table 5-8 below.

Table 5-8.	Pad Types
------------	-----------

Pad	Туре	DTPDHL <sup>(1)</sup>	DTPDLH <sup>(2)</sup>	TPDHL <sup>(3)</sup>	TPDLH <sup>(4)</sup>	Output Current
MC5B01	5 V CMOS bidirectional pad	0.144 ns/pF	0.131 ns/pF	2.327 ns	2.192 ns	2 mA AC 2 mA DC
MC5B02	5 V CMOS bidirectional pad	0.072 ns/pF	0.066 ns/pF	2.298 ns	2.179 ns	4 mA AC 4 mA DC
MC5B03	5 V CMOS bidirectional pad	0.036 ns/pF	0.033 ns/pF	2.727 ns	2.034 ns	8 mA AC 8 mA DC
MC5B04	5 V CMOS bidirectional pad	0.018 ns/pF	0.017 ns/pF	3.265 ns	2.449 ns	16 mA AC 16 mA DC
MC5O01	5 V CMOS output pad	0.144 ns/pF	0.131 ns/pF	2.310 ns	2.174 ns	2 mA AC 2 mA DC
MC5D00	5 V CMOS non-inverting input pad					
MC5D20	5 V CMOS schmitt non-inverting input pad					
PC3D01D	3 V CMOS non-inverting input pad with pull- down resistor					
PC3D01U	3 V CMOS non-inverting input pad with pull- up resistor					
PC3D21	3 V CMOS schmitt non-inverting input pad					
PC3D21U	3V CMOS schmitt non-inverting input pad with pull-up resistor					
PC3T01	3 V CMOS three state output pad	0.120 ns/pF	0.116 ns/pF	1.357 ns	1.011 ns	2 mA AC 0.3 mA DC
PC3T02	3 V CMOS three state output pad	0.060 ns/pF	0.058 ns/pF	1.002 ns	0.781 ns	4 mA AC 0.3 mA DC
PC3T03	3 V CMOS three state output pad	0.040 ns/pF	0.039 ns/pF	0.943 ns	0.800 ns	6 mA AC 0.3 mA DC





#### Table 5-8. Pad Types

Pad	Туре	DTPDHL <sup>(1)</sup>	DTPDLH <sup>(2)</sup>	TPDHL <sup>(3)</sup>	TPDLH <sup>(4)</sup>	Output Current
PC3B01D	3 V CMOS bidirectional pad with pull-down resistor	0.118 ns/pF	0.116 ns/pF	1.357 ns	1.040 ns	2 mA AC 0.3 mA DC
PC3B01	3 V CMOS non-inverting bidirectional pad	0.120 ns/pF	0.116 ns/pF	1.372 ns	1.033 ns	2 mA AC 0.3 mA DC
PC3B02	3 V CMOS non-inverting bidirectional pad	0.060 ns/pF	0.058 ns/pF	1.010 ns	0.789 ns	6 mA AC 0.3 mA DC
PC3B03	3 V CMOS non-inverting bidirectional pad	0.040 ns/pF	0.039 ns/pF	0.948 ns	0.808 ns	6 mA AC 0.3 mA DC
OSCK33	32.768 kHz crystal oscillator pad					
OSC16M	2-6 MHz crystal oscillator pad					
PLL080M 1	20 MHz to 80 MHz single pad Phase- Locked Loop					
AIMUX1	Analog input pad					

Notes: 1. Differential (load-dependent) propagation delay, high-to-low or high impedance-to-low (V<sub>DD</sub> = 3.3 V, Temp. = 25°C, Input Slope = 1 ns)

 Differential (load-dependent) propagation delay, low-to-high or high impedance-to-high (V<sub>DD</sub> = 3.3 V, Temp. = 25°C, Input Slope = 1 ns)

3. Propagation delay, high-to-low ( $V_{DD}$  = 3.3 V, Temp. = 25°C, Input Slope = 1 ns)

4. Propagation delay, low-to-high ( $V_{DD}$  = 3.3 V, Temp. = 25°C, Input Slope = 1 ns)

#### 5.4.1 Propagation Time

The propagation delay time shown in Table 5-8, "Pad Types," on page 15, is the time in nanoseconds from the 50% point of the input to the 50% point of the output.

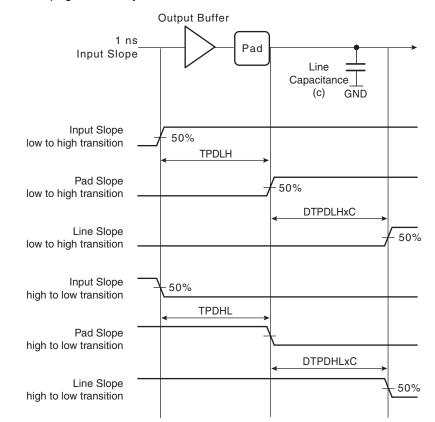


Figure 5-1. Propagation Delay





# 6. Clocks

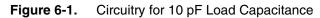
### 6.1 Overview

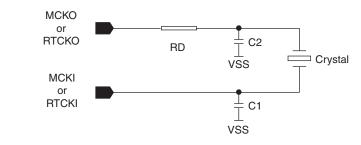
The AT91SAM7A2 microcontroller provides:

- 32.768 kHz oscillator
- 4 MHz to 8 MHz oscillator

#### 6.2 Crystals

Crystals with 10 pF load capacitance can be directly connected to the oscillator pads. Nevertheless, it is recommended to implement the circuitry as described hereafter and in Figure 6-1 below.





If the crystal load capacitance Cx is greater than 10 pF, then C1 and C2 must be added. Cx can be approximated to:  $Cx = (C1 \times C2)/(C1 + C2)$ .

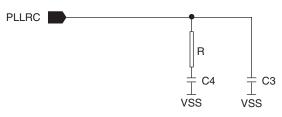
Table 6-1.	Typical Crystal Series Resistor
------------	---------------------------------

Signal	RD	Conditions
МСКО	0 Ω	Crystal: CP12A-4MHz-S1-4085-1050 (NDK <sup>®</sup> )
RTCKO	10 kΩ	Crystal: MC-306 32.768K-A (EPSON®)

#### 6.3 Phase Locked Loop

The AT91SAM7A2 microcontroller integrates a programmable PLL with a default ratio value of x5. The PLL requires an external RC network as described hereafter and in Figure 6-2 below.

Figure 6-2. External RC Network



The optimum response with a simple RC filter is obtained when:

Equation 1:  $0.4 < \sqrt{\left(\frac{K_0 \times I_P}{n \times (C_3 + C_4)}\right)} \times \frac{R \times C_4}{2} < 1 \text{ with an optimum value of } 0.707$ 

18 **AT91SAM7A2** 

Where:

- K0 is the PLL  $V_{CO}$  gain
- IP is the peak current delivered by the charge pump into the filter
- n is the division ration of the divider (i.e. PLL multiplication factor)
- Stability can be improved with an additional capacitor C3. The value of C3 must be chosen so that:

Equation 2:

$$4 < \frac{C_4}{C_3} < 15$$

Equation 3:

$$\sqrt{\left(\frac{K_0 \times I_p}{n \times (C_3 + C_4)}\right)} \leq \frac{\mathsf{II} \times f_{CKR}}{5}$$

Where:

• f<sub>CKR</sub> is the PLL input frequency (i.e. MCK): Phase jitter for the PLL is 200 ps typically.

Recommended values: R = 100 Ohm, C3 = 10 nF and C4 = 2.2 nF.

#### 6.3.1 PLL Characteristics

Table 6-2.	PLL Characteristics
------------	---------------------

Code	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>CKR</sub>	Input frequency		0.02		30	MHz
f <sub>ск</sub>	Output frequency		20		30	MHz
Wlow	Duty cycle			50		%
Јск	Jitter	With ratio 1:1		200		ps
n	Division ratio		1:1		1:1024	
K <sub>0</sub>	V <sub>CO</sub> gain		45	105	172	MHz/V
l <sub>P</sub>	CHP current		50	350	800	mA





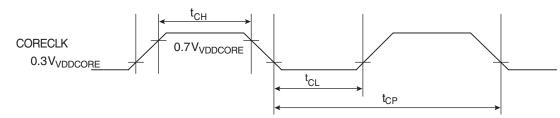
## 6.4 Clock Timings

#### 6.4.1 Core Clock

#### Table 6-3.Core Clock Timings

Symbol	Parameter	Minimum	Maximum	Unit
1/t <sub>CP</sub>	Oscillator frequency	32.768	30000	kHz
t <sub>CP</sub>	Main clock period	33		ns
t <sub>CH</sub>	Main clock high time	12		ns
t <sub>CL</sub>	Main clock low time	12		ns

#### Figure 6-3. Core Clock Waveform



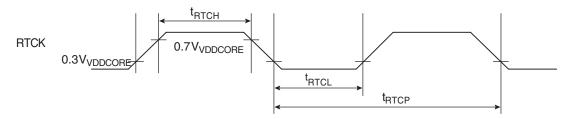
#### 6.4.2 32.768 kHz Frequency Clock

The 32.768 kHz clock is the clock generated by the real time clock oscillator. The real-time clock (RTCK) characteristics are given below in Table 6-4.

#### **Table 6-4.**Low Frequency Clock Timings

Symbol	Parameter	Minimum	Typical	Maximum	Unit
1/t <sub>RTCP</sub>	32.768 kHz oscillator frequency		32.768		kHz

#### Figure 6-4. 32.768 kHz Clock Waveform



# 6.5 Internal Oscillator Characteristics

#### 6.5.1 Core Clock Oscillator

#### Table 6-5.Core Clock Oscillator

Code	Parameter	Conditions	Min	Тур	Max	Unit
Du	Duty cycle	Crystal @ 4 MHz	40	50	60	%
Opf	Operating frequency		4		8	MHz
t <sub>SU</sub>	Startup time	Crystal @ 4 MHz			10	ms
t <sub>SU</sub>	Startup time	Crystal @ 8 MHz			5	ms
C1	Internal capacitance (MCKI/GND)			10		pF
C2	Internal capacitance (MCKO/GND)			10		pF
CL	Equivalent load capacitance (MCKI/MCKO)			5		pF
DL	Drive level				50	W
Rs	Equivalent Series Resistance	Fundamental @ 8 Mhz			100	
Rs	Equivalent Series Resistance	Fundamental @ 4 Mhz			50	
Cs	Shunt capacitance	Crystal			6	pF
CL	Load capacitance	Crystal @ 4 MHz		10		pF
Cm	Motional capacitance	Crystal @ 4 MHz			3	fF

#### 6.5.2 Real Time Clock Oscillator

#### Table 6-6.Real Time Clock Oscillator

Code	Parameter	Conditions	Min	Тур	Мах	Unit
Du	Duty cycle	@ 32.768 kHz	40	50	60	%
tsu	Startup time				1.5	S
C1	Internal capacitance (RTCKI/GND)			20		pF
C2	Internal capacitance (RTCKO/GND)			20		pF
CL	Equivalent load capacitance (RTCKI/RTCKO)			10		pF
DL	Drive level				1	W
Rs	Series resistance	Crystal			60	k
Cs	Shunt capacitance	Crystal	0.8		1.7	pF
	Load capacitance	Crystal @ 32.768 kHz		10		pF
Cm	Motional capacitance	Crystal @ 32.768 kHz	1		4	fF





# 7. Memory Map

When the AT91SAM7A2 microcontroller is reset, the ARM core is in reboot mode to access the external memory (usually ROM) on NCS0 at address 0x00000000. The internal RAM is located at address 0x00300000.

When the software executes the remap command (write 1 in RCB field in AMC\_RCR register), the internal RAM is automatically located at address 0x00000000 and the external memory accessed on the NCS0 is located in the memory space from 0x40000000 to 0x7FFFFFFF, depending on the AMC\_NCS0 register in the Advanced Memory Controller, then the chip is in remap mode.

#### 7.1 Reboot Mode

The memory map in reboot mode is described below.

Memory Space	Application	Abort
0xFFFFFFF _ 0xFFE00000	Peripheral devices	No
0xFFDFFFFF 	Reserved	Yes
0x003FFFFF _ 0x00300000	Internal RAM 16 kbytes repeated 64 times	No
0x002FFFFF _ 0x00200000	Reserved (Read as '0')	No
0x001FFFFF _ 0x00100000	Reserved	Yes
0x000FFFFF 	External memory on NCS0	No

 Table 7-1.
 Memory Map in Reboot Mode

No

#### 7.2 Remap Mode

The memory map in remap mode is described below.

**Memory Space** Application Abort 0xFFFFFFF Peripheral devices No 0xFFE00000 0xFFDFFFFF Reserved Yes 0x80000000 0x7FFFFFFF External memories (up to 4) Yes, outside of page defined in Memory values repeated within the AMC 0x4000000 the page size programmed 0x3FFFFFFF Reserved Yes 0x00300000 0x002FFFFF Reserved (Read as '0') No 0x00100000 0x000FFFFF Internal RAM 16 kbytes

**Table 7-2.**Memory Map in Remap Mode

#### 7.3 External Memory

The AT91SAM7A2 external memories can be relocated in the address space from 0x40000000 to 0x7FFFFFFF. The configuration of the base address and the page size of each EBI chip select line (NCS[2:0], CS3) is done through the Advanced Memory Controller (AMC) registers. It is to be noted that the two most significant bits of the base address are fixed to 01b, allocating these memories between 0x40000000 to 0x7FFFFFF.

repeated 64 times

The maximum external memory space is 6 Mbytes, while CS3/A20 is used as address line A20.

Table 7-3.External Memory Map

0x0000000

Memory Space	Size	Application
0x(01XXb)XXFFFFF 	Up to 1 Mbytes	External memory on CS3
0x(01XXb)X1FFFFF 0x(01XXb)XX00000	Up to 2 Mbytes	External memory on NCS2
0x(01XXb)X1FFFFF 0x(01XXb)XX00000	Up to 2 Mbytes	External memory on NCS1
0x(01XXb)X1FFFFF 	Up to 2 Mbytes	External memory on NCS0





# 7.4 Peripheral Memory

The peripheral memory map is described below.

Table 7-4. Peripheral Memory Map	Table 7-4.	Peripheral Memory Map
----------------------------------	------------	-----------------------

Peripheral	Address	IRQ
AMC	0xFFE00000	-
SFM	0xFFF00000	-
Watchdog	0xFFFA0000	2
Watch Timer	0xFFFA4000	3
USART0	0xFFFA8000	4
USART1	0xFFFAC000	5
CAN3 (16 channels)	0xFFFB0000	6
SPI	0xFFFB4000	7
CAN1 (16 channels)	0xFFFB8000	8
CAN2 (32 channels)	0xFFFBC000	9
ADC0 (8 channels 10-bit)	0xFFFC0000	10
ADC1 (8 channels 10-bit)	0xFFFC4000	11
	0xFFFC8000	12
GPT0 (3 channels)		13
		14
GPT1 (1 channel)	0xFFFCC000	18
PWM	0xFFFD0000	19
CAN0 (16 channels)	0xFFFD4000	20
UPIO	0xFFFD8000	21
Capture CAPT0	0xFFFDC000	22
Capture CAPT1	0xFFFE0000	23
Simple Timer ST0	0xFFFE4000	24
Simple Timer ST1	0xFFFE8000	25
Clock Manager	0xFFFEC000	-
PMC	0xFFFF4000	-
PDC	0xFFFF8000	-
GIC	0xFFFFF000	-

### 8. Power Management Block

In order to reduce power consumption, the AT91SAM7A2 microcontroller provides a power management block in some peripherals used to switch on/off the peripheral clocks (peripheral and PIO block).

This function is independent of the Power Management Controller (peripheral) used to switch on/off the ARM7TDMI core and the PDC clocks.

Three registers are provided:

- PERIPHERAL\_ECR (at peripheral offset 0x0050) enables the clock.
- PERIPHERAL\_DCR (at peripheral offset 0x0054) disables the clock.
- PERIPHERAL\_PMSR (at peripheral offset 0x0058) gives the status of the clock.

Two bits are provided in these registers:

- Bit 0 controls the PIO block of the peripheral.
- Bit 1 controls the peripheral function.

When the peripheral clock (and/or the PIO clock) is disabled, the clock is immediately stopped. When the clock is re-enabled, the peripheral controller (and/or the PIO controller) resumes action where it left off.

The interrupt registers are common to the peripheral controller and its PIO controller. The clock on the interrupt registers and its associated logic are stopped only if both the peripheral controller clock and the PIO controller clock are stopped.

Module	Power Management Block Present
AMC	No
SFM	No
Watchdog	No
Watch Timer	No
USART0	Yes
USART1	Yes
CAN3	Yes
SPI	Yes
ADC0	Yes
ADC1	Yes
GPT0 Ch0	Yes
GPT0 Ch1	Yes
GPT0 Ch2	Yes
GPT1 Ch0	Yes
PWM	Yes
CANO	Yes
CAN1	Yes
CAN2	Yes

**Table 8-1.**Power Management Blocks





#### Table 8-1. Power Management Blocks

Module	Power Management Block Present
UPIO	Yes
CAPT0	Yes
CAPT1	Yes
Simple Timer ST0	Yes
Simple Timer ST1	Yes
СМ	No
PMC	Yes
PDC	No
GIC	No

AT91SAM7A2

## 9. PIO Controller Block

To match different applications, the AT91SAM7A2 peripherals have their dedicated pins multiplexed with general purpose I/O pins (MPIO).

The following table lists the module sharing the dedicated pins with MPIOs.

Module	PIO Block Present	Number of MPIOs	Name of PIO Lines
AMC	No	-	-
SFM	No	-	-
Watchdog	No	-	-
Watch Timer	No	-	-
USART0	Yes	3	TXD0, RXD0, SCK0
USART1	Yes	3	TXD1, RXD1, SCK1
CAN3	No	-	-
SPI	Yes	7	MISO, MOSI, SPCK, NPCS[3:0]
ADC0	No	-	-
ADC1	No	-	-
GPT0 TC0	Yes	3	TIOA0, TIOB0, TCLK0
GPT0 TC1	Yes	3	TIOA1, TIOB1, TCLK1
GPT0 TC2	Yes	3	TIOA2, TIOB2, TCLK2
PWM	No	-	-
CAN0	No	-	-
CAN1	No	-	-
CAN2	No	-	-
UPIO	Yes	32	UPIO[31:0]
CAPT0	No	-	-
CAPT1	No	-	-
Simple Timer ST0	No	-	-
Simple Timer ST1	No	-	-
СМ	No	-	-
PMC	No	-	-
PDC	No	-	-
GIC	No	-	-
GPT1 TC0	Yes	3	TIOA, TIOB, TCLK

Table 9-1.PIO Blocks

Each PIO block in the peripheral is controlled through the peripheral interface. The PIO block clock is enabled/disabled by the peripheral power management controller. See Table 7-4, "Peripheral Memory Map," on page 24.





# **10. Multiplexed I/O Lines**

All I/O lines are multiplexed with an I/O signal of the peripheral. After reset, the pin is controlled by the peripheral PIO controller. When a peripheral signal is not used in an application, the corresponding pin can be used as a parallel I/O.

Each parallel I/O line is bidirectional, whether, the peripheral defines the signal as input or output.

Figure 10-1 "Parallel I/O Block," on page 30 shows the multiplexing of the peripheral signals with the PIO controller signal.

Each pin of the peripheral can be independently controlled using the Peripheral\_PER (PIO Enable) and Peripheral\_PDR (PIO Disable) registers.

The Peripheral\_PSR (PIO Status) register indicates whether the pin is controlled by the peripheral or by the PIO controller block.

#### 10.1 Output Selection

The user can select the direction of each individual I/O signal (input or output) using the Peripheral\_OER (Output Enable) and Peripheral\_ODR (Output Disable) registers. The output status of the I/O signal can be read in the Peripheral\_OSR (Output Status) register. The direction defined has effect only if the pin is configured to be controlled by the PIO controller block.

#### 10.2 I/O Levels

Each pin can be configured to be independently driven high or low. The level is defined in different ways, according to the following conditions.

If a pin is controlled by the PIO controller block and is defined as an output (see "Output Selection" above), the level is programmed using the Peripheral\_SODR (Set Output Data) and Peripheral\_CODR (Clear Output Data) registers. In this case, the programmed value can be read in the Peripheral\_ODSR (Output Data Status) register.

If a pin is controlled by the PIO controller block and is not defined as an output, the level is determined by the external circuit. If a pin is not controlled by the PIO controller block, the state of the pin is defined by the Peripheral controller. In all cases, the level on the pin can be read in the Peripheral\_PDSR (Pin Data Status) register.

### 10.3 Interrupts

Each PIO controller block also provides an internal interrupt signal shared with the peripheral interrupt.

Each PIO can be programmed to generate an interrupt when a level change occurs. This is controlled by the Peripheral\_IER (Interrupt Enable) and Peripheral\_IDR (Interrupt Disable) registers which enable/disable the I/O interrupt (and the peripheral interrupts) by setting/clearing the corresponding bit in the Peripheral\_IMR.

When a change in level occurs, the corresponding bit in the Peripheral\_SR (Interrupt Status) register is set whether the pin is used as a PIO or a peripheral signal and whether it is defined as input or output.

If the corresponding interrupt in Peripheral\_IMR (Interrupt Mask) register is enabled, the PIO interrupt is asserted.

The PIO interrupt is cleared when:

• a write access is performed on the Peripheral\_CISR register (with the corresponding bit set at a logical 1),

or

• a read access is performed in the Peripheral\_SR register (if no Peripheral\_CISR register is present in the peripheral).

#### 10.4 User Interface

Each individual MPIO is associated with a bit position in the PIO controller user interface registers. Each of these registers is 32 bits wide. If a parallel I/O line is not defined, writing to the corresponding bit has no effect. Undefined bits read zero.

#### 10.5 Multi-driver (Open Drain)

Each PIO can be programmed for a multi-driver option. This means that the PIO is configured as open drain (can only drive a low level) in order to support external drivers on the same pin. An external pull-up is necessary to guarantee a logic level (logical one) when the pin is not being driven. The Peripheral\_MDER (Multi-Driver Enable) and Peripheral\_MDDR (Multi-Driver Disable) registers control this option. The Multi-driver can be selected whether the I/O pin is controlled by the PIO controller or the peripheral controller. Peripheral\_MDSR (Multi-driver Status) indicates which pins are configured to support external drivers.





## 10.6 MPIO Block Diagram

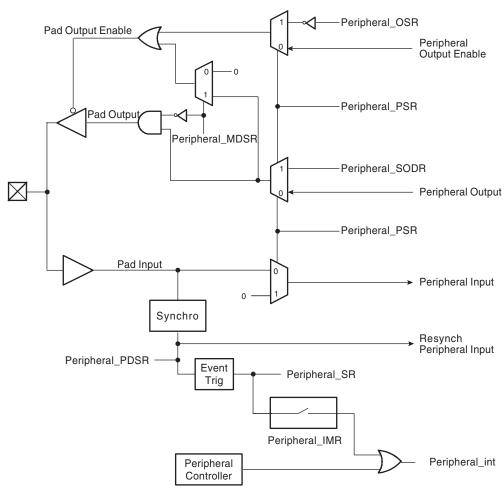


Figure 10-1. Parallel I/O Block

# 11. Advanced Memory Controller (AMC)

#### 11.1 Overview

The AT91SAM7A2 microcontroller is provided with an Advanced Memory Controller (AMC) enabling the software to configure external and internal memory mapping (at boot level).

The external 16-bit data bus interface is called the External Bus Interface (EBI) and is the physical layer used to connect external devices to the AT91SAM7A2 microcontroller. Subsequently, the EBI generates the signals which control the access to the external memory or peripheral devices.

The EBI is fully programmable through the Advanced Memory Controller and can address up to 6 Mbytes. It has four chip selects and a 20-bit address bus.

The AT91SAM7A2 can only boot on a 16-bit external memory device connected to the NCS0 signal. All the other chip select lines can be configured to access 8 or 16-bit memory devices.

#### 11.2 Boot on NCS0

Automatically, the AT91SAM7A2 boots on a 16-bit external memory device connected on NCS0.

At reset, access through NCS0 is configured as follows (in the AMC\_CSR0 register).

- 8 wait states (WSE = 1, NWS = 7)
- 16-bit data bus width
- Base address is at 0x00000000
- Byte access type is configured as Byte Write Access, BAT = 0
- The number of data float time is 0.
- The EBI is configured in normal read protocol (DRP = 0 in AMC\_MCR register).

The user can modify the chip select 0 configuration, programming the AMC\_CSR0 with exact boot memory characteristics. The base address becomes effective after the remap command (set to a logical 1 the RCB in AMC\_RCR), but the other parameters are changed immediately after the write access in the AMC\_CSR0 register.

#### 11.3 External Memory Mapping

The memory map associates the internal 32-bit address space with the external 20-bit address bus.

The memory map is defined by programming the base address and page size of the external memories.

If the physical memory device is smaller than the programmed page size, it wraps around and appears to be repeated within the page. The AMC correctly handles any valid access to the memory device within the page.

In the event of an access request to an address outside any programmed page, a data abort signal is generated. Two types of abort are possible: instruction prefetch abort and data abort. The corresponding exception vector addresses are respectively 0x0000000C and 0x00000010.

It is up to the system programmer to program the error handling routine to use in case of an abort (see the ARM7TDMI datasheet for further information).





The AT91SAM7A2 microcontroller must be wired so that the NCS0 accesses a non volatile 16bit memory as shown in Figure 11-1 "EBI Connection for External 16-bit Memory Device, 16-bit Access Only," on page 33 or Figure 11-2 "EBI Connection for 2x8-bit External Memory Devices, 16-bit Access Only," on page 33.

#### 11.4 External Memory Device Connection

Each chip select can operate with one of two different types of write access by setting the Byte Access Type bit.

1. Byte select access (BAT = 1): uses one write signal, one read signal, and two signals to select upper and lower memory bank (used for SRAM) in a 16-bit memory.

**Typically used with 16-bit memories**, except when user want to connect 2x8-bit memories in parallel, in that case seen by the AMC this is a 16-bit memory.

 Byte write access (BAT = 0): uses two write signals for selecting two different 8-bit memories and a read signal.

**Typically used with 2x8-bit memories**, this mode is used at reset to boot on the memory connected on NCS0 (Chip Select 0).

#### 11.4.1 Byte Select Access (BAT = 1)

This mode is selected by setting the BAT bit to 1 in AMC\_CSRX registers and is typically used to connect the EBI with a 16-bit memory. All 2X8-bit memories can be connected in this mode.

Users can use the upper/lower bank selection signal to get either an 8-bit or a 16-bit access.

The signal NOE is used for reading and the signal NWE is used for writing. Signals NUB (upper bank selection) and NLB (lower bank selection) are used to have a 8-bit access.

The following illustrations Figure 11-1, Figure 11-2 on page 33 and Figure 11-3 on page 34 show how to connect a typical 16-bit memory and 2x8-bit memories with 16-bit access and a 16-bit memory with 8-bit access.

#### 11.4.1.1 16-bit Access Device Connection

A typical 16-bit memory (e.g. Flash) device connection with 16-bit access is listed below, NLB and NUB are not used.

- The A0/NLB signal is not used
- The NWR1/NUB signal is not used
- The NWR0/NWE signal is used as NWE and enables half-word writes.
- The NRD/NOE signal is used as NOE and enables half-word and byte reads.

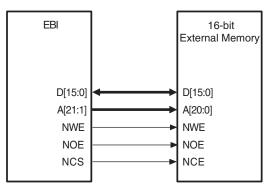
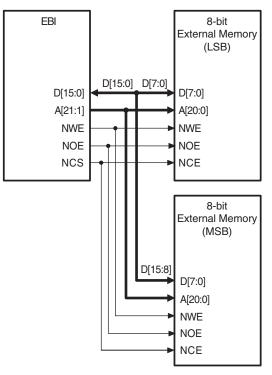


Figure 11-1. EBI Connection for External 16-bit Memory Device, 16-bit Access Only

In the same configuration as shown in Figure 11-1 above, it is possible to connect 2x8-bit memory devices with 16-bit access. The configuration shown in Figure 11-2 on page 33 demonstrates how to interface the EBI with 2x8-bit memories (for example 2x8-bit ROM memory) as a 16-bit memory page.





#### 11.4.1.2 8-bit or 16-bit Access Device Connection

A typical 16-bit memory (e.g. 16-bit SRAM) device connection with 8- or 16-bit access is shown below.

This 16-bit memory allows upper/lower bank selection and NUB, NLB are used to achieve an 8-bit access.

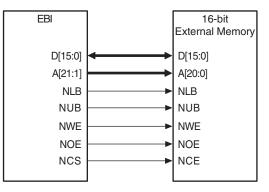
• The A0/NLB signal is used as NLB and enables the lower byte for both read and write operations.





- The NWR1/NUB signal is used as NUB and enables the upper byte for both read and write operations.
- The NWR0/NWE signal is used as NWE and enables half-word or byte writes.
- The NRD/NOE signal is used as NOE and enables half-word and byte reads.

Figure 11-3. EBI Connection for External 16-bit Memory Devices, 8- or 16-bit Access



#### 11.4.2 Byte Write Access (BAT = 0)

This mode is selected by setting the BAT bit to 0 in AMC\_CSRX registers. This is the mode selected at reset.

In this mode users can interface the EBI with one or two 8-bit memories.

If the EBI is interfaced with two 8-bit memories, then users can choose to have either an 8- or 16-bit access.

The NRD signal is used for reading and two signals are used for writing, NWR0 for lower byte writes and NWR1 for upper byte writes.

Figure 11-4 shows how to connect one and two 8-bit memories with the EBI.

The example shown in Figure 11-5 demonstrates what happens when AT91SAM7A2 boots in this mode on a 16-bit memory (type Flash).

#### 11.4.2.1 8-bit Access Device Connection

A typical 8-bit memory device connection with 8-bit access is shown here.

DBW[1:0] should be set for a 8-bit-data bus width and only NWR0 is used.

- The A0/NLB signal is used as A0.
- The NWR1/NUB signal is not used.
- The NWR0/NWE signal is used as NWR0 and enables lower byte writes.
- The NRD/NOE signal is used as NRD and enables half-word and byte reads.

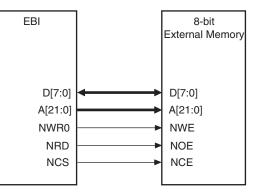
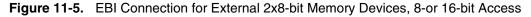


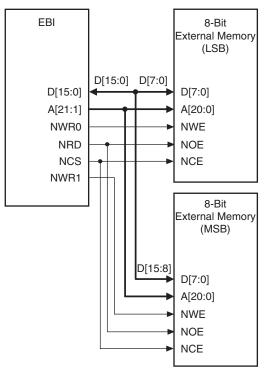
Figure 11-4. EBI Connection for External 8-bit Memory Device, 8-bit Access Only

#### 11.4.2.2 8-bit or 16-bit Access Device Connection

A typical 2x8-bit memory device connection with 8-bit or 16-bit access is shown below.

- The A0/NLB signal is not used.
- The NWR1/NUB signal is used as NWR1 and enables upper byte writes.
- The NWR0/NWE signal is used as NWR0 and enables lower byte writes.
- The NRD/NOE signal is used as NRD and enables half-word and byte reads.





#### 11.4.2.3 16-bit Access Device Connection

A typical 16-bit memory device connection with 16-bit access only is shown below.

This is typically the configuration of the memory after a reset or at power up when using a 16-bit flash memory on NCS0, in that case AT91SAM7A2 is in byte write access mode and boots on



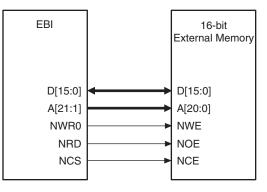


the 16-bit memory. NWR1 and NWR0 are used by the EBI but only NWR0 is used by the memory enabling a 16-bit access.

The correct mode to use with this configuration is byte select access and should be set in the boot.

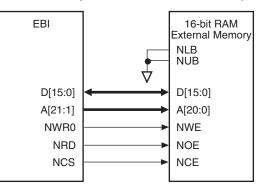
- The A0/NLB signal is not used.
- The NWR1/NUB signal is not used.
- The NWR0/NWE signal is used as NWR0 and enables half-word writes.
- The NRD/NOE signal is used as NRD and enables half-word and byte reads.

Figure 11-6. EBI Connection for External 16-bit Memory Devices, 16-bit Access Only



If users want to boot on a RAM memory for debug purposes, the RAM memory should be connected the same way as a flash memory (NUB and NLB of the RAM memory connected to the ground) to emulate a pure 16-bit Flash memory as shown in Figure 11-5.

Figure 11-7. EBI Connected to an External 16-bit RAM Memory Device, 16-bit Access Only and Used as Boot Memory for Debug Purposes



#### 11.5 External Bus Interface Timings

Simple read and write access cycles are explained in detail where read access can be done through two modes as follows:

- The standard read protocol.
- The early read protocol, which increases the EBI performance for read access.

The EBI can automatically insert wait states during the external access cycles. These wait states are applied within the actual access cycle.

AT91SAM7A2

Data float wait states can also be inserted and applied in between cycles. The data float wait states depend strongly on the previous and next access contingent upon whether the state is a write or read cycle (early or standard) and if it is on the same chip select. These conditions are detailed in the pages that follow.

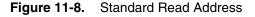
### 11.5.1 Read Access

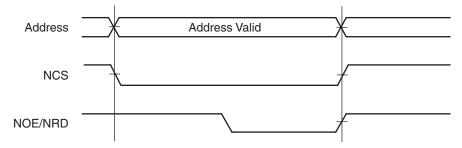
#### 11.5.1.1 Standard Read Protocol

Standard read protocol (default read mode) implements a read cycle in which NRD/NOE is active during the second half of the read cycle.

The first half of the read cycle allows time to ensure completion of the previous access as well as address and NCS output before the read cycle begins.

During a standard read protocol external memory access, NCS is set low and ADDR is valid at the beginning of the access while NRD/NOE goes low only in the second half of the read cycle to avoid bus conflict.



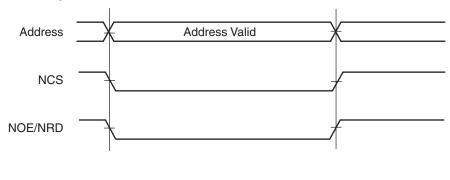


#### 11.5.1.2 Early Read Protocol

Early read protocol provides more memory access time for a read access by asserting NRD at the beginning of the read cycle. This mode is selected by setting the DRP bit in the AMC\_MCR register.

In the case of successive read cycles in the same memory, NRD remains active continuously. Since a read cycle normally limits the speed of operation of the external memory system, early read protocol can allow a faster timing of the EBI to be used. However, an extra data float wait state is required in some cases to avoid contentions on the external bus, this is explained in "Data Float Wait State" on page 40.









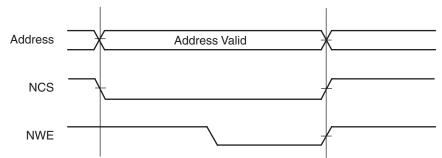
#### 11.5.2 Write Access

In a write access cycle, NWE (or NWR0, NWR1) is active during the second half of the write cycle.

The first half of the write cycle allows time to ensure completion of the previous access as well as the address and NCS set up time before NEW (or NWR0, NWR1) is asserted.

During a write external memory access, NCS is set low and ADDR is valid at the beginning of the access while NWE (or NWR0, NWR1) goes low only in the second half of the write cycle to avoid bus conflict.

#### Figure 11-10. Write Access



NWE (or NWR0, NWR1) goes high at the end of the write cycle, this is not true if a wait state is asserted.

#### 11.5.3 Wait State

Each chip select line can be programmed to insert one or more wait states during an external access. This is done by setting the WSE bit in the corresponding AMC\_CSRx register. The number of cycles to insert is programmed in the NWS[2:0] field in the same register.

The correspondence between the number of standard wait states programmed and the number of cycles during which the NWE pulse is held low is as follows:

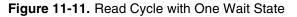
- 0 wait states 1/2 cycle
- 1 wait state 1 cycle

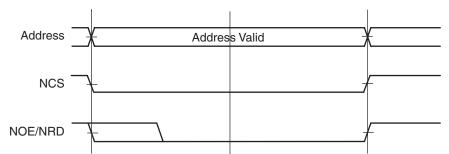
For each additional wait state programmed, an additional cycle is added.

#### 11.5.3.1 Wait state with Read Cycle

The read cycle is delayed one cycle for each wait state programmed. In early mode, NOE/NRD goes low at the start of the read cycle while in standard mode, this signal goes low at the half of the first cycle.

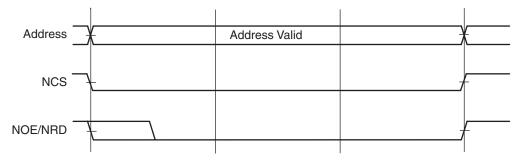
The following figure shows a read cycle with one wait state.





The following figure shows a read cycle with two wait states.

Figure 11-12. Read Cycle with Two Wait States

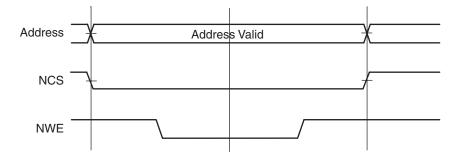


#### 11.5.3.2 Wait State with Write Cycle

The write cycle is delayed one cycle for each wait state programmed. NWE (or NWR0, NWR1) goes high one half cycle before the end of the write cycle.

The following figure shows a write cycle with one wait state.

Figure 11-13. Write Cycle with One Wait State

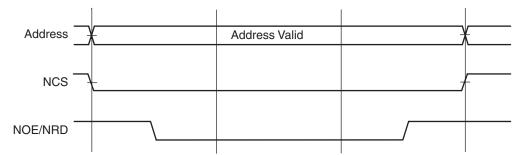


The following figure shows a write cycle with two wait states.





#### Figure 11-14. Write Cycle with Two Wait States



#### 11.5.4 Data Float Wait State

Data float wait states are added to avoid data bus conflict.

After a read access, the data float wait state gives more time for the external memory to release the data bus.

After a write access the data float wait state gives more time for the EBI to release the data bus.

The Data Float Output Time  $(t_{DF})$  for each external memory device is programmed in the TDF field of the AMC\_CSR register for the corresponding chip select. The value (0-7 clock cycles) indicates the number of data float wait states to be inserted.

Data float wait state are asserted in between accesses.

Data float wait state insertion depends strongly on the previous access and the next access contingent upon whether the state is a write or read cycle (early or standard) and if it is on the same chip select.

The following table describes the data float wait state applied.

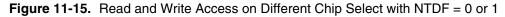
		Number of Data Flo	Number of Data Float Wait State Applied				
Previous Access	Next Access	Early Read Mode	Standard Read Mode				
NCSx Read	NCSx Read	0	0				
NCSx Read	NCSx Write	nTDF	NTDF				
NCSx Write	NCSx Read	1	0				
NCSx Write	NCSx Write	0	0				
NCSx Read	NCSy Read	Max(1, nTDFx)	Max(1, nTDFx)				
NCSx Read	NCSy Write	Max(1, nTDFx)	Max(1, nTDFx)				
NCSx Write	NCSy Read	1	1				
NCSx Write	NCSy Write	1	1				

#### Table 11-1. Data Float State Applied

Table 11-2.	Examples
-------------	----------

Previous		Early Read Mode					Standard Read Mode						
Access	Next Access	TDFx = 0	TDFx = 1	TDFx = 2	TDFx = 3	TDFx = 0	TDFx = 1	TDFx = 2	TDFx = 3				
NCSx Read	NCSx Read	0	0	0	0	0	0	0	0				
NCSx Read	NCSx Write	0	1	2	3	0	1	2	3				
NCSx Write	NCSx Read	1	1	1	1	0	0	0	0				
NCSx Write	NCSx Write	0	0	0	0	0	0	0	0				
NCSx Read	NCSy Read	1	1	2	3	1	1	2	3				
NCSx Read	NCSy Write	1	1	2	3	1	1	2	3				
NCSx Write	NCSy Read	1	1	1	1	1	1	1	1				
NCSx Write	NCSy Write	1	1	1	1	1	1	1	1				

The waveforms appearing below and on the following pages give an exhaustive description of how data float wait states apply.



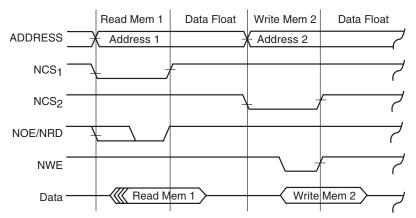
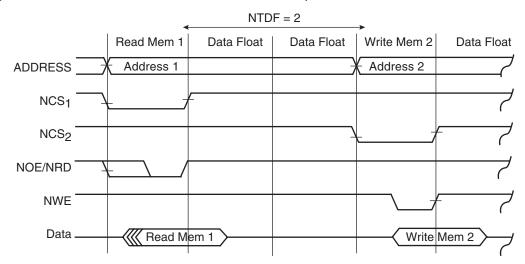


Figure 11-16. Read and Write Access on Different Chip Select with NTDF = 2







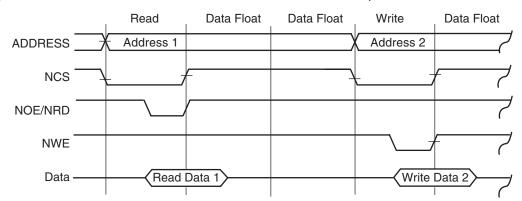


Figure 11-17. Standard Read and Write Access on the Same Chip Select with NTDF = 2

Figure 11-18. Sequential Early Read Access on the Same Chip Select with One Wait State

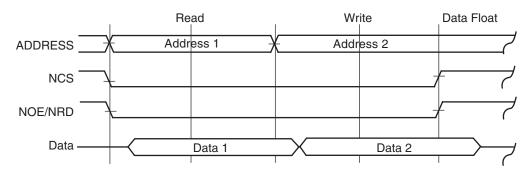
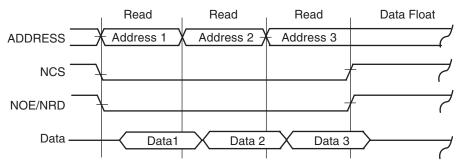


Figure 11-19. Sequential Early Read Access on the Same Chip Select with No Wait State



# AT91SAM7A2

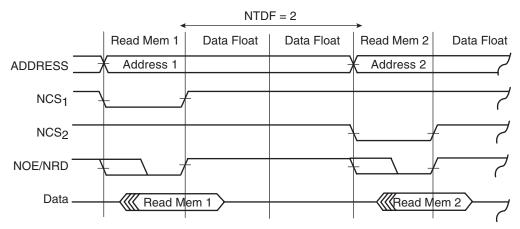


Figure 11-20. Sequential Read Access on Different Chip Select with NTDF = 2

Figure 11-21. Sequential Read Access on Different Chip Select with NTDF = 0 or 1

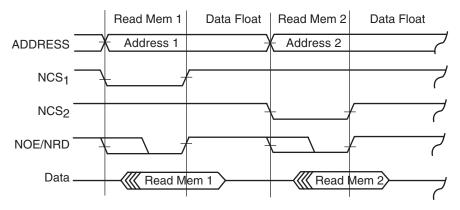
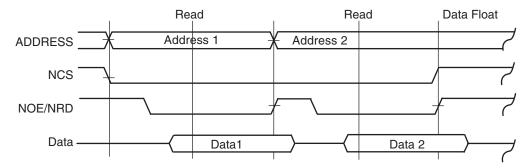


Figure 11-22. Sequential Standard Read Access on the Same Chip Select with One Wait State







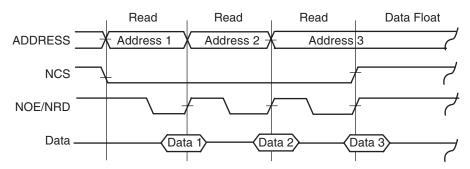
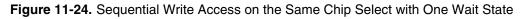


Figure 11-23. Sequential Standard Read Access on the Same Chip with No Wait State



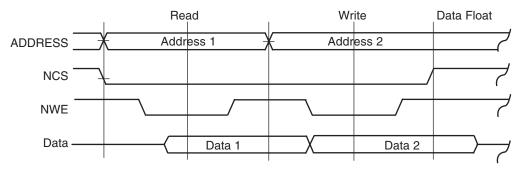


Figure 11-25. Sequential Write Access on the Same Chip Select with No Wait State

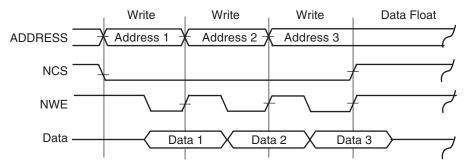
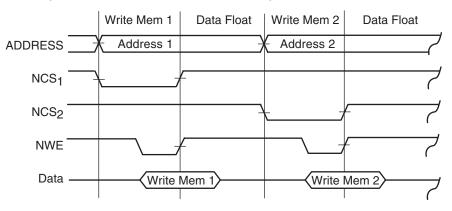


Figure 11-26. Sequential Write Access on Different Chip Select



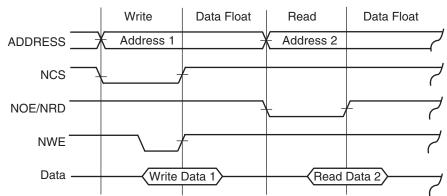


Figure 11-27. Write and Early Read on the Same Chip Select



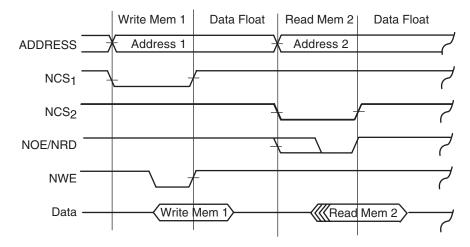
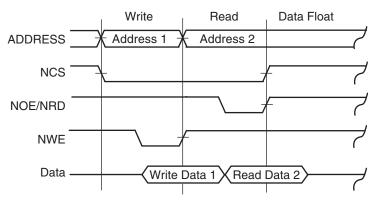


Figure 11-29. Write and Standard Read on the Same Chip Select







## 11.5.5 Timings

Table 11-3 below and Table 11-5 on page 48 show the minimum and maximum timings for external memory Read/Write cycles (valid over the recommended operating conditions) for a capacitive load of 15 pF and 30 pF. These timings are represented on the relevant waveform.

Table 11-3.         Asynchronous Timings for Read Access	ss (+85° C)
--	-------------

			Load = 15p	f		Load = 30p	f
	Read Access	MIN	ТҮР	МАХ	MIN	ТҮР	МАХ
tr <sub>ADCRDV2</sub>	Address Change to Read Data Valid (Read/Write Memory, 0 Wait State, standard read)			t <sub>CYCLE</sub> - 27 ns			t <sub>CYCLE</sub> - 31 ns
tr <sub>ADCRDV1</sub>	Address Change to Read Data Valid (All other cases)			(1 + NWS) * t <sub>CYCLE</sub> - 15 ns			(1 + NWS) * t <sub>CYCLE</sub> - 17 ns
tr <sub>CSLRDV</sub>	Chip Select Low to Read Data Valid			(1 + NWS) * t <sub>CYCLE</sub> - 14 ns			(1 + NWS) * t <sub>CYCLE</sub> - 16 ns
tr <sub>OELRDV1</sub>	Output Enable Low to Read Data Valid (standard read)			(0.5 + NWS) * t <sub>CYCLE</sub> - 12 ns			(0.5 + NWS) * t <sub>CYCLE</sub> - 13 ns
tr <sub>OELRDV2</sub>	Output Enable Low to Read Data Valid (early read)			(1 + NWS) * t <sub>CYCLE</sub> - 12 ns			(1 + NWS) * t <sub>CYCLE</sub> - 13 ns
tr <sub>BSLRDV2</sub>	Byte Select Low to Read Data Valid (Read/Write Memory, 0 Wait State, standard read)			(1 + NWS) * t <sub>CYCLE</sub> - 26 ns			(1 + NWS) * t <sub>CYCLE</sub> - 31 ns
tr <sub>BSLRDV1</sub>	Byte Select Low to Read Data Valid (all other cases)			(1 + NWS) * t <sub>CYCLE</sub> - 16 ns			(1 + NWS) * t <sub>CYCLE</sub> - 18 ns
tr <sub>DH</sub>	Data Hold-time from Address Change/NCS High/NOE High	0			0		
tr <sub>DHZOEL1</sub>	Data Hi-Z to Output Enable Low (previous is a write cycle - standard read)	0.5 * t <sub>CYCLE</sub> - 1 ns			0.5 * t <sub>CYCLE</sub> - 3 ns		
tr <sub>DHZOEL2</sub>	Data Hi-Z to Output Enable Low (previous is a write cycle - early read)	-3 ns			-4 ns		
tr <sub>DHZCSL</sub>	Data Hi-Z to Chip Select Low (previous is a write cycle - standard)	0.5 * t <sub>CYCLE</sub> + 1 ns			0.5 * t <sub>CYCLE</sub> + 1 ns		

			Load = 15pt	f	Load = 30pf			
	Read Access	MIN TYP		МАХ	MIN	ТҮР	MAX	
tr <sub>ADCRDV2</sub>	Address Change to Read Data Valid (Read /Write Memory, 0 Wait State, standard read)			t <sub>CYCLE</sub> - 29 ns			t <sub>CYCLE</sub> - 31 ns	
tr <sub>ADCRDV1</sub>	Address Change to Read Data Valid (All other cases)			(1 + NWS) * t <sub>CYCLE</sub> - 16 ns			(1 + NWS) * t <sub>CYCLE</sub> - 17 ns	
tr <sub>CSLRDV</sub>	Chip Select Low to Read Data Valid			(1 + NWS) * t <sub>CYCLE</sub> - 15 ns			(1 + NWS) * Tcycle – 16 ns	
tr <sub>OELRDV1</sub>	Output Enable Low to Read Data Valid (standard read)			(0.5 + NWS) * t <sub>CYCLE</sub> - 13 ns			(0,5 + NWS) * t <sub>CYCLE</sub> - 13 ns	
tr <sub>OELRDV2</sub>	Output Enable Low to Read Data Valid (early read)			(1 + NWS) * t <sub>CYCLE</sub> - 13 ns			(1 + NWS) * t <sub>CYCLE</sub> – 13 ns	
tr <sub>BSLRDV2</sub>	Byte Select low to read data Valid (Read /Write Memory, 0 Wait State, standard read)			(1 + NWS) * t <sub>CYCLE</sub> - 28 ns			(1 + NWS) * t <sub>CYCLE</sub> - 31 ns	
tr <sub>BSLRDV1</sub>	Byte Select low to read data Valid (All other cases)			(1 + NWS) * t <sub>CYCLE</sub> - 17 ns			(1 + NWS) * t <sub>CYCLE</sub> - 18 ns	
tr <sub>DH</sub>	Data Hold-time from Address change/NCS High/NOE High			0			0	
tr <sub>DHZOEL1</sub>	Data Hi-Z to Output Enable low (Previous is a write Cycle - standard read)			0.5 * t <sub>CYCLE</sub> - 3 ns			0.5 * t <sub>CYCLE</sub> - 3 ns	
tr <sub>DHZOEL2</sub>	Data Hi-Z to Output Enable low (Previous is a write Cycle - early read)			- 2 ns			- 4 ns	
tr <sub>DHZCSL</sub>	Data Hi-Z to Chip Select low (Previous is a write Cycle - standard)			0.5 * t <sub>CYCLE</sub> + 1 ns			0.5 * t <sub>CYCLE</sub> + 1ns	

Table 11-4.Asynchronous Timings for Read Access (+105° C)





			Load = 15pf		Load = 30pf			
	Write Access	MIN	TYPICAL	MAX	MIN	TYPICAL	МАХ	
tw <sub>ADSWL</sub>	Address/NCS/NUB/NLB Setup-time to Write Low	0.5 * t <sub>CYCLE</sub> - 1 ns			0.5 * t <sub>CYCLE</sub>			
tw <sub>WPL1</sub>	Write Pulse Low (one or more Wait States)	NWS * t <sub>CYCLE</sub> - 1 ns			NWS * t <sub>CYCLE</sub> - 1 ns			
tw <sub>WPL2</sub>	Write Pulse low (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 2 ns			0.5 * t <sub>CYCLE</sub> - 2ns			
tw <sub>DSWH1</sub>	Data Setup-time to Write High (one or more Wait States)	NWS * t <sub>CYCLE</sub>			NWS * t <sub>CYCLE</sub>			
tw <sub>DSWH2</sub>	Data Setup time to Write High (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 1 ns			0.5 * t <sub>CYCLE</sub> - 1 ns			
tw <sub>ADHWH</sub>	Address/CS/NUB/NLB Hold-time from Write High	5 ns			6 ns			
tw <sub>OEHDD</sub>	Output Enable High (previous is a read cycle) to Data Drive	0.5 * t <sub>CYCLE</sub> - 3 ns			0.5 * t <sub>CYCLE</sub> - 3 ns			
tw <sub>CSHDD</sub>	Chip Select High (previous is a read cycle) to Data Drive	1.5 * t <sub>CYCLE</sub> - 4 ns			1.5 * t <sub>CYCLE</sub> - 6 ns			
tw <sub>DHWH1</sub>	Data Hold-time from Write High (one or more Wait States)	t <sub>CYCLE</sub> - 6 ns			t <sub>CYCLE</sub> - 8 ns			
tw <sub>DHWH2</sub>	Data Hold-time from Write High (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 6 ns			0.5 * t <sub>CYCLE</sub> - 8 ns			

# Table 11-5.Asynchronous Timings for Write Access (+85° C)

			Load = 15pf		Load = 30pf			
	Write Access	MIN	TYPICAL	MAX	MIN	TYPICAL	MAX	
tw <sub>ADSWL</sub>	Address/NCS/NUB/NLB Setup-time to Write Low	0.5 * t <sub>CYCLE</sub> - 1 ns			0.5 * t <sub>CYCLE</sub>			
tw <sub>WPL1</sub>	Write Pulse Low (one or more Wait States)	NWS * t <sub>CYCLE</sub> - 1 ns			NWS * t <sub>CYCLE</sub> - 1 ns			
tw <sub>WPL2</sub>	Write Pulse low (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 2 ns			0.5 * t <sub>CYCLE</sub> - 3 ns			
tw <sub>DSWH1</sub>	Data Setup-time to Write High (one or more Wait States)	NWS * t <sub>CYCLE</sub>			NWS * t <sub>CYCLE</sub>			
tw <sub>DSWH2</sub>	Data Setup time to Write High (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 3 ns			0.5 * t <sub>CYCLE</sub> - 1 ns			
tw <sub>ADHWH</sub>	Address/CS/NUB/NLB Hold-time from Write High	5 ns			6 ns			
tw <sub>OEHDD</sub>	Output Enable High (previous is a read cycle) to Data Drive	0.5 * t <sub>CYCLE</sub> - 3 ns			0.5 * t <sub>CYCLE</sub> - 3 ns			
tw <sub>CSHDD</sub>	Chip Select High (previous is a read cycle) to Data Drive	1.5 * t <sub>CYCLE</sub> - 5 ns			1.5 * t <sub>CYCLE</sub> - 7 ns			
tw <sub>DHWH1</sub>	Data Hold-time from Write High (one or more Wait States)	t <sub>CYCLE</sub> - 7 ns			t <sub>CYCLE</sub> - 9 ns			
tw <sub>DHWH2</sub>	Data Hold-time from Write High (0 Wait State)	0.5 * t <sub>CYCLE</sub> - 6 ns			0.5 * t <sub>CYCLE</sub> - 8 ns			

# **Table 11-6.**Asynchronous Timings for Write Access $(+105^{\circ} C)$





#### Figure 11-30. Asynchronous Read Access Waveform

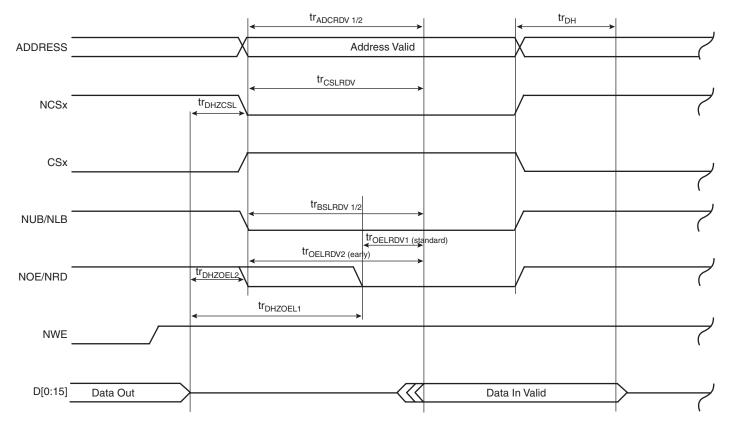
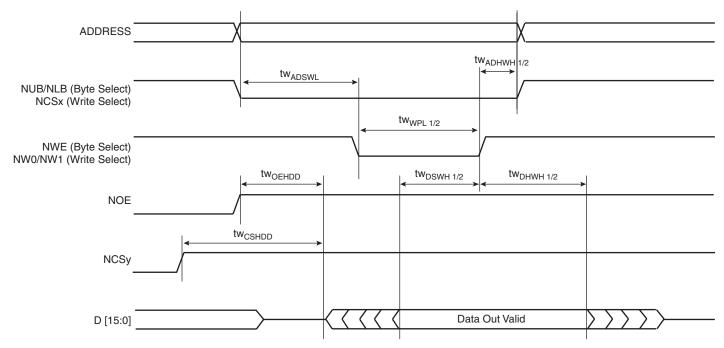


Figure 11-31. Asynchronous Write Access Waveform



		Load = 15 pf			Load = 30 pf		
	Read Access	MIN	ТҮР	МАХ	MIN	ТҮР	MAX
tr <sub>CKLCSL</sub>	Core clock low to NCS low			4 ns			4 ns
tr <sub>CKLADV</sub>	Core clock low to Address valid			5 ns			5 ns
tr <sub>CKLOEL1</sub>	Core clock low to NOE low (early Read)			2 ns			1 ns
tr <sub>CKLOE2</sub>	Core clock high to NOE low (standard read)			2 ns			1 ns
tr <sub>wasckh</sub>	NWAIT Active set-up time to Core clock high	16 ns			18 ns		
tr <sub>WNASCKH</sub>	NWAIT Not Active set-up time to Core clock high	16 ns			18 ns		
tr <sub>WAHCKH</sub>	NWAIT Active hold time from Core clock high	0 ns			0 ns		
tr <sub>DVBCKL</sub>	Data Valid before Core clock low	11 ns			13 ns		
tr <sub>DVACK</sub> I	Data Valid after Core clock low	0 ns			0 ns		

# Table 11-7.Synchronous Timings for Read Access (+85 $^{\circ}$ C)

# Table 11-8.Synchronous Timings for Read Access (+105° C)

		Load = 15 pf			Load = 30 pf		
	Read Access		ТҮР	МАХ	MIN	ТҮР	MAX
tr <sub>CKLCSL</sub>	Core clock low to NCS low			4 ns			4 ns
tr <sub>CKLADV</sub>	Core clock low to Address valid			5 ns			5 ns
tr <sub>CKLOEL1</sub>	Core clock low to NOE low (early Read)			2 ns			1 ns
tr <sub>CKLOE2</sub>	Core clock high to NOE low (standard read)			2 ns			1 ns
tr <sub>WASCKH</sub>	NWAIT Active set-up time to Core clock high	17 ns			19 ns		
tr <sub>WNASCKH</sub>	NWAIT Not Active set-up time to Core clock high	17 ns			19 ns		
tr <sub>wahCKH</sub>	NWAIT Active hold time from Core clock high	0 ns			0 ns		
tr <sub>DVBCKL</sub>	Data Valid before Core clock low	12 ns			14 ns		
tr <sub>DVACK</sub> I	Data Valid after Core clock low	0 ns			0 ns		





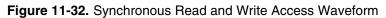
# Table 11-9.Synchronous Timings for Write Access (+85° C)

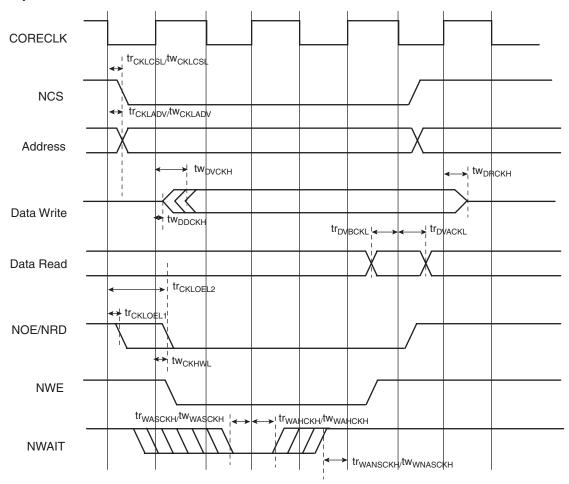
		Load = 15 pf			Load = 30 pf		
	Read Access	MIN	ТҮР	МАХ	MIN	ТҮР	MAX
tw <sub>CKLCSL</sub>	Core clock low to NCS low			4 ns			4 ns
tw <sub>CKLADV</sub>	Core clock low to Address valid			5 ns			5 ns
tw <sub>CKHWL</sub>	Core clock low to NWE low			6 ns			6 ns
tw <sub>WASCKH</sub>	NWAIT Active set-up time to Core clock high	16 ns			18 ns		
tw <sub>WNASCKH</sub>	NWAIT Not Active set-up time to Core clock high	16 ns			18 ns		
tw <sub>WAHCKH</sub>	NWAIT Active hold time from Core clock high	0 ns			0 ns		
tw <sub>DVCKH</sub>	Data Valid from Core clock high			5 ns			5 ns
tw <sub>DRCKH</sub>	Data Release from Core clock high			3 ns			3 ns
tw <sub>DDCKH</sub>	Data Drive from Core clock high	-2 nsns			-4 ns		

# Table 11-10. Synchronous Timings for Write Access (+105 $^{\circ}$ C)

		L	oad = 15 p	of	Load = 30 pf		
	Read Access	MIN	ТҮР	МАХ	MIN	ТҮР	MAX
tw <sub>CKLCSL</sub>	Core clock low to NCS low			4 ns			4 ns
tw <sub>CKLADV</sub>	Core clock low to Address valid			5 ns			5 ns
tw <sub>CKHWL</sub>	Core clock low to NWE low			6 ns			6 ns
tw <sub>WASCKH</sub>	NWAIT Active set-up time to Core clock high	17 ns			18 ns		
tw <sub>WNASCKH</sub>	NWAIT Not Active set-up time to Core clock high	17 ns			19 ns		
tw <sub>WAHCKH</sub>	NWAIT Active hold time from Core clock high	0 ns			0 ns		
tw <sub>DVCKH</sub>	Data Valid from Core clock high			5 ns			5 ns
tw <sub>DRCKH</sub>	Data Release from Core clock high			4 ns			3 ns
tw <sub>DDCKH</sub>	Data Drive from Core clock high	-2 nsns			-4 ns		

AT91SAM7A2









# 11.6 Advanced Memory Controller (AMC) Memory Map

Base Address: 0xFFE0000

Offset	Register	Name Acce		Reset State
0x000	AMC Chip Select Register 0	AMC_CSR0	Read/Write	0x4000203D
0x004	AMC Chip Select Register 1	AMC_CSR1	Read/Write	0x48000000
0x008	AMC Chip Select Register 2	AMC_CSR2	Read/Write	0x5000000
0x00C - 0x018	Reserved	_	_	_
0x01C	AMC Chip Select Register 3	AMC_CSR3	Read/Write	0x78000000
0x020	AMC Remap Control Register	AMC_RCR	Read/Write	0x0000000
0x024	AMC Memory Control Register	AMC_MCR	Read/Write	0x0000000

Table 11-11. AMC Memory Map  $^{(1)}$   $^{(2)}$ 

Notes: 1. The software must set the AMC Registers for correct operation.

2. In register tables, individual bits are defined: W: Write, R: Read, -0: 0 after reset, -1: 1 after reset, -U: undefined after reset.

# 11.7 AMC Chip Select Register 0

Name: Access: Base Address:	AMC_C Read/W 0x000						
31	30	29	28	27	26	25	24
-	—			BA	[9:4]		
23	22	21	20	19	18	17	16
	BA	[3:0]		-	_	-	_
15	14	13	12	11	10	9	8
-	_	CSEN	BAT		TDF[2:0]		PAGES1
7	6	5	4	3	2	1	0
PAGES0	_	WSE		NWS[2:0]		DB	W[1:0]

### • BA[9:0]: Base Address

Bits [31:30] are set by hardware, so the base address can only be in the memory space 0x40000000-0x7FFFFFF. The other bits contain the highest bits of the base address. If the page size is larger than 1Mbyte, the unused bits of the base address are ignored by the AMC decoder.

## • CSEN: Chip Select Enable

0: Chip select is disabled.

1: Chip select is enabled.

### • BAT: Byte Access Type

0: Byte write access type.

1: Byte select access type.

# • TDF[2:0]: Data Float Output Time

These bits select the number of cycles added after a memory transfer.

TDF[2:0]			Cycles Added
0	0	0	0
0	0	1	1
0	1	0	2
0	1	1	3
1	0	0	4
1	0	1	5
1	1	0	6
1	1	1	7





## • PAGES[1:0]: Page Size

These bits select the memory page size.

PAGES[1:0]		Page Size	Active Bits in Base Address
0	0	1 Mbytes	12 (31-20)
0	1	4 Mbytes	10 (31-22)
1	0	16 Mbytes	8 (31-24)
1	1	64 Mbytes	6 (31-26)

#### • WSE: Wait State Enable

0: Wait state generation is disabled. No wait state is inserted.

1: Wait state generation is enabled.

### • NWS[2:0]: Number of Wait States

These bits select the number of wait states added. This field is only valid if the WSE bit is set.

	NWS[2:0]				
0	0	0	1		
0	0	1	2		
0	1	0	3		
0	1	1	4		
1	0	0	5		
1	0	1	6		
1	1	0	7		
1	1	1	8		

# • DBW[1:0]: Data Bus Width

Type of data bus selected.

DBW[1:0]		Data Bus Width		
0	0	Reserved		
0	1	16-bit Data Bus		
1	0	8-bit Data Bus		
1	1	Reserved		

# 11.8 AMC Chip Select Register

Name: Access: Base Address:	- AMC_C Read/V	CSR1AMC_CS Vrite 0x008, 0x01C	SR3				
31	30	29	28	27	26	25	24
-	_			BA	[9:4]		
23	22	21	20	19	18	17	16
	BA	[3:0]		-	_	_	_
15	14	13	12	11	10	9	8
_	_	CSEN	BAT		TDF[2:0]		PAGES1
7	6	5	4	3	2	1	0
PAGES0	_	WSE		NWS[2:0]		DB	W[1:0]

#### • BA[9:0]: Base Address

Bits [31:30] are set by hardware, so the base address can only be in the memory space 0x40000000-0x7FFFFFF. The other bits contain the highest bits of the base address. If the page size is larger than 1Mbyte, the unused bits of the base address are ignored by the AMC decoder.

#### • CSEN: Chip Select Enable

0: Chip select is disabled.

1: Chip select is enabled.

### • BAT: Byte Access Type

0: Byte write access type.

1: Byte select access type.

## • TDF[2:0]: Data Float Output Time

These bits select the number of cycles added after a memory transfer.

TDF[2:0]			Cycles Added
0	0	0	0
0	0	1	1
0	1	0	2
0	1	1	3
1	0	0	4
1	0	1	5
1	1	0	6
1	1	1	7





## • PAGES[1:0]: Page Size

These bits select the memory page size.

PAGES[1:0]		Page Size	Active Bits in Base Address
0	0	1 Mbytes	12 (31-20)
0	1	4 Mbytes	10 (31-22)
1	0	16 Mbytes	8 (31-24)
1	1	64 Mbytes	6 (31-26)

#### • WSE: Wait State Enable

0: Wait state generation is disabled. No wait state is inserted.

1: Wait state generation is enabled.

#### • NWS[2:0]: Number of Wait States

These bits select the number of wait states added. This field is only valid if the WSE bit is set.

NWS	5[2:0]	WS A	dded
0	0	0	1
0	0	1	2
0	1	0	3
0	1	1	4
1	0	0	5
1	0	1	6
1	1	0	7
1	1	1	8

# • DBW[1:0]: Data Bus Width

Type of data bus selected.

DBW	Data Bus Width	
0	0	Reserved
0	1	16-bit Data Bus
1	0	8-bit Data Bus
1	1	Reserved

# 11.9 AMC Remap Control Register

Name: Access: Base Address:	AMC_R Read/W 0x020						
31	30	29	28	27	26	25	24
_	_	_	-	_	-	_	_
23	22	21	20	19	18	17	16
15	14	13	12	11	10	9	8
_	_	-	-	-	_	_	-
7	6	5	4	3	2	1	0
_	_	-	-	—	—	_	RCB

### • RCB: Remap Command Bit

0: No effect

1: Cancel the remapping (performed at reset) of the two memory devices (internal RAM and external memory on NCS0).

This bit is read at a logical 0 during remapping and read at logical 1 when remapping has been canceled.

# 11.10 AMC Memory Control Register

Name: Access: Base Address:	AMC_M Read/W 0x024						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	DRP	-		ALE[2:0]	

# • DRP: Data Read Protocol

0: standard read protocol for all external memory devices enabled.

1: early read protocol for all external memory devices enabled.

### • ALE[2:0]: Address Line Enable





These bits indicate the number of valid chip select lines.

ALE2	ALE1	ALE0	Valid Address Bits	Maximum Addressable Space per Chip Select Line	Valid Chip Select
0	х	х	ADD[20:0]	2 Mbytes	NCS[2:0]
1	0	х	ADD[20:0]	2 Mbytes	NCS[2:0]
1	1	0	ADD[20:0]	2 Mbytes	NCS[2:0]
1	1	1	ADD[19:0]	1 Mbytes	NCS[2:0], CS3

# 12. Clock Manager (CM)

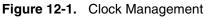
# 12.1 Overview

The AT91SAM7A2 microcontroller provides:

- 32.768 kHz Oscillator (real time clock oscillator)
- 4 MHz to 8 MHz Oscillator
- Programmable PLL (x2 to x20)
- Programmable Master Clock Divider

The clock management is done through the Clock Manager (CM). This allows the user to select between the different working modes; LPM, SLM and OPE.

At power up, the master clock oscillator and the real time clock oscillator are enabled. As the application can use (or not) the real time clock oscillator, the DIVCLK clock is used as the low frequency clock (LFCLK). This ensures that both the CORECLK and the LFCLK clock can be used at power up. The master clock (MCK) is multiplied by 10 through the PLL and divided by 2 giving a core clock frequency (CORECLK) equal to MCK x 5.



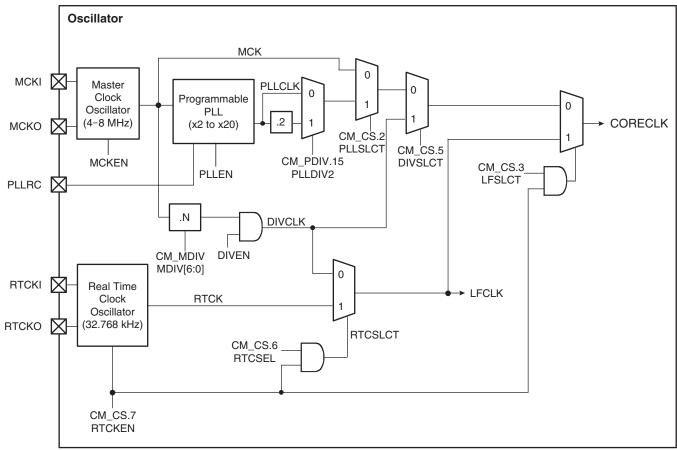






Table 12-1 below lists the different working modes according to value set in the CM\_CS register. Line 14 show values at reset.

Table 12-1. Clock Selection

Mode Numbe r	CM_C S.7RT CKEN	CM_C S.6RT CSEL	CM_C S.3 LFSLC T	CM_C S.5DIV SLCT	CM_C S.2PL LSLCT	CM_C S.1 PLLDI V2	CM_C S.8MC KEN	CM_C S.9 PLLEN	CM_CS. 13DIVE N	CM_CS .14RTC SLCT	COREC LK	LFCL K	Mo de
1	0	х	х	1	х	х	1	0	1	0	DIVCLK	DIVCL K	SL M
2	0	х	х	0	0	х	1	0	1	0	МСК	DIVCL K	SL M
3	0	х	х	0	1	0	1	1	1	0	PLLCLK	DIVCL K	OP E
4	0	х	х	0	1	1	1	1	1	0	PLLCLK /2	DIVCL K	OP E
5	1	1	1	х	х	х	0	0	0	1	RTCK	RTCK	LP M
6	1	1	0	1	х	х	1	0	1	1	DIVCLK	RTCK	SL M
7	1	1	0	0	0	х	1	0	0	1	MCK	RTCK	SL M
8	1	1	0	0	1	0	1	1	0	1	PLLCLK	RTCK	OP E
9	1	1	0	0	1	1	1	1	0	1	PLLCLK /2	RTCK	OP E
10	1	0	1	х	х	х	1	0	1	0	DIVCLK	DIVCL K	SL M
11	1	0	0	1	х	х	1	0	1	0	DIVCLK	DIVCL K	SL M
12	1	0	0	0	0	х	1	0	1	0	MCK	DIVCL K	SL M
13	1	0	0	0	1	0	1	1	1	0	PLLCLK	DIVCL K	OP E
14	1	0	0	0	1	1	1	1	1	0	PLLCLK /2	DIVCL K	OP E

For each mode listed in Table 12-1 above, the corresponding equation is calculated for LFCLK and CORECLK as shown below in Table 12-2.

Table 12-2.	Corresponding Equation for LFCLK and CORECLK
-------------	--

Mode Number	CORECLK	LFCLK		
1	MCK/(2*(MDIV+1))	MCK/(2*(MDIV+1))		
2	МСК	MCK/(2*(MDIV+1))		
3	MCK*PLL	MCK/(2*(MDIV+1))		
4	MCK*PLL/DIV2	MCK/(2*(MDIV+1))		
5	RTCK	RTCK		
6	MCK/(2*(MDIV+1))	RTCK		

Mode Number	CORECLK	LFCLK
7	MCK	RTCK
8	MCK*PLL	RTCK
9	MCK*PLL/DIV2	RTCK
10	MCK/(2*(MDIV+1))	MCK/(2*(MDIV+1))
11	MCK/(2*(MDIV+1))	MCK/(2*(MDIV+1))
12	МСК	MCK/(2*(MDIV+1))
13	MCK*PLL	MCK/(2*(MDIV+1))
14	MCK*PLL/DIV2	MCK/(2*(MDIV+1))

#### **Table 12-2.** Corresponding Equation for LFCLK and CORECLK

#### 12.1.1 Example

To switch from the default mode, OPE, to the LPM mode, do the following:

Configuration:

- 1. Enable the real time clock oscillator and the RTCSEL switch by writing the RTCKEN, RTCSEL bits and CLKEKEY field in the CM\_CE register.
- 2. Then wait that the RTCSEL status flag is set in the CE\_CS register. Once this bit is set, LFCLK clock is derived from the RCTK clock (see Figure 12-1 on page 61).
- Then set the LFSLCT bit in the CM\_CE register to obtain a core clock derived from RTCK.

To change the PLL multiplier from the default mode OPE, do the following:

- Disable the PLL by writing the PLLSLCT bit in CM\_CD. This this enables writing in the CM\_PDIV and CM\_PST registers. Now the core clock is equal to the master oscillator clock.
- 2. Then change the PLL multiplier and the divider by 2 in the CM\_PDIV register. Users can also optimize the stabilization time of the PLL of CM\_PST. This time is used in the next step when enabling the PLL. During that time the core clock will be cut.
- Enable the PLL by writing the PLLSLCT bit in the CM\_CE register. After this action, the core clock will be stopped during the time described by the PSTB field in the CM\_PST register.





# 12.2 Clock Manager (CM) Memory Map

Base Address: 0xFFEC000

# Table 12-3. Clock Manager Memory Map

Offset	Register	Name	Access	Reset State
0x000	CM Clock Enable	CM_CE	Write-only	-
0x004	CM Clock Disable	CM_CD	Write-only	-
0x008	CM Clock Status	CM_CS	Read-only	0x00002384
0x00C	CM PLL Stabilization Time	CM_PST	Read/Write	0x00000B0
0x010	CM PLL Divider	CM_PDIV	Read/Write	0x0000800A
0x014	CM Oscillator Stabilization Time	CM_OST	Read/Write	0x00000B0
0x018	CM Master Clock Divider	CM_MDIV	Read/Write	0x0000001F

# 12.3 CM Clock Enable Register

Name: Access: Base Address:	CM_CE Write-or 0x000						
31	30	29	28	27	26	25	24
			CLKEYK	EY[15:8]			
23	22	21	20	19	18	17	16
			CLKEK	EY[7:0]			
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
RTCKEN	RTCSEL	DIVSLCT	_	LFSLCT	PLLSLCT	_	_

## • CLKEKEY[15:0]: Key for Write Access into the CM\_CE Register

Any write in the CM\_CD register bits will be effective only if CLKEKEY[15:0] is equal to 0x2305.

# 12.4 CM Clock Disable Register

Name: Access: Base Address:	CM_CD Write-or 0x004								
31	30	29	28	27	26	25	24		
	CLKEYKEY[15:8]								
23	22	21	20	19	18	17	16		
			CLKEK	EY[7:0]					
15	14	13	12	11	10	9	8		
-	_	-	-	-	-	-	-		
7	6	5	4	3	2	1	0		
RTCKEN	RTCSEL	DIVSLCT	_	LFSLCT	PLLSLCT	_	—		

# • CLKDKEY[15:0]: Key for Write Access into the CM\_CD Register

Any write in the CM\_CD register bits will be effective only if CLKDKEY[15:0] is equal to 0x1807.





# 12.5 CM Clock Status Register

Name: Access: Base Address:	CM_CE Read-or 0x008						
31	30	29	28	27	26	25	24
-	_	—	—	—	-	_	-
23	22	21	20	19	18	17	16
-	_	-	-	-	—	-	-
15	14	13	12	11	10	9	8
-	RTCSLCT	DIVEN	-	-	—	PLLEN	MCKEN
7	6	5	4	3	2	1	0
RTCKEN	RTCSEL	DIVSLCT	-	LFSLCT	PLLSLCT	-	-

#### PLLSLCT: PLL/Master Clock Selection

0: Selects MCK clock (deselects PLLCLK or PLLCLK/2 clock).

1: Selects PLLCLK or PLLCLK/2 clock (deselects MCK clock).

#### • LFSLCT: Low Frequency Clock Selection

0: Allows selection of MCK, PLLCLK, PLLCK/2 or DIVCLK.

1: Selects low frequency clock LFCLK (also disables master clock oscillator and PLL).

### • DIVSLCT: Programmable Clock Selection

0: Allows selection of MCK, PLLCK or PLLCLK/2 (also deselects the DIVCLK clock).

1: Selects DIVCLK i.e. MCK divided by MDIV[6:0] (also deselects the master clock or PLL clock).

#### • RTCSEL: RTC frequency clock selection

0: Selects the DIVCLK clock for low power clock (deselects the RTCK clock).

1: Selects the RTCK clock for low power clock (deselects the DIVCLK clock).

### • RTCKEN: Low Frequency Clock Oscillator

- 0: The low frequency clock oscillator is disabled.
- 1: The low frequency clock oscillator is enabled.

#### • MCKEN: Master Clock Oscillator Enable

0: MCKEN signal is at a logical 0. The master clock oscillator is disabled and bypassed.

1: MCKEN signal is at a logical 1. The master clock oscillator is activated.

#### • PLLEN: PLL Enable

0: PLLEN signal is at a logical 0. PLL is deactivated.

1: PLLEN signal is at a logical 1. PLL is enabled.

#### • DIVEN: Programmable Divider Enable

0: DIVEN signal is at a logical 0. The programmable divider is disabled.

# 66 AT91SAM7A2

1: DIVEN signal is at a logical 1. The programmable divider is enabled.

# • RTCSLCT: Low Frequency Clock Selection

- 0: RTCSLCT signal is at a logical 0. The DIVCLK is selected for LFCLK.
- 1: RTCSLCT signal is at a logical 1. The RTCK is selected for LFCLK.





Name: Access: Base Address:	CM_PS Read/W : 0x00C								
31	30	29	28	27	26	25	24		
PSTKEY[15:8]									
23	22	21	20	19	18	17	16		
			PSTK	EY[7:0]					
15	14	13	12	11	10	9	8		
-	_	_	-	-	-	PSTB[9:8]			
7	6	5	4	3	2	1	0		
	PSTB[7:0]								

# 12.6 CM PLL Stabilization Timer Register

### • PSTB[9:0]: PLL Stabilization Time

Number of clock cycles needed before PLL stabilization. This register gives the possibility to optimize the PLL stabilization time when changing the PLL multiplier. The PLL must be disabled before writing this register.

This stabilization time is used when the PLL is enabled, after a reset and at power up, during the stabilization time the clock is halted.

The default value is 0x000000B0 guarantying 176 clock cycles with MCK/256 clock (i.e. Tsetup = 176x(1/4MHz)\*256 = 11.264 ms with MCK = 4.0 MHz). This default value includes the stabilization time for the oscillator and the PLL at start up and after a reset.

When the clock manager is configured in LPM mode and the program enables both the PLL and the master oscillator, PSTB should include the oscillator stabilization time and the PLL stabilization time. This is due to the fact that PSTB counter and OSTB counter decrement in parallel.

The PLL transient behavior before mathematical locking (phase error between the reference signal and derived signal less than  $\pm 2\pi$ ) is complex and difficult to describe using simple mathematical expression. Thus, there is no general formula giving the set-up time for any step-response transient behavior that unlocks the loop. Nevertheless, this set-up time can be approximated by a simple loop filter capacitor charging time Tsetup in the worst case:

Tsetup 
$$\leq \alpha \cdot \left[\frac{C1 + C2}{I_P}\right] \cdot \frac{VDDPLL}{2}$$

where:

C1 and C2 are the loop filter capacitors,

 $I_{p}$  the charge pump current (see "PLL Characteristics" on page 19),

 $\alpha$  is a margin factor, set to 3 or 4 as a minimum,

VDDPLL / 2 (approximately equal to 1.6V) is chosen because the PLL's VCO operates linearly.

This formula over estimates the required time, but gives an easy way to approximate this setup time.

- PLL stabilization time will be effective when PLL value is modified.
- During PLL stabilization time, CORECLOCK is stopped.

AT91SAM7A2

### • PSTKEY[15:0]: Key for Write Access into the CM\_PST Register

Any write in PSTB[9:0] will be effective only if PSTKEY[15:0] is equal to 0x59C1. These bits are always read at 0.

Note: Write accesses to this register are only valid if PLLEN is at logical 0 (i.e. PLL not enabled).

# 12.7 CM PLL Divider Register

Name: Access: Base Address:	CM_PD Read/W 0x010							
31	30	29	28	27	26	25	24	
PDIVKEY[15:8]								
23	22	21	20	19	18	17	16	
			PDIVK	EY[7:0]				
15	14	13	12	11	10	9	8	
PLLDIV2	_	-	-	-	-	—	-	
7	6	5	4	3	2	1	0	
-	_	-	PMUL[4:0]					

### • PMUL[4:0]: PLL Multiplier

These bits select the PLL multiplier.

PMUL[4:0]	PLL multiplier			
0	Remains in previous state			
1	Remains in previous state			
2	2			
3	3			
19	19			
20	20			
21 to 31	Remains in previous state			

### • PLLDIV2: PLL Divider

0: Selects PLLCLK clock (deselects PLLCLK/2)

1: Selects PLLCLK/2 clock (deselects PLLCLK)

# • PDIVKEY[15:0]: Key for Write Access into the CM\_PDIV Register

Any write in the PMUL[4:0] and PLLDIV2 bits will only be effective if the PDIVKEY[15:0] is equal to 0x762D. These bits are always read at 0.

The output frequency of the PLL is equal to: MCK x PMUL[4:0], where MCK is the PLL input clock.

Note: Write accesses to this register are only valid if PLLEN is at logical 0 (i.e. PLL disabled).





# 12.8 CM Oscillator Stabilization Timer Register

Name: Access: Base Address:	CM_OS Read/W 0x014							
31	30	29	28	27	26	25	24	
OSTKEY[15:8]								
23	22	21	20	19	18	17	16	
OSTKEY[7:0]								
15	14	13	12	11	10	9	8	
-	_	_	-	_	_	OSTB[9:8]		
7	6	5	4	3	2	1	0	
OSTB[7:0]								

#### • OSTB[9:0]: Oscillator Stabilization Time

Number of clock cycles needed before master oscillator stabilization. This register provides optimization of the oscillator stabilization time during the stabilization time the clock is halted.

This stabilization time is used when changing from low power mode (Core clock derived from RTCK) to slow mode (PLL not used).

The required time for master oscillator stabilization is 4 ms maximum and is based on the MCK/256 clock.

The default value is 0x000000B0 guarantying 176 clock cycles with MCK/256 clock (i.e. Tsetup =  $176x(1/4MHz)^{*}256 = 11.264$  ms with MCK = 4.0 MHz)

The oscillator stabilization time can be estimated at  $32/F_{osc}$  ms with  $F_{osc}$  in MHz (i.e. Tsetup = 32/4 = 8 ms with MCK = 4.0 MHz).

### • OSTKEY[15:0]: Key for Write Access into the CM\_OST Register

Any write in the OSTB[9:0] bits will only be effective if the OSTKEY[15:0] bits are equal to 0xDB5A. These bits are always read at 0.

Name: Access: Base Address:	CM_MD	CM_MDIV Read/Write 0x00C							
31	30	29	28	27	26	25	24		
			MDIVKI	EY[15:8]					
23	22	21	20	19	18	17	16		
			MDIVK	EY[7:0]					
15	14	13	12	11	10	9	8		
-	-	-	-	-	-	-	-		
7	6	5	4	3	2	1	0		
-	MDIV[6:0]								

# 12.9 CM Master Clock Divider Register

### • MDIV[6:0]: Master Clock Divider

MDIV[6:0] is used to divide the MCK clock and generate the DIVCLK. Default value for MDIV[6:0] is 0x1F.

$$\mathsf{DIVCLK} = \frac{\mathsf{MCK}}{2 \times (\mathsf{MDIV}[6:0]+1)}$$

# • MDIVKEY[15:0]: Key for Write Access into the CM\_MDIV Register

Any write in the MDIV[6:0] bits will only be effective if the MDIVKEY[15:0] bits are equal to 0xACDC. These bits are always read at 0.





# 13. Special Function Mode (SFM)

# 13.1 Overview

The AT91SAM7A2 provides registers which implement the following special functions:

- Chip identification
- Reset status

# 13.2 Chip Identification

Chip identification is done via the Chip ID register (SFM\_CIDR). This register gives information on the internal memories used (type and size) and the architecture of the device.

## 13.3 Reset status

The AT91SAM7A2 includes the Reset Status (SFM\_RSR) register to give the last cause of reset (i.e. hardware reset or internal watchdog reset).

# 13.4 Special Function Mode (SFM) Memory Map

Base Address: 0xFFF0000

### Table 13-1. SFM Memory Map

Offset	Register	Name	Access	Reset State
0x000	SFM Chip ID	SFM_CIDR	Read-only	0x80000500
Reserved	-	_	_	_
0x008	SFM Reset Status	SFM_RSR	Read-only	0x0000006C or 0x00000053

13.5 SFM C Name: Access: Base Address:	SFM_CII Read-on	DR					
31	30	29	28	27	26	25	24
EXT	-	-	-	_	—	—	-
23	22	21	20	19	18	17	16
					ARCI	H[3:0]	
15	14	13	12	11	10	9	8
	NVPMT[3:0]				IRS	[3:0]	
7	6	5	4	3	2	1	0
	NVDM	S[3:0]			NVPN	IS[3:0]	

### • NVPMS[3:0]: Non Volatile Program Memory Size

0000<sub>b</sub>: None.

Other: memory size =  $2^{(14+NVPMS[3:0])}$  bytes.

### • NVDMS[3:0]: Non Volatile Data Memory Size

0000<sub>b</sub>: None.

Other: Reserved.

### • IRS[3:0]: Internal RAM Size

Internal RAM size = 2<sup>(9+IRS[3:0])</sup> bytes.

# • NVPMT[3:0]: Non Volatile Program Memory Type

0000<sub>b</sub>: ROM less.

0001<sub>b</sub>: Mask ROM.

Other: reserved.

### • ARCH[3:0]: Core Architecture

0000<sub>b</sub>: ARM7TDMI.

Other: Reserved.

### • EXT: Extension Flag

0: No extended chip ID.

1: Extended chip ID existing.





# 13.6 SFM Reset Status Register

Name: Access: Base Address:	SFM_R Read-or 0x008						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	_	
23	22	21	20	19	18	17	16
-	_	-	-	-	-	_	
15	14	13	12	11	10	9	8
-	_	-	-	-	-	_	
7	6	5	4	3	2	1	0
			RESI	ET[7:0]			

This register gives the last cause of reset.

# • RESET[7:0]: Cause of Reset

0x6C: External reset on NRESET pin.

0x53: Internal watchdog reset.

# 14. Watchdog (WD)

### 14.1 Overview

The watchdog timer is used to prevent the system from locking-up (for example in infinite software loops). If the software does not write to the watchdog during the programmed time, then it can generate an interrupt (WDOVF) or an internal reset.

The watchdog timer has a programmable 16-bit down counter in WD\_MR register.

The software can control the action to perform when the WD counter overflows (i.e. reaches 0):

- If the RSTEN bit is set in the WD\_OMR register, an internal reset is generated.
- If the WDOVF bit is set in the WD\_IMR register, an interrupt is generated on the Generic Interrupt Controller (GIC).

In this case, an "overflow" occurs when the watchdog down-counter reaches zero.

The low frequency clock from the clock manager supplies the watchdog counter (see Figure 14-1 "Watchdog Block Diagram," on page 76).

It is possible to set a programmable pending window that provides users with the option to restart the watchdog counter only from within this window. This protection is set with the RSTALW bit, otherwise users can restart the watchdog counter at any time. When the pending windows is reached, the WDPEND bit is set followed by the PENDING bit.

The WDPDIVCLK is then divided by the WDPDIV[2:0] divider and provided to the down-counter input WDCLK.

All write accesses are protected by control access keys to help prevent corruption of the watchdog should an error condition occur.

To update the contents of the mode and control registers it is necessary to write the correct bit pattern to the control access key bits at the same time as the control bits are written (the same write access).

**Note:** Due to internal synchronization of the restart command (write restart key in the WD\_CR register), no further restart commands can be taken in account during 2 WDCLK and ½ LFCLK periods.

### 14.2 Architecture

The WD contains a programmable length down-counter. The count length determines the timeout period, and is controlled by loading the PCV field of the WD\_MR register. The time out period (in seconds) is:

$$\frac{(\text{PCV[15:0]}) + 1}{WDCLK_{\text{freg}}}$$

When the counter reaches the value programmed in the pending window PWL[15:0] of WD\_PWR register, the watchdog can generate a watchdog pending interrupt. The pending interrupt occurs after:

 $\frac{(PCV[15:0]) - (PWL[15:0])}{WDCLK_{freq}} seconds$ 

If the previous time is negative, the WD pending interrupt should not be used.





In order to prevent an internal reset (if the RSTEN bit is set in the WD\_OMR register) or interrupt (if the WDOVF bit is set in the WD\_IMR), the software must reset the counter before it reaches 0 by writing the correct key in the WD\_CR register (0xC071). The time (in seconds) between the WD pending interrupt and the WD overflow is:

When the counter reaches 0, it triggers the programmed action (internal reset or interrupt).

If no WD reset is programmed (i.e. RSTEN is at a logical 0) when the WD reaches 0, it is reset to the programmed value and continues to count, unless it is disabled. This enables it to be used to generate periodic interrupts.

### 14.3 Block Diagram

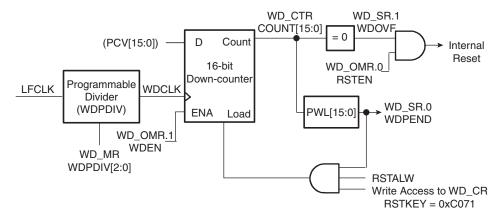


Figure 14-1. Watchdog Block Diagram

### 14.3.1 Internal Reset Pulse Generation

If the RSTEN bit is set in the WD\_OMR register, an internal reset pulse is generated when the overflow occurs. This pulse is 8 clock cycles long (CORECLK) and is not affected by hardware reset.

After a reset (hardware or WD), the clock selected by the watchdog is LFCLK/2.

### 14.3.2 Internal Interrupt Request

The watchdog can generate an internal interrupt request when an overflow occurs. The software can enable or disable this interrupt either in the WD module (WD\_IMR register) or in the GIC registers.

### 14.4 Example

The LFCLK clock source from the clock manager is used to select the WDSCLK.

The following example uses LFCLK as a value of 32.768 kHz and a pending window of 0xFF.

WD Counter Start Value	WDPDIV[2:0]	WDCLK	Time to WD Pending	Time to WD Overflow
0x07FF	000 <sub>b</sub>	LFCLK/2	0.109 s	0.125 s
0x0FFF	001 <sub>b</sub>	LFCLK/4	0.468 s	0.500 s
0x17FF	010 <sub>b</sub>	LFCLK/8	1.437 s	1.500 s
0xAFFF	101 <sub>b</sub>	LFCLK/128	175 s	176 s
0xFFFF	111 <sub>b</sub>	LFCLK/1024	1040 s	2048 s

**Table 14-1.**Watch Dog Example

### 14.4.1 Example use of the Watchdog

Use of the Pending Window to generate an interrupt and reload the watchdog counter within the window only.

If the interrupt is not considered due to a bug, a reset occurs when the watchdog counter reaches 0.

### 14.4.2 Configuration

- Configuration of WD\_MR: Choice of the clock to decrement counter, preload value from which the counter starts to decrement.
- Configuration of WD\_ PWR: Upper limit of the window from which it generates an interrupt when reached and the bit which restarts the counter only within this window.
- Configuration of WD\_IER: Enable Interrupt at the peripheral level when the window is reached (WDPEND bit) or when the counter overflow (WDOVF bit if watchdog reset is not enabled), GIC must be configured.
- Configuration of WD\_OMR: Enable the watchdog (start decrementing the counter) and enable the watchdog reset in case of counter overflow.

### 14.4.3 Interrupt Handling

- IRQ Entry and call C function.
- Read WD\_SR and verify the source of the interrupt.
- Clear the corresponding interrupt at peripheral level by writing in the WD\_CSR.
- Interrupt treatment. If this is a pending window interrupt, restart the watchdog by writing in WD\_CR.
- IRQ Exit.





# 14.5 Watchdog (WD) Memory Map

Base Address: 0xFFA0000

Table 14-2.	Watchdog Memory Map
-------------	---------------------

Offset	Register	Name	Access	Reset State
0x000 _ 0x05C	Reserved	_	-	_
0x060	Control Register	WD_CR	Write-only	_
0x064	Mode Register	WD_MR	Read/Write	0x0007FF00
0x068	Overflow Mode Register	WD_OMR	Read/Write	0x0000000
0x06C	Clear Status Register	WD_CSR	Write-only	_
0x070	Status Register	WD_SR	Read-only	0x00000000
0x074	Interrupt Enable Register	WD_IER	Write-only	-
0x078	Interrupt Disable Register	WD_IDR	Write-only	_
0x07C	Interrupt Mask Register	WD_IMR	Read-only	0x00000000
0x080	Pending Window Register	WD_PWR	Read-only	0x00000000

### 14.6 WD Control Register

Name: Access: Base Address:	WD_CR Write-or 0x060	ł					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	
23	22	21	20	19	18	17	16
-	_	—	-	-	_	—	
15	14	13	12	11	10	9	8
			RSTKE	EY[15:8]			
7	6	5	4	3	2	1	0
			RSTK	EY[7:0]			

### • RSTKEY[15:0]: Restart Key

0xC071: Watchdog counter is restarted if its value is equal or less than the pending window length or if the pending window is disabled.

Other value: No effect.

**Note:** A restart command (write the restart key in WD\_CR register) will not be effective if it occurs less than 2 WDCLK and ½ LFCLK periods after a previous start command.

# 14.7 WD Mode Register

Name: Access: Base Address:	WD_MR Read/Wri 0x064						
31	30	29	28	27	26	25	24
			CKE	Y[7:0]			
23	22	21	20	19	18	17	16
			PC\	/[7:0]			
15	14	13	12	11	10	9	8
			PC\	/[7:0]			
7	6	5	4	3	2	1	0
—	—	_	_	_		WDPDIV[2:0]	

### • WDPDIV[2:0]: WD Clock Divider

	WDPDIV[2:0]					
0	0	0	LFCLK/2			
0	0	1	LFCLK /4			
0	1	0	LFCLK /8			
0	1	1	LFCLK /16			
1	0	0	LFCLK /32			
1	0	1	LFCLK /128			
1	1	0	LFCLK /256			
1	1	1	LFCLK /1024			

# • PCV[15:0]: Preload Counter Value

Counter is preloaded when watchdog counter is restarted.

### • CKEY[7:0]: Clock Access Key

Used only when writing in WD\_MR. CKEY is read as 0.

Write access in WD\_MR is allowed only if CKEY[7:0] = 0x37.





# 14.8 WD Overflow Mode Register

Name: Access: Base Address:	WD_ON Read/W 0x068						
31	30	29	28	27	26	25	24
-	-	_	-	_	—	-	—
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
			OKE	/[11:4]			
7	6	5	4	3	2	1	0
	OKE	Y[3:0]		-	—	TSTEN	WDEN

### • WDEN: Watchdog Enable

0: Watchdog is disabled.

1: Watchdog is enabled.

### • RSTEN: Reset Enable

0: Generation of an internal reset by the Watchdog is disabled.

1: When overflow occurs, the Watchdog generates an internal reset.

### • OKEY[11:0]: Overflow Access Key

Used only when writing WD\_OMR. OKEY is read as 0.

0x234: Write access in WD\_OMR is allowed.

Other value: Write access in WD\_OMR is prohibited.

# 14.9 WD Clear Status Register

Name: Access: Base Address:	WD_CS Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	_	_	_	—
15	14	13	12	. 11	10	9	8
-	_	_	_	-	_	-	—
7	6	5	4	3	2	1	0
-	—	_	-	-	_	WDOVF	WDPEND

### • WDPEND: Watchdog Pending Clear

0: No effect.

1: Clear Watchdog pending interrupt.

### • WDOVF: Watchdog Overflow Clear

0: No effect.

1: Clear Watchdog overflow interrupt.





# 14.10 WD Status Register

Name: Access: Base Address:	WD_SR Read-or : 0x070						
31	30	29	28	27	26	25	24
—	_	_	-	_	_	—	-
23	22	21	20	19	18	17	16
-	-	-	-	-	—	-	-
15	14	13	12	11	10	9	8
_	-	-	-	-	-	RESTART	PENDING
7	6	5	4	3	2	1	0
-	-	-	-	-	-	WDOVF	WDPEND

### • WDPEND: Watchdog Pending

0: No Watchdog pending.

1: A Watchdog pending has occurred.

### • WDOVF: Watchdog Overflow

0: No Watchdog overflow.

1: A Watchdog overflow has occurred.

### • PENDING: Watchdog Pending Status

0: Watchdog counter is over pending window length.

1: Watchdog counter is equal or less than pending window length.

### • RESTART: Watchdog Restart Status

0: Watchdog available for new restart.

1: Watchdog restart executing.

# 14.11 WD Interrupt Enable Register

Name: Access: Base Address:	WD_IEF Write-or 0x074						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	-	-	-	_	WDOVF	WDPEND

# 14.12 WD Interrupt Disable Register

Name: Access: Base Address:	WD_IDF Write-or 0x078						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	-	-	-	-	WDOVF	WDPEND





# 14.13 WD Interrupt Mask Register

Name: Access: Base Address:	WD_IMI Read-or 0x07C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	_	-	-	-	WDOVF	WDPEND

### • WDPEND: Watchdog Pending Interrupt Mask

0: The WDPEND interrupt is disabled.

1: The WDPEND interrupt is enabled.

### • WDOVF: Watchdog Overflow Interrupt Mask

0: The WDOVF interrupt is disabled.

1: The WDOVF interrupt is enabled.

# 14.14 WD Pending Window Register

Name: Access: Base Address:	WD_PW Read/W 0x080	/R					
31	30	29	28	27	26	25	24
			PWK	EY[7:0]			
23	22	21	20	19	18	17	16
			PWI	_[158]			
15	14	13	12	11	10	9	8
			PW	L[7:0]			
7	6	5	4	3	2	1	0
-	-	-	-	—	-	-	RSTALW

### • RSTALW: Restart Allowed

0: restart allowed every time.

1: restart allowed only within Pending Window.

This bit does not disable the bit WDPEND in the interrupt register.

### • PWL[15:0]: Pending Window Length

Length of the window.

### • PWKEY[7:0]: Pending Window Access Key

Used only when writing in WD\_PWR. PWKEY is read as 0.

Write access in WD\_PWR is allowed only if PWKEY[7:0] = 0x91.





# 15. Watch Timer (WT)

### 15.1 Overview

The Watch Timer provides a seconds counter and an alarm function.

### 15.1.1 Seconds Counter

The seconds counter is a 32-bit counter that indicates the number of low frequency clock (LFCLK) pulses elapsed since the last time it was reset to zero.

The seconds counter is incremented every 30.518 µs (one period of 32 KHz clock).

The counter is reset to 0x00000000 when the counter reaches 0xA8C00000 (86400 seconds or 24 hours) or 0xFFFFFFF (it is configurable on the Mode Register).

A write access can only be performed when the seconds counter is disabled because of an asynchronous interface (see "Asynchronous Interface" below).

### 15.1.2 Alarm

The alarm register has the same resolution as the seconds counter. This enables a 32-bit register to have sufficient range to cater for a 24 hour period.

An interrupt is generated at the end of the period at which the value in the seconds register equals the value in the alarm register.

A write access can only be performed when the alarm counter is disabled because of an asynchronous interface. An invalid data (i.e. value greater or equal to 0xA8C00000) will not be written into the alarm register in 24 hours mode.

### 15.1.3 Asynchronous Interface

As the seconds counter is an asynchronous counter (can use the RTCK clock), some precautions must be taken with it.

When enabling or disabling the alarm or seconds counter, software must wait for an enabled or disabled interrupt to be sure that the alarm or the counter is really enabled or disabled.

### 15.1.4 CAN Time Stamp

The 32-bit register forming the seconds counter is provided to the CAN module. After each transmission or reception of a CAN frame, the value of the current seconds counter will be automatically written in the corresponding CAN channel CAN\_STPx register.

### 15.2 Example

An example use of the Watch Timer: Use of the Watch Timer to generate an interrupt after 24 hours. The low frequency clock should be 32 KHz.

### 15.2.1 Configuration

- Do a software reset of the watch timer to be in a known state by writing SWRST bit in WT\_CR and wait about four LFCLK periods for the circuitry to be stabilized.
- Configuration of WT\_MR: Selects the 24 hour mode by writing the SECRST bit. The seconds counter will increment and reset when equal to the value of 0xA8BFFFF.
- Configuration of WT\_ ALARM: When the seconds counter is equal to this value, an interrupt can be generated. For 24 hours at 32 KHz, the Alarm value should be 0xA8BFFFF.

# 86 AT91SAM7A2 I

- Configuration of WT\_SECS: Starting value from which the seconds counter will start to increment. Should be left to 0.
- Configuration of WT\_IER: The ALARM bit enables an interrupt at the peripheral level when the seconds counter is equal to the programmed value in WT\_ALARM. Other interrupts can be activated to indicate when the seconds counter or alarm functionality are really enabled or disabled, as the watch timer is clocked on the LFCLK. GIC must be configured.
- Configuration of WT\_CR: Starts the seconds counter and enables the alarm (ALARMEN and SECSEN bits). The counter will start to increment when the SECSEN bit is set in WT\_SR. The same is true for the alarm functionality bit, ALARMEN. An interrupt can be programmed with these events.

### 15.2.2 Interrupt Handling

- IRQ Entry and call C function.
- Read WT\_SR and verify the source of the interrupt.
- Clear the corresponding interrupt at the peripheral level by writing in WD\_CSR.
- Interrupt treatment: The seconds counter will restart automatically, counting from 0 when it reaches 0xA8BFFFFF (WT\_MR). If the user wants to set a lower value in WT\_ALARM and desires to restart the seconds counter; the seconds counter must first be disabled, set to 0 and once again enabled.
- IRQ Exit.

### 15.3 Watch Timer (WT) Memory Map

Base Address: 0xFFA4000

 Table 15-1.
 Watch Timer Memory Map

Offset	Register	Name	Access	Reset State
0x000 _ 0x05C	Reserved	_	_	_
0x060	Control Register	WT_CR	Write-only	_
0x064	Mode Register	WT_MR	Read/Write	0x0000000
0x068	Reserved	_	_	_
0x06C	Clear Status Register	WT_CSR	Write-only	_
0x070	Status Register	WT_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	WT_IER	Write-only	_
0x078	Interrupt Disable Register	WT_IDR	Write-only	_
0x07C	Interrupt Mask Register	WT_IMR	Read-only	0x00000000
0x080	Seconds Register	WT_SECR	Read/Write	0x0000000
0x084	Alarm Register	WT_ALR	Read/Write	0x00000000





# 15.4 WT Control Register

Name: Access: Base Address:	WT_CF Write-or 0x060						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	_	-
15	14	13	12	11	10	9	8
-	_	-	-	_	_	_	-
7	6	5	4	3	2	1	0
-	_	_	ALARMDIS	ALARMEN	SECSDIS	SECSEN	SWRST

### • SWRST: WT Software Reset

0: No effect.

### 1: Reset the WT.

A Software triggered hardware reset of the WT is performed. It resets all the registers. The software must wait for LFCLK to set up properly before using other registers.

### • SECSEN: WT Seconds Counter Enable

- 0: No effect.
- 1: Enables the WT seconds counter.

### • SECSDIS: WT Seconds Counter Disable

- 0: No effect.
- 1: Disables the WT seconds counter.

In case both SECSEN and SECSDIS are equal to one when the control register is written, the WT seconds counter will be disabled.

### • ALARMEN: WT Alarm Enable

- 0: No effect.
- 1: Enables the WT alarm.

### • ALARMDIS: WT Alarm Disable

- 0: No effect.
- 1: Disables the WT alarm.

In case both ALARMEN and ALARMDIS are equal to one when the control register is written, the WT alarm will be disabled

# 15.5 WT Mode Register

Name: Access: Base Address:	WT_MF Read/W : 0x064						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	_	_	_	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	—	-
7	6	5	4	3	2	1	0
-	_	-	-	_	_	-	SECRST

### • SECRST: Second Reset

0: The seconds counter is reset to 0x00000000 at the end of the period when it reaches 0xA8BFFFFF.

1: The seconds counter is reset to 0x00000000 at the end of the period when it reaches 0xFFFFFFF.





# 15.6 WT Clear Status Register

Name: Access: Base Address:	WT_CS Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	_
23	22	21	20	19	18	17	16
-	_	-	-	-	_	-	-
15	14	13	12	11	10	9	8
-	—	—	—	-	Ι	-	-
7	6	5	4	3	2	1	W O
_	_	_	ALARMDIS	ALARMEN	SECSDIS	SECSEN	ALARM

### • ALARM: Clear Alarm Interrupt

0: No effect.

1: Clear ALARM interrupt.

### SECSEN: Clear Seconds Counter Enabled

0: No effect.

1: Clear the seconds counter enabled interrupt.

### • SECSDIS: Clear Seconds Counter Disabled

0: No effect.

1: Clear the seconds counter disabled interrupt.

### • ALARMEN: Clear Alarm Enabled

0: No effect.

1: Clear the alarm enabled interrupt.

### • ALARMDIS: Clear Alarm Disabled

0: No effect.

1: Clear the alarm disabled interrupt.

# 15.7 WT Status Register

Name: Access: Base Address:	WT_SR Read-or 0x070						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	ALARMENS	SECENS
7	6	5	4	3	2	1	0
-	_	WSEC	ALARMDIS	ALARMEN	SECSDIS	SECSEN	ALARM

### • ALARM: Alarm Interrupt

0: No alarm occurred.

1: An alarm occurred since last clear of the status register.

### • SECSEN: Seconds Counter Enabled Interrupt

0: No seconds counter enabled interrupt.

1: A seconds counter enabled interrupt occurred since last clear of the status register.

### • SECSDIS: Seconds Counter Disabled Interrupt

0: No seconds counter disabled interrupt.

1: A seconds counter disabled interrupt occurred since last clear of the status register.

### • ALARMEN: Alarm Enabled Interrupt

0: No alarm enabled interrupt.

1: An alarm enabled interrupt occurred since last clear of the status register.

### • ALARMDIS: Alarm Disabled Interrupt

0: No alarm disabled interrupt.

1: An alarm disabled interrupt occurred since last clear of the status register.

### • WSEC: Write Second

0: No effect.

1: A write is occurring on the seconds counter register.

### • SECSENS: Seconds Counter Enable Status

- 0: Seconds counter is disabled.
- 1: Seconds counter is enabled.

### • ALARMENS: Alarm Enable Status

0: Alarm is disabled.





1: Alarm is enabled.

# 15.8 WT Interrupt Enable Register

Name: Access: Base Address:	WT_IEF Write-or 0x074						
31	30	29	28	27	26	25	24
_	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	—	—	-	—	—	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	—	ALARMDIS	ALARMEN	SECSDIS	SECSEN	ALARM

# 15.9 WT Interrupt Disable Register

Name: Access: Base Address:	WT_IMF Write-or	3					
31	30	29	28	27	26	25	24
-	_	_	-	_	-	-	-
23	22	21	20	19	18	17	16
_	_	_	_	_	_	_	_
15	14	13	12	11	10	9	8
-	_	-	-	—	-	-	-
7	6	5	4	3	2	1	0
—	_	-	ALARMDIS	ALARMEN	SECSDIS	SECSEN	ALARM

# 15.10 WT Interrupt Mask Register

Name: Access: Base Address:	WT_IMF Read-or	7					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
_	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	ALARMDIS	ALARMEN	SECSDIS	SECSEN	ALARM

### ALARM: Alarm Interrupt Mask

0: ALARM interrupt is disabled.

1: ALARM interrupt is enabled.

### • SECSEN: Seconds Counter Enabled Interrupt Mask

- 0: SECSEN interrupt is disabled.
- 1: Seconds counter enabled interrupt is enabled.

### • SECSDIS: Seconds Counter Disabled Interrupt Mask

- 0: SECSDIS interrupt is disabled.
- 1: SECSDIS interrupt is enabled.

### • ALARMEN: Alarm Enabled Interrupt Mask

0: ALARMEN interrupt is disabled.

1: ALARMEN interrupt is enabled.

### • ALARMDIS: Alarm Disabled Interrupt Mask

- 0: ALARMDIS interrupt is disabled.
- 1: ALARMDIS interrupt is enabled.



### 15.11 WT Seconds Register

Name: Access: Base Address:	WT_SECR Read/Write 0x080						
31	30	29	28	27	26	25	24
			SECOND	S[31:24]			
23	22	21	20	19	18	17	16
			SECOND	S[23:16]			
15	14	13	12	11	10	9	8
			SECON	DS[15:8]			
7	6	5	4	3	2	1	0
			SECON	DS[7:0]			

### • SECONDS[31:0]: Seconds Register

Number of LFCLK clock cycle periods elapsed since last reset to zero.

This register can only be written when SECSENS = 0.

An invalid data (i.e. value greater or equal to 0xA8C00000) will not be written into the seconds register if in 24 hours mode.

15.12 WT Alarr Name: Access: Base Address:	m Register WT_ALR Read/Write 0x084						
31	30	29	28	27	26	25	24
			ALARMRE	EG[31:24]			
23	22	21	20 ALARMRI	19 -G[23:16]	18	17	16
				_0[20.10]			
15	14	13	12	11	10	9	8
			ALARMR	EG[15:8]			
7	6	5	4 ALARMF	3 REG[7:0]	2	1	0

### ALARMREG[31:0]: Alarm Register

An interrupt can be generated when the seconds register reaches this value.

This register can only be written when ALARMENS = 0.

An invalid data (i.e. value greater or equal to 0xA8C00000) will not be written into the alarm register if in 24 hours mode.

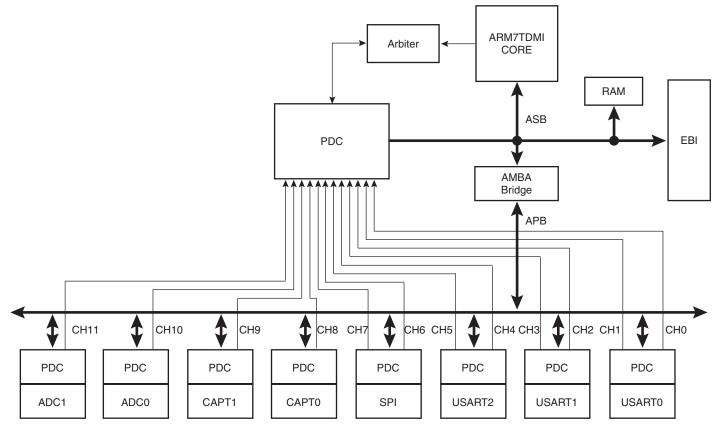
# 16. Peripheral Data Controller (PDC)

# 16.1 Overview

The Peripheral Data Controller (PDC) permits easy and quick transfers of large blocks of words from memory to peripheral or from peripheral to memory.

### 16.2 Block Diagram

Figure 16-1. PDC Block Diagram



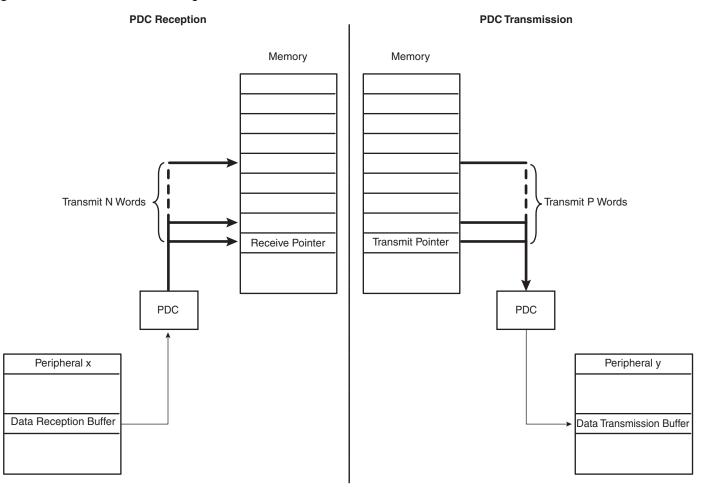
Each PDC channel, with a start address and a counter (number of words), can automatically put/get a block of transmitted words in/from a specific memory area. Each peripheral that transfers data can have a channel corresponding to a Peripheral Data Controller channel.

The peripherals associated to the PDC channels are listed in Table 16-1, "PDC Connection," on page 98.





### Figure 16-2. PDC Functional Diagram



The PDC has data transfer ports which connect to the ASB and the AMBA Bridge. It is programmed via the APB. The ASB bus request and grant signals are used to request bus access, and to detect when that access has been granted according to the standard AMBA bus arbitration scheme.

Data transfer to a peripheral is made directly via the APB (AMBA Bridge, to avoid tying up the ASB). A simple arbitration scheme is implemented between the ASB and PDC, to control APB access.

Before programming any PDC transfer, the PDC module must be enabled. This is done by setting the PDC bit to a logical 1 in the PMC\_ECR register (see "Power Management Controller (PMC)" on page 269.

The PDC channel is programmed using PDC\_CRx (Control Register x, x = 0 to 9), PDC\_MPRx (Memory Pointer x) and PDC\_TCRx (Transfer Counter x).

The status of the PDC transfer is given in the Status Register of the associated peripheral.

The pointer registers (PDC\_MPRx) are used to store the address of the buffer.

The counter registers (PDC\_TCRx) are used to store the size of these buffers (i.e. the number of data to be transferred).

When a transfer is performed, the counter is decremented and the pointer is incremented. When the counter reaches 0 (i.e. all the data have been sent/received to/from the module), the end status bit is set in the peripheral status register and can be programmed to generate an interrupt.

### 16.3 PDC Transfers

PDC transfers consist of byte (8-bit), half-word (16-bit) or word (32-bit) data transmitted from peripheral to memory or from memory to peripheral.

Transfers are triggered by the peripheral signals.

The PDC makes block transfers one byte, one half-word or one word at a time (programmed in the PDC\_CRx register, with direction). Each transfer is triggered by a PDC request from the peripheral. The PDC releases the AMBA bus after each transfer. A new trigger is needed for each transfer.

Between transfers, the ASB memory pointer is incremented and the transfer counter is decremented.

Each block of data can be programmed to be up to 64 Kbytes.

Transfers stop when the transfer counter reaches zero, and the trigger is disabled.

Figure 16-3.	Transfer Example (Byte)
--------------	-------------------------

Memory Pointer (MPR)	XXXXXXX X 00001000
Counter (TCR)	0 X 4 X 3 X 2 X 1 X 0
RDY	
END	
Memory Address	XXXXXXXX X X X X X X X X X X X X X X X

### 16.4 Memory Pointers

There is one 32-bit memory address pointer for each channel (PDC\_MPRx). Each memory pointer points to a location in the AT91SAM7A2 memory space (on chip RAM or external memory on the EBI).

The PDC\_MPRx is automatically incremented by 1, 2 or 4 after each transfer, for byte, half-word or word transfers. The PDC\_MPRx must be initialized before any transfers are started.

If PDC\_MPRx is reprogrammed while the PDC is operating, the address of the transfers will be changed. The PDC will continue to perform transfers when triggered, from the newly programmed address.

### 16.5 Transfer Counter

There is one 16-bit Transfer Counter (PDC\_TCRx) for each channel, which is used to count the size of the block of transfers. The PDC\_TCRx is decremented after every data transfer. When the PDC\_TCRx reaches zero the block transfer is complete, the PDC stops transferring data and disables the trigger.





It is not possible to trigger a block of transfers if the PDC\_TCRx value is zero. When the PDC\_TCRx is programmed to a non-zero value, transfers can be triggered by the peripheral for that particular channel.

The number of transfers required is programmed in the PDC\_TCRx which is memory mapped as a 16-bit Read/Write register. The number of transfers remaining can be read in the PDC\_TCRx register.

If the PDC\_TCRx is reprogrammed while the PDC is operating, the number of transfers will be changed. The PDC will continue to count transfers when triggered, from the newly programmed value.

The end of transfer is signaled to the peripheral via the PDC\_END signal. The PDC does not have any dedicated status registers.

While the PDC is operating, in order to stop PDC transfers correctly and to get a fixed value in the PDC\_MPRx register, user should write two consecutives '0' value in the PDC\_TCRx register. In case the second write is not done and if a PDC transfer has started before the first write in the PDC\_TCRx register then PDC\_MPRx register value will change during the next core clock periods.

### 16.6 PDC Configuration

For emulation purposes, each PDC channel can be software configured to be attached to a different peripheral.

In the AT91SAM7A2 microcontroller, each PDC channel is attached to a dedicated peripheral (with a fixed direction and fixed address). Software must configure each PDC channel so that accesses are correctly done by the PDC module.

Peripheral	PDC Channel	Transfer Direction	DIR Bit in PDC_CRx	Associated Peripheral Register	Associated Peripheral Address
USABT0	RX: Ch0	Reception	0	USART0_RHR	0xFFFA8080
USARIU	TX: Ch1	Transmission	1	USART0_THR	0xFFFA8084
	RX: Ch2	Reception	0	USART1_RHR	0xFFFAC080
USART1	TX: Ch3	Transmission	1	USART1_THR	0xFFFAC084
ADC0 (8-channel 10-bit)	Ch4	Reception	0	ADC0_DR	0xFFFB0080
ADC1 (8-channel 10-bit)	Ch5	Reception	0	ADC1_DR	0xFFFB0084
CDI	RX: Ch6	Reception	0	SPI_RDR	0xFFFB4080
SPI	TX: Ch7	Transmission	1	SPI_TDR	0xFFFB4084

Table 16-1. PDC Connection

The end of transmission or reception for each PDC channel transfer is indicated in the status register of the attached peripheral. The PDC\_TCRx is decremented with the peripheral trigger when a word has been transferred either from memory to peripheral or from peripheral to memory.

Peripheral	PDC Channel	Transfer Direction	Associated Peripheral Status Register	End of Transfer Bit in Status Register	Status Bit for PDC_TCRx Decrement (Trigger)
USART0	RX: Ch0	Reception	USART0_SR	ENDRX	RXRDY
USARTU	TX: Ch1	Transmission	USART0_SR	ENDTX	TXRDY
USART1	RX: Ch2	Reception	USART1_SR	ENDRX	RXRDY
USARTI	TX: Ch3	Transmission	USART1_SR	ENDTX	TXRDY
ADC0 (8-channel 10-bit)	Ch4	Reception	ADC0_SR	TEND	EOC
ADC1 (8-channel 10-bit)	Ch5	Reception	ADC1_SR	TEND	EOC
CDI	RX: Ch6	Reception	SPI_SR	REND	RDRF
SPI	TX: Ch7	Transmission	SPI_SR	TEND	TDRE
Capture CAPT0	Ch8	Reception	CAPT0_SR	PDCEND	DATACAPT
Capture CAPT1	Ch9	Reception	CAPT1_SR	PDCEND	DATACAPT

#### **Table 16-2.**PDC Transfer Status

### 16.6.1 Configuration Steps

- Enable PDC clock by writing the PDC bit in PMC\_PMSR of the PMC peripheral
- Configuration of PDC\_PRA: Address of targeted register TX or RX
- Configuration of PDC\_CR: Flux direction and element size (8-bit, 16-bit, 32-bit)
- Configuration of PDC\_MR: Address of a memory space to receive or transmit data
- Configuration of PDC\_TC: Number of transmissions or receptions to do and start PDC

### 16.7 PDC Transfer Example

#### 16.7.1 Transmission on SPI

Assuming the following:

- SPI bits per transfer = 10 on NPCS0 (i.e. BITS[3:0] = 0010b in SPI\_CRS0)
- Number of 10-bit words to transfer: 15
- Address of buffer in internal RAM for 10-bit words to be transmitted 0x00000100 (first 10-bit word is at address 0x00000100, second 10-bit word is at address 0x00000102, ...)
- SPI clock is enabled (SPI = 1 in SPI\_PMSR)
- SPI is enabled (SPIENS = 1 in SPI\_SR)

PDC channel 7 must be configured as follows:

- PDC\_PRA7 = 0xFFFB4084 (i.e. address of the SPI\_TDR register)
- PDC\_CR7 = 0x00000003 (i.e. 10-bit words cater in 16-bit words so PDC transfer size is a half-word incrementing the address pointer by 2 after each transfer, transfers are done from memory to peripheral so DIR = 1)
- PDC\_MPR7 = 0x00000100 (address of buffer)
- PDC\_TCR7 = 0x0000000F (number of 10-bit words to transfer)





As soon as the software writes the number of bytes to transfer in the PDC\_TCR7 register, the PDC starts transmitting the 15 10-bit words.

When all the 10-bit words have been transferred to the SPI\_TDR register, the TEND bit in the SPI\_SR register will be set to a logical 1 informing the software that the transfer is completed. The TEND bit in the SPI\_SR register can also generate an interrupt if the corresponding bit is set in the SPI\_IMR register.

**Note:** If a module is used in reception and transmission, the reception channel must be configured before the transmission channel.

### 16.7.2 Reception on SPI

Assuming the following:

- SPI bits per transfer = 8 on NPCS0 (i.e. BITS[3:0] = 0000b in SPI\_CRS0)
- Number of 8-bit words to transfer: 69
- Address of buffer in external RAM for 8-bit words to be transmitted 0x48000000 (first 8-bit word is at address 0x48000000, second 8-bit word is at address 0x48000001, ...)
- .SPI clock is enabled (SPI = 1 in SPI\_PMSR)
- SPI is enabled (SPIENS = 1 in SPI\_SR)

PDC channel 6 must be configured as follows:

- PDC\_PRA6 = 0xFFFB4084 (i.e. address of the SPI\_TDR register)
- PDC\_CR6 = 0x00000000 (i.e. 8-bit words cater in 8-bit words so PDC transfer size is a byte incrementing the address pointer by 1 after each transfer, transfers are done from peripheral to memory so DIR = 0)
- PDC\_MPR6 = 0x48000000 (address of buffer in external RAM)
- PDC\_TCR6 = 0x00000045 (number of 8-bit words to transfer)

As soon as the software writes the number of bytes to transfer in the PDC\_TCR6 register, the PDC starts receiving the 69 8-bit words.

When all the 8-bit words have been received (i.e. all bytes have been written in external RAM), the REND bit in the SPI\_SR register will be set to a logical 1 informing the software that the transfer is completed. The REND bit in the SPI\_SR register can also generate an interrupt if the corresponding bit is set in the SPI\_IMR register.

# 16.8 Peripheral Data Controller (PDC) Memory Map

Base Address: 0xFFFF8000

### Table 16-3.PDC Memory Map

Offset	Register	Name	Access	Reset State
0x000				
0x07C	Reserved	_	_	_
0x080	CH0 Peripheral Register Address	PDC_PRA0	Read/Write	0xFFE00000
0x084	CH0 Control Register	PDC_CR0	Read/Write	0x0000000
0x088	CH0 Memory Pointer	PDC_MPR0	Read/Write	0x0000000
0x08C	CH0 Transfer Counter	PDC_TCR0	Read/Write	0x0000000
0x090	CH1 Peripheral Register Address	PDC_PRA1	Read/Write	0xFFE00000
0x094	CH1 Control Register	PDC_CR1	Read/Write	0x0000000
0x098	CH1 Memory Pointer	PDC_MPR1	Read/Write	0x0000000
0x09C	CH1 Transfer Counter	PDC_TCR1	Read/Write	0x0000000
0x0A0	CH2 Peripheral Register Address	PDC_PRA2	Read/Write	0xFFE00000
0x0A4	CH2 Control Register	PDC_CR2	Read/Write	0x0000000
0x0A8	CH2 Memory Pointer	PDC_MPR2	Read/Write	0x0000000
0x0AC	CH2 Transfer Counter	PDC_TCR2	Read/Write	0x0000000
0x0B0	CH3 Peripheral Register Address	PDC_PRA3	Read/Write	0xFFE00000
0x0B4	CH3 Control Register	PDC_CR3	Read/Write	0x0000000
0x0B8	CH3 Memory Pointer	PDC_MPR3	Read/Write	0x0000000
0x0BC	CH3 Transfer Counter	PDC_TCR3	Read/Write	0x0000000
0x0C0	CH4 Peripheral Register Address	PDC_PRA4	Read/Write	0xFFE00000
0x0C4	CH4 Control Register	PDC_CR4	Read/Write	0x0000000
0x0C8	CH4 Memory Pointer	PDC_MPR4	Read/Write	0x0000000
0x0CC	CH4 Transfer Counter	PDC_TCR4	Read/Write	0x0000000
0x0D0	CH5 Peripheral Register Address	PDC_PRA5	Read/Write	0xFFE00000
0x0D4	CH5 Control Register	PDC_CR5	Read/Write	0x0000000
0x0D8	CH5 Memory Pointer	PDC_MPR5	Read/Write	0x0000000
0x0DC	CH5 Transfer Counter	PDC_TCR5	Read/Write	0x0000000
0x0E0	CH6 Peripheral Register Address	PDC_PRA6	Read/Write	0xFFE00000
0x0E4	CH6 Control Register	PDC_CR6	Read/Write	0x0000000
0x0E8	CH6 Memory Pointer	PDC_MPR6	Read/Write	0x0000000
0x0EC	CH6 Transfer Counter	PDC_TCR6	Read/Write	0x0000000
0x0F0	CH7 Peripheral Register Address	PDC_PRA7	Read/Write	0xFFE00000
0x0F4	CH7 Control Register	PDC_CR7	Read/Write	0x0000000
0x0F8	CH7 Memory Pointer	PDC_MPR7	Read/Write	0x0000000





### Table 16-3.PDC Memory Map

Offset	Register	Name	Access	Reset State
0x0FC	CH7 Transfer Counter	PDC_TCR7	Read/Write	0x0000000
0x100	CH8 Peripheral Register Address	PDC_PRA8	Read/Write	0xFFE00000
0x104	CH8 Control Register	PDC_CR8	Read/Write	0x0000000
0x108	CH8 Memory Pointer	PDC_MPR8	Read/Write	0x0000000
0x10C	CH8 Transfer Counter	PDC_TCR8	Read/Write	0x0000000
0x110	CH9 Peripheral Register Address	PDC_PRA9	Read/Write	0xFFE00000
0x114	CH9 Control Register	PDC_CR9	Read/Write	0x0000000
0x118	CH9 Memory Pointer	PDC_MPR9	Read/Write	0x0000000
0x11C	CH9 Transfer Counter	PDC_TCR9	Read/Write	0x0000000

# 16.9 PDC CH0...CH9 Peripheral Register Address

Name: Access: Base Address:	PDC_P Read/W 0xXX0	RA0PDCPRA9 /rite					
31	30	29	28	27	26	25	24
			CHPRA	A[31:24]			
23	22	21	20	19	18	17	16
			CHPR	A[23:16]			
15	14	13	12	11	10	9	8
			CHPR	A[15:8]			
7	6	5	4	3	2	1	0
			CHPF	RA[7:0]			

### • CHPRA[31:0] Peripheral Register Address

CHPRA[31:0] must be loaded with the address of the target register (peripheral receive or transmit register).

# 16.10 PDC CH0...CH9 Control Register

Name: Access: Base Address:	PDC_C Read/W	R0PDC_CR9					
31	30	29	28	27	26	25	24
-	-	—	-	-	-	-	-
23	22	21	20	19	18	17	16
_	_	_	-	_	_	_	-
15	14	13	12	11	10	9	8
-	_	_	-	-	-	-	_
7	6	5	4	3	2	1	0
-	—	_	-	-	SIZE	[1:0]	DIR

### • DIR: Transfer direction

0: Peripheral to memory.

1: Memory to peripheral.

# • SIZE[1:0]: Transfer size

Defines the size of the transfer.

SIZE	Transfer Size	
0	0	Byte (8-bit)
0	1	Half-word (16-bit)
1	0	Word (32-bit)
1	1	Reserved





# 16.11 PDC CH0...CH9 Memory Pointer Register

Name: Access: Base Address:	PDC_M Read/W 0xXX8	PR0PDC_MP /rite	R9				
31	30	29	28	27	26	25	24
			CHPTF	R[31:24]			
23	22	21	20	19	18	17	16
			CHPTF	R[23:16]			
15	14	13	12	11	10	9	8
			CHPT	R[15:8]			
7	6	5	4	3	2	1	0
			CHPT	R[7:0]			

# • CHPTR[31:0]: Channel Pointer

CHPTR[31:0] must be loaded with the address of the target buffer (memory address).

Name: Access: Base Address:	PDC_T Read/W 0xXXC		R9				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	_	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
			CHC1	FR15:8]			
7	6	5	4	3	2	1	0
			CHC	TR[7:0]			

# 16.12 PDC CH0...CH9 Transfer Register

### • CHCTR[15:0]: Channel Counter

CHCTR[15:0] must be loaded with the size of the receive buffer.

0: Stop Peripheral Data Transfer dedicated to the peripheral X.

1 to 65535: Start immediately Peripheral Data Transfer.





# **17. Generic Interrupt Controller (GIC)**

# 17.1 Overview

The GIC is an 8-level priority, individually maskable, vectored interrupt controller. It can substantially reduce the software and real time overhead in handling internal and external interrupts.

The interrupt controller is connected to the nFIQ (fast interrupt request) and the nIRQ (standard interrupt request) inputs of the ARM7TDMI processor (see Figure 17-1 below). The processor's nFIQ line can only be asserted by the external fast interrupt request input, FIQ. The nIRQ line can be asserted by all other internal and external interrupt sources.

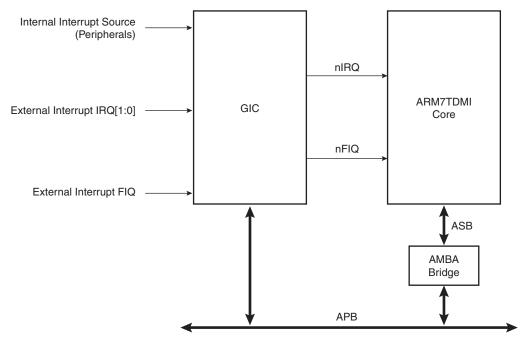


Figure 17-1. GIC Connection to Core

An 8-level priority encoder allows the user to define the priority between the different nIRQ interrupt sources. The internal interrupt sources are programmed to be level sensitive or edge triggered. The external interrupt sources can be programmed to be positive or negative edge triggered or high or low level sensitive.

Interrupt Source	Name	Description	GIC bit
0	FIQ	Fast interrupt	FIQ
1	SWIRQ0	Software interrupt 0	SWIRQ0
2	INT_0	Watch Dog	WD
3	INT_1	Watch Timer	WT
4	INT_2	USART0	USART0
5	INT_3	USART1	USART1
6	INT_4	CAN3	CAN3
7	INT_5	SPI	SPI
8	INT_6	CAN1	CAN1
9	INT_7	CAN2	CAN2
10	INT_8	ADC0	ADC0
11	INT_9	ADC1	ADC1
12	INT_10	General Purpose Timer 0 channel 0	GPT0CH0
13	INT_11	General Purpose Timer 0 channel 1	GPT0CH1
14	INT_12	General Purpose Timer 0 channel 2	GPT0CH2
15	SWIRQ0	Software interrupt 1	SWIRQ0
16	SWIRQ1	Software interrupt 2	SWIRQ1
17	SWIRQ2	Software interrupt 3	SWIRQ2
18	INT_13	General Purpose Timer 1 channel 0	GPT1CH0
19	INT_14	Pulse Width Modulation	PWM
20	INT_18	CANO	CAN0
21	INT_19	UPIO	UPIO
22	INT_20	Capture 0	CAPT0
23	INT_21	Capture 1	CAPT1
24	INT_22	Simple Timer 0	ST0
25	INT_23	Simple Timer 1	ST1
26	SWIRQ4	Software interrupt 4	SWIRQ4
27	SWIRQ5	Software interrupt 5	SWIRQ5
28	EXT_0	External interrupt IRQ0	IRQ0
29	EXT_1	External interrupt IRQ1	IRQ1
30	SWIRQ6	Software interrupt 6	SWIRQ6
31	SWIRQ7	Software interrupt 7	SWIRQ7

 Table 17-1.
 Interrupt Sources





# 17.2 Interrupt Handling

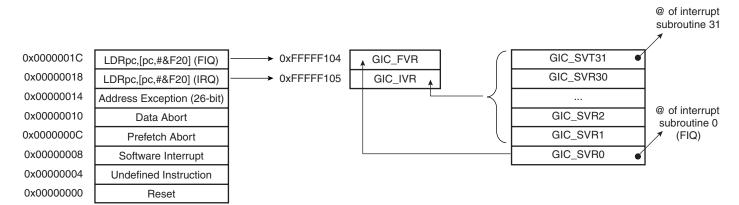
### 17.2.1 Hardware Interrupt Vectoring

Hardware interrupt vectoring reduces the number of instructions to reach the interrupt handler to only one. By storing the following instruction at address 0x00000018, the processor loads the program counter with the interrupt handler address stored in the GIC\_IVR register. Execution is then vectored to the interrupt handler corresponding to the current interrupt (see Figure 17-2 below).

ldr PC,[PC,# -&F20]

The current interrupt is the interrupt with the highest priority when the Interrupt Vector Register (GIC\_IVR) is read. The value read in the GIC\_IVR corresponds to the address stored in the Source Vector Register (GIC\_SVR) of the current interrupt. Each interrupt source has its corresponding GIC\_SVR. In order to take advantage of the hardware interrupt vectoring, it is necessary to store the address of each interrupt handler in the corresponding GIC\_SVR at system initialization.





### 17.2.2 Priority Controller

The nIRQ line is controlled by an 8-level priority encoder. Each source has a programmable priority level of 7 to 0. Level 7 is the highest priority and level 0 the lowest.

When the GIC receives more than one unmasked interrupt at a time, the interrupt with the highest priority is serviced first. If both interrupts have equal priority, the interrupt with the lowest interrupt source number is serviced first (see Table 17-1, "Interrupt Sources," on page 107).

The current priority level is defined as the priority level of the current interrupt at the time the GIC\_IVR register is read (the interrupt that will be serviced).

If a higher priority unmasked interrupt occurs while an interrupt already exists, there are two possible outcomes depending on whether the GIC\_IVR has been read:

• If the nIRQ line has been asserted but the GIC\_IVR has not been read, then the processor will read the new higher priority interrupt handler address in the GIC\_IVR register and the current interrupt level is updated.

If the processor has already read the GIC\_IVR, then the nIRQ line is reasserted. When the
processor has authorized nested interrupts to occur and reads the GIC\_IVR again, it reads
the new higher priority interrupt handler address. At the same time the current priority value
is pushed onto a first-in last-out stack and the current priority is updated to the higher priority.

When the end of Interrupt Command Register (GIC\_EOICR) is written, the current interrupt level is updated with the last stored interrupt level from the stack (if any). Hence at the end of a higher priority interrupt, the GIC returns to the previous state corresponding to the preceding lower priority interrupt that had been interrupted.

#### 17.2.3 Software Interrupt Handling

The interrupt handler must read the GIC\_IVR as soon as possible. This de-asserts the nIRQ request to the processor and clears the interrupt in case it is programmed to be edge triggered. This permits the GIC to assert the nIRQ line again when a higher priority unmasked interrupt occurs.

At the end of the interrupt service routine, the end of Interrupt Command Register (GIC\_EOICR) must be written. This allows pending interrupts to be serviced.

#### 17.2.4 Interrupt Masking

Each interrupt source, including FIQ, can be enabled or disabled using the command registers GIC\_IECR and GIC\_IDCR. The interrupt mask can be read in the read only register GIC\_IMR. A disabled interrupt does not affect the servicing of other interrupts.

#### 17.2.5 Interrupt Clearing and Setting

All interrupt sources that are programmed to be edge triggered (including FIQ) can be individually set or cleared by respectively writing to the GIC\_ISCR and GIC\_ICCR registers. This function of the interrupt controller is available for auto-test or software debug purposes.

#### 17.2.6 Configuration Steps

- Initialization of the interrupt at the peripheral level
- Configuration of GIC\_SMR: Priority and detection mode
- Configuration of GIC\_SVR: Address of the function associated to this interrupt
- Configuration of GIC\_IER: Enable the corresponding peripheral

#### 17.2.7 Fast Interrupt Request

The external FIQ line is the only source which can raise a fast interrupt request to the processor. Therefore it has no priority controller. It can be programmed to be positive or negative edge triggered or high or low level sensitive in the GIC\_SMR0 register.

The fast interrupt handler address can be stored in the GIC\_SVR0 register. The value written into this register is available by reading the GIC\_FVR register when an FIQ interrupt is raised. By storing the following instruction at address 0x0000001C, the processor will load the program counter with the interrupt handler address stored in the GIC\_FVR register.

ldr PC, [PC, #-&F20]

Alternatively the interrupt handler can be stored starting from address 0x0000001C as described in the ARM7TDMI datasheet.





#### 17.2.8 Software Interrupt

Any interrupt source of the GIC can be a software interrupt. It must be programmed to be edge triggered in order to set or clear it by writing to the GIC\_ISCR and GIC\_ICCR. This is totally independent of the SWI instruction of the ARM7TDMI processor.

#### 17.2.9 Spurious Interrupt

A spurious interrupt is a signal of very short duration on one of the interrupt input lines.

### 17.3 Standard Interrupt Sequence

For details on the registers mentioned in the steps below, refer to the ARM7TDMI Embedded Core Datasheet.

It is assumed that:

- The Generic Interrupt Controller has been programmed, GIC\_SVR are loaded with corresponding interrupt service routine addresses and interrupts are enabled.
- The Instruction at address 0x18 (IRQ exception vector address) is:

```
ldr pc, [pc, #-&F20].
```

When nIRQ is asserted, if the bit I of CPSR is 0, the sequence is:

- 1. The CPSR is stored in SPSR\_irq, the current value of the Program Counter is loaded in the IRQ link register (r14\_irq) and the Program Counter (r15) is loaded with 0x18. In the following cycle during fetch at address 0x1C, the ARM core adjusts r14\_irq, incrementing it by 4.
- 2. The ARM core enters IRQ mode, if it is not already.
- 3. When the instruction loaded at address 0x18 is executed, the Program Counter is loaded with the value read in GIC\_IVR. Reading the GIC\_IVR has the following effects:
  - Sets the current interrupt to be the one pending with the highest priority. The current level is the priority level of the current interrupt.
  - De-asserts the nIRQ line on the processor (even if vectoring is not used, GIC\_IVR must be read in order to de-assert nIRQ).
  - Automatically clears the interrupt, if it has been programmed to be edge triggered; pushes the current level on to the stack, returns the value written in the GIC\_SVR corresponding to the current interrupt.
- 4. The previous step branches to the corresponding interrupt service routine. This should start by saving the Link Register (r14\_irq) and the SPSR (SPSR\_irq). Note that the Link Register must be decremented by 4 when it is saved, if it is to be restored directly into the Program Counter at the end of the interrupt. The instruction: sub pc, lr, #4 may be used, for example.
- 5. Further interrupts can then be unmasked by clearing the I bit in CPSR, allowing reassertion of the nIRQ to be taken into account by the core. This can arise if an interrupt with a higher priority than the current one occurs.
- 6. The Interrupt Handler can then proceed as required, saving the registers which will be used and restoring them at the end. During this phase, an interrupt of higher priority than the current level will restart the sequence from step 1. Note that if the interrupt is programmed to be level sensitive, the source of the interrupt must be cleared during this phase.
- 7. The I bit in CPSR must be set in order to mask interrupts before exiting, to ensure that the interrupt is completed in an orderly manner.

- 8. The End Of Interrupt Command Register (GIC\_EOICR) must be written in order to indicate to the GIC that the current interrupt is finished. This causes the current level to be popped from the stack, restoring the previous current level if one exists on the stack. If another interrupt is pending, with lower or equal priority than the old current level but with higher priority than the new current level, the nIRQ line is reasserted, but the interrupt sequence does not immediately start because the I bit is set in the core.
- 9. The SPSR (SPSR\_irq) is restored. Finally, the saved value of the Link Register is restored directly into the PC. This has the effect of returning from the interrupt to whatever was being executed before, and of loading CPSR with the stored SPSR, masking or unmasking the interrupts depending on the state saved in the SPSR (the previous state of the ARM core).

**Note:** The I bit in the SPSR is significant. If it is set, it indicates that the ARM core was just about to mask IRQ interrupts when the mask instruction was interrupted. Hence, when the SPSR is restored, the mask instruction is completed (IRQ is masked).

### 17.4 Fast Interrupt Sequence

For details on the registers mentioned in the steps below, refer to the ARM7TDMI Embedded Core Datasheet.

It is assumed that:

- The Generic Interrupt Controller has been programmed, GIC\_SVR[0] is loaded with a fast interrupt service routine address and the fast interrupt is enabled.
- The Instruction at address 0x1C (FIQ exception vector address) is:

ldr pc, [pc, #-&F20]

Nested Fast Interrupts are not needed by the user.

When nFIQ is asserted, if the F bit of CPSR is 0, the sequence is:

- 1. CPSR is stored in SPSR\_fiq, the current value of the Program Counter is loaded in the FIQ link register (r14\_fiq) and the Program Counter (r15) is loaded with 0x1C. In the following cycle, during fetch at address 0x20, the ARM core adjusts r14\_fiq, incrementing it by 4.
- 2. The ARM core enters FIQ mode.
- 3. When the instruction loaded at address 0x1C is executed, the Program Counter is loaded with the value read in GIC\_FVR. Reading the GIC\_FVR automatically clears the fast interrupt (source 0 connected to the FIQ line), if it has been programmed to be edge triggered. In this case only, it de-asserts the nFIQ line on the processor.
- The previous step branches to the corresponding interrupt service routine. It is not necessary to save the Link Register (r14\_fiq) and the SPSR (SPSR\_fiq) if nested fast interrupts are not needed.
- 5. The Interrupt Handler can then proceed as required. It is not necessary to save registers r8 to r13 because FIQ mode has its own dedicated registers and the user registers r8 to r13 are banked. The other registers, r0 to r7, must be saved before being used, and restored at the end (before the next step). Note that if the fast interrupt is programmed to be level sensitive, the source of the interrupt must be cleared during this phase in order to de-assert the nFIQ line.
- 6. Finally, the Link Register (r14\_fiq) is restored into the PC after decrementing it by 4 (with instruction sub pc, lr, #4 for example). This has the effect of returning from the interrupt to whatever was being executed before, and of loading CPSR with the SPSR, masking or unmasking the fast interrupt depending on the state saved in the SPSR.





**Note:** The F bit in the SPSR is significant. If it is set, it indicates that the ARM core was just about to mask FIQ interrupts when the mask instruction was interrupted. Hence when the SPSR is restored, the interrupted instruction is completed (FIQ is masked).

### 17.5 Spurious Interrupt Sequence

A spurious interrupt is a signal of very short duration on one of the interrupt input lines. It is handled by the following sequence of actions.

- 1. When an interrupt is active, the GIC asserts the IRQ (or nFIQ) line and the ARM7TDMI enters IRQ (or FIQ) mode. At this moment, if the interrupt source disappears, the nIRQ (or nFIQ) line is de-asserted but the ARM7TDMI continues with the interrupt handler.
- 2. If the IRQ Vector Register (GIC\_IVR) is read when the nIRQ is not asserted, the GIC\_IVR is read with the contents of the Spurious Interrupt Vector Register.
- 3. If the FIQ Vector Register (GIC\_FVR) is read when the nFIQ is not asserted, the GIC\_FVR is read with the contents of the Spurious Interrupt Vector Register.
- 4. The Spurious Interrupt Routine must at least write into the GIC\_EOICR to perform an end of interrupt command. Until the GIC\_EOICR write is received by the interrupt controller, the nIRQ (or nFIQ) line is not reasserted.
- 5. This causes the ARM7TDMI to jump into the Spurious Interrupt Routine.
- 6. During a Spurious Interrupt Routine, the Interrupt Status Register GIC\_ISR reads 0.

# 17.6 Generic Interrupt Controller (GIC) Memory Map

Base Address: 0xFFFFF000

# Table 17-2.GIC Memory Map

Offset	Register	Name	Access	Reset State
0x000 _	GIC Source Mode Register 0	GIC_SMR0	Read/Write	0x00000000
0x07C	GIC Source Mode Register 31	GIC_SMR31		
0x080 -	GIC Source Vector Register 0 -	GIC_SVR0 -	Read/Write	0x00000000
0x0FC	GIC Source Vector Register 31	GIC_SVR31		
0x100	GIC IRQ Vector	GIC_IVR	Read-only	0x0000000
0x104	GIC FIQ Vector	GIC_FVR	Read-only	0x0000000
0x108	GIC Interrupt Status	GIC_ISR	Read-only	0x00000000
0x10C	GIC interrupt Pending	GIC_IPR	Read-only	0xXXXXXXXX
0x110	GIC Interrupt Mask	GIC_IMR	Read-only	0x00000000
0x114	GIC Core Interrupt Status	GIC_CISR	Read-only	0x00000000
0x118  0x11C	Reserved	_	_	-
0x120	GIC Interrupt Enable Command	GIC_IECR	Write-only	_
0x124	GIC Interrupt Disable Command	GIC_IDCR	Write-only	_
0x128	GIC Interrupt Clear Command	GIC_ICCR	Write-only	_
0x12C	GIC Interrupt Set Command	GIC_ISCR	Write-only	_
0x130	GIC End of Interrupt Command	GIC_EOICR	Write-only	_
0x134	GIC Spurious Vector	GIC_SPU	Read/Write	0x0000000





# 17.7 GIC Source Mode Register

Name: Access: Base Address:	Read/V	GIC_SMR0GIC_SMR31 Read/Write 0x0000x07C							
31	30	29	28	27	26	25	24		
-	_	-	-	-	-	-	-		
23	22	21	20	19	18	17	16		
-	_	-	-	_	-	-	-		
15	14	13	12	11	10	9	8		
-	-	-	-	_	—	_	_		
7	6	5	4	3	2	1	0		
_	SRCT	YP[1:0]	_	_		PRIOR[2:0]			

### • PRIOR[2:0]: Priority Level

These bits program the priority level (from 0-lowest to 7-highest) of all the interrupt sources. The priority level is not used for the FIQ in the SMR0.

### • SRCTYP[1:0]: Interrupt Source Type

SRCTY	'P[1:0] <sup>(1)</sup>	Internal Sources	External Sources		
0	0	High level sensitive	Low level sensitive		
0	1	Positive edge triggered	Negative edge triggered		
1	0	High level sensitive	High level sensitive		
1	1	Positive edge triggered	Positive edge triggered		

Note: 1. All the interrupts used by internal peripherals are considered as internal interrupts and subsequently the SRCTYP1 bit is always read at 0.

# 17.8 GIC Source Vector Register

Name: Access: Base Address:	GIC_SVR0GIC_SVR31 Read/Write ss: 0x0800x0FC									
31	30	29	28	27	26	25	24			
			VECT	[31:24]						
23	22	21	20	19	18	17	16			
			VECT	[23:16]						
15	14	13	12	11	10	9	8			
			VECT	[15:8]						
7	6	5	4	3	2	1	0			
			VEC	T[7:0]						

### • VECT[31:0]: Interrupt Handler Address

Address of the corresponding handler for each interrupt source.

17.9 GIC Inte Name: Access: Base Address:	rrupt Vector GIC_IVR Read-only 0x100	Registe	r				
31	30	29	28	27	26	25	24
			IRQV[	31:24]			
23	22	21	20	19	18	17	16
			IRQV[	23:16]			
15	14	13	12	11	10	9	8
			IRQV	[15:8]			
7	6	5	4	3	2	1	0
			IRQV	/[7:0]			

### • IRQV[31:0]: Interrupt Vector Address

Address of the currently serviced interrupt vector (user programmed in the GIC\_SVR register).

Note: GIC\_IVR = 0x0000000 when there is no current interrupt.

Note: When debugging, to read the GIC\_IVR register clears the IRQ interrupt if present at the GIC. To avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).





# 17.10 GIC FIQ Vector Register

Name: Access: Base Address:	GIC_FV Read-or 0x104	R					
31	30	29	28	27	26	25	24
			FIQV[	31:24]			
23	22	21	20	19	18	17	16
			FIQV[	23:16]			
15	14	13	12	11	10	9	8
			FIQV	[15:8]			
7	6	5	4	3	2	1	0
			FIQ\	/[7:0]			

### • FIQV[31:0]: FIQ Vector Address

Address of the FIQ serviced interrupt (user programmed in the GIC\_SVR0 register).

Note: When debugging, to read the GIC\_FVR register clears the FRQ interrupt if present at the GIC. To avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

# 17.11 GIC Interrupt Status Register

Name: Access: Base Address:	GIC_ISI Read-or 0x108						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	-	-	—	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	—	-	-	-
7	6	5	4	3	2	1	0
_					IRQID[4:0]		

### • IRQID[4:0]: Current IRQ Identifier

Current interrupt source number.

# 17.12 GIC Interrupt Pending Register

Name: Access: Base Address:	GIC_IPF Read-or 0x10C						
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

### • Interrupt Pending

0: Corresponding interrupt is inactive.

1: Corresponding interrupt is pending.

# 17.13 GIC Interrupt Mask Register

Name: Access: Base Address:	GIC_IM Read-or 0x110						
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

# • Interrupt Mask

0: Corresponding interrupt is disabled.

1: Corresponding interrupt is enabled.





# 17.14 GIC Core Interrupt Status Register

Name: Access: Base Address:	GIC_CI Read-or 0x114	SR	-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	_	—	-	-	-	-
15	14	13	12	11	10	9	8
-	—	—	—	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	-	_	-	NIRQ	NFIQ

### • NIRQ: NIRQ Status

0: nIRQ line is inactive.

1: nIRQ line is active.

### • NFIQ: NFIQ Status

0: nFIQ line is inactive.

1: nFIQ line is active.

# 17.15 GIC Interrupt Enable Command Register

	ten upt Ena		nu negister				
Name: Access:	GIC_IE0 Write-or						
		пу					
Base Address:	0x120						
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

# 17.16 GIC Interrupt Disable Command Register

Name: Access: Base Address:	GIC_ID Write-or 0x124						
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

# 17.17 GIC Interrupt Clear Command Register

Name: Access:	GIC_IC Write-or						
		าโง					
		·· y					
Base Address:	0x128						
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

# • Software Interrupt Clear

0: No effect.

1: Clears corresponding software interrupt.





# 17.18 GIC Interrupt Set Command Register

Name: Access: Base Address:	GIC_IC Write-or 0x12C		-				
31	30	29	28	27	26	25	24
SWIRQ7	SWIRQ6	IRQ1	IRQ0	SWIRQ5	SWIRQ4	ST1	ST0
23	22	21	20	19	18	17	16
CAPT1	CAPT0	UPIO	CAN0	PWM	GPT1CH0	SWIRQ3	SWIRQ2
15	14	13	12	11	10	9	8
SWIRQ1	GPT0CH2	GPT0CH1	GPT0CH0	ADC1	ADC0	CAN2	CAN1
7	6	5	4	3	2	1	0
SPI	CAN3	USART1	USART0	WT	WD	SWIRQ0	FIQ

### • Software Interrupt Set

0: No effect.

1: Clears corresponding software interrupt.

# 17.19 GIC End of Interrupt Command Register

Name: Access: Base Address:	GIC_EC Write-or 0x130						
31	30	29	28	27	26	25	24
-	_	_	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	-	-	-	-	_	-

A write access to this register (with any value) indicates that the interrupt treatment is completed.

17.20 GIC Spi	irious Veo	ctor Register	•					
Name: Access: Base Address:	GIC_SF Read/W 0x080	/rite						
31	30	29	28	27	26	25	24	
	SPUVECT[31:24]							
23	22	21	20	19	18	17	16	
			SPUVEC	CT[23:16]				
15	14	13	12	11	10	9	8	
	SPUVECT[15:8]							
7	6	5	4	3	2	1	0	
			SPUVE	CT[7:0]				

# 7.20 GIC Spurious Ventor Pagistor

• SPUVECT[31:0]: Spurious Interrupt Vector Handler Address

Address of the spurious interrupt handler.





# 18. 10-bit Analog to Digital Converter (ADC)

## 18.1 Overview

The AT91SAM7A2 includes two 10-bit ADCs (8 channels).

The 10-bit Analog to Digital Converter (ADC) can be configured in different modes as follows.

- Single input/one shot mode: One input selected and a single conversion.
- Single input/continuous mode: One input selected, the microprocessor gives the first start to the peripheral which is then completely independent. The PDC can be used to save the resulting data in memory.

The conversion is stopped in two ways:

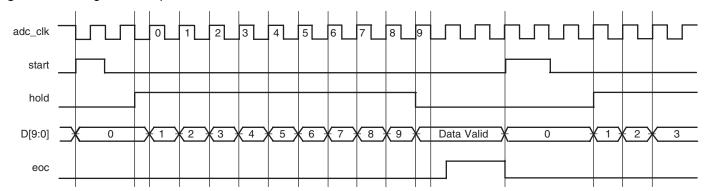
- a. By the microprocessor, by setting the STOP bit of the active control register.
- b. By the PDC: TEND can stop the conversion if the STOPEN bit of the mode register is active. This allows the user to order a conversion of a certain number of data without any software intervention.
- Multiple input/one shot mode: The user selects which analog inputs will be converted and specifies the names of the inputs that will be considered by the ADC and in which order they will be converted. This allows the user to make conversions of some of the eight inputs in the order of preference, in one shot. The PDC can be used to save each result.
- Multiple input/continuous mode: Several analog inputs are converted, the microprocessor first starts up the peripheral which then becomes completely independent. The conversion is stopped in two ways:
  - By the microprocessor, by setting the STOP bit of the active control register.
  - By the PDC: TEND can stop the conversion if the STOPEN bit of the mode register is active. This allows the user to order a conversion of a certain number of data without any software intervention.

If the PDC is active, a flag is set when the transfer of all the data is finished.

The converter is composed of a 10-bit cascaded potentiometric digital to analog converter connected to the negative input of a Sample and Hold comparator. It is based on a string of 64 polysilicon resistors connected between reference inputs VREF. So, the analog input to be converted needs to be in the interval [GND:VREFP].

In the terminology, the Integral Non-Linearity (INL) is a measure of the maximum deviation from a straight line passing through the end-points of the transfer function. It does not include the full scale error and the zero error. It is calculated from the real value of the LSB which is calculated from the output range (output variation between the minimal value 0 and the maximal one).

The Differential Non-linearity (DNL) is the difference between the measured change and the ideal change between any two adjacent codes. A specified DNL of  $\pm 1$  LSB max over the operating point temperature range ensures monotony. The DNL is relative to the real value of the LSB.



#### Figure 18-1. Signal Description

When start is high, the cell is reset and the internal clock is inhibited. The D[9:0] signal is at 512, so the internal ADC output is at (VREFP - GND)/2. The hold output is low, so the input voltage at the Sample and Hold stage of the comparator is sampled.

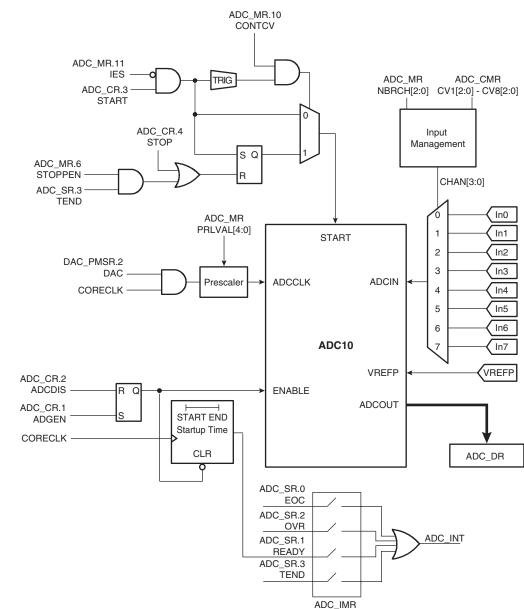
When the start goes low, the hold signal goes high with the next falling clock edge, and the input voltage is stored on the Sample and Hold capacitor. The comparator performs a comparison between the stored input voltage and the ADC output.

With the next rising clock signal, the comparator output becomes valid, this value being stored as D[9] in the internal register. The internal shift register now sets D[8] high, and a new comparison is performed. The next rising edge of clock stores the result in D[8]. After another 8 lowpulse of clock, all 10 bits are valid at output D[9:0]. The End of Conversion signal (EOC) is set high. The input is sampled again as the hold signal goes low and a new conversion can be started with a high-pulse of the start signal.





### 18.2 Block Diagram



#### Figure 18-2. ADC Block Diagram

#### 18.2.1 Conversion Details

The conversion of a single analog value to 10-bit digital data requires 11 ADC clock cycles comprised of the following:

- One ADC clock cycle to sample the analog input signal.
- Ten ADC clock cycles to fix the ten resultant bits.

The clock input to the ADC has to be lower than 700 kHz. The user can choose a frequency by writing in the ADC mode register (ADC\_MR) the preload value of the counter (PRLVAL[4:0]). The master clock is divided by this value, and the result is the ADC clock. The preload value is coded on 7 bits with the 2 LSBs always low to guarantee a duty cycle of ½. The user can divide

master clock by 0 (so ADC clock = CORECLK), 4, 8, ..., 48, ..., 64, ..., 124. This allows adapting the ADC clock as well as possible with a master clock comprised between 700 kHz and 30 MHz. For example, if CORECLK = 30 MHz with a preload value of 48 the ADC clock is 625 kHz.

A single conversion at the maximum clock rate permitted (i.e. 700 kHz) will occur in 15.7 ms.

The ADC starts the conversion by writing 1 in the START bit of the control register (ADC\_CR).

Writing 1 to the START bit starts the conversion even if the analog structure begins the conversion when start goes low, the interface transmits the opposite of the start command to the analog part.

For a conversion, different input combinations can be selected. The 3-bit NBRCH[2:0] indicates how many inputs will be converted (the real number is the value of NBRCH incremented by one).

The result of the conversion is stored in the convert data register (ADC\_DR). When the conversion is complete, the analog part activates the EOC bit in the ADC Status register (ADC\_SR) and sends an EOC signal to the PDC which can take the result and write at a memory location. The EOC bit in ADC\_SR (Status Register) is cleared when the ADC\_DR (Convert Data Register) is read. If a new result arrives before the PDC or the CPU read the old data, the Overrun bit (OVR) is set active to specify to the microprocessor that data is lost. If the PDC is used to save the results and if the transfer of all the data is finished, the PDC sets the TEND bit to a logical 1.

The READY bit is set after an absolute time of 4  $\mu$ s after an enable command, which corresponds to the initialization time of the analog part. This time is necessary to stabilize the analog structure and does not depend on the choice of the ADC clock or the names of analog inputs considered. The number of master clock periods necessary to wait during 4  $\mu$ s is remembered in the STARTUPTIME bits of the mode register.

The user can make conversions in continuous mode. This status is indicated by the CONTCV bit of the ADC Mode Register (ADC\_MR). In this case, the microprocessor gives the first start to the ADC and the peripheral does not stop the conversion until the STOP bit of the ADC Control Register is set, or when the TEND bit of the status register is set if STOPEN is active. However, the user should be vigilant, because after a stop command in continuous mode, the ADC finishes the ongoing conversion and this may appear to be an extra conversion. The digital interface between the analog part and the APB bus is in stand alone mode; this permits conversion without any help. This mode can be associated with multiple inputs as well as a single input. The different steps of the conversion are equivalent to those of a single conversion.

If the ADC is configured in continuous mode, a particular sequence should be observed at the end of a PDC transfer. When a set of PDC transfers have reached the end, the ADC runs an extra conversion. The CPU must clear the TEND flag before the end (EOC in ADC\_SR) of the extra conversion. If the software can not ensure clearing the TEND flag before the EOC of the extra conversion, two solutions are available:

- When a set of PDC transfers is completed, before starting an additional set of ADC conversions associated with PDC transfers, software must reset the ADC (SWRST in ADC\_CR).
- When a set of PDC transfers is completed, before starting an additional set of ADC conversions associated with PDC transfers, software should start a conversion, wait for the EOC flag in the ADC\_SR register and read ADC\_DR to clear the EOC flag.





#### 18.2.2 Modes of Operation

The ADC can be active or shutdown; in the latter case it is in a power saving mode.

At any time the software can program the ADC to be disabled to save power. Setting the ADC-DIS bit of the ADC Control register will put the ADC Analog circuitry into standby mode. To reduce the power consumption near to 0, the user can switch off the ADC peripheral clock in the Disable Clock Register (ADC\_DCR), thus also disabling the Digital part of the ADC.

When the ADC is re-enabled, a minimum of 4  $\mu$ s is required before the analog circuitry is stabilized and ready for reliable usage. The user has to initialize the STARTUPTIME in the ADC\_MR register value by indicating how many clock periods of the master clock are necessary to make 4  $\mu$ s.

When the ADC is enabled for the first time (after standby mode but not after wait mode) the interface starts counting and then sets the READY bit in ADC\_SR (Status Register) which allows a conversion. The ready flag also indicates if the ADC is converting data, or is waiting for a start, because this flag is low when ADC is disabled or converting and is high when ADC is waiting for a start.

#### Wait Mode

When the analog part of the ADC peripheral is in standby mode, but the digital part of the peripheral is active, the circuitry is in wait mode. To leave this status, the user can do one of the following.

- Disable the CPU clock. The peripheral is then in standby mode.
- Enable the analog circuitry by setting the ADCEN of the control register. The peripheral is active.

#### Standby Mode

When the analog part as well as the digital part are in standby mode, the ADC peripheral is really in standby mode. The digital part is in standby mode when the clock is disabled. The microcontroller disables the peripheral clock by writing in the Disable Clock register (ADC\_DCR) The microcontroller disables the analog part by setting the ADCDIS bit of the control register.

#### Warnings

As the standby mode can be effective only after having been in wait mode, a specific order in the different actions has to be respected. If the user disables the clock of the interface before disabling the analog part of the peripheral, the peripheral is not in standby mode.

#### START Mode

The peripheral is commanded by an internal start (given by the microprocessor). In this mode, the analog part of the ADC peripheral is waiting for a start.

#### NBRCH[2:0]

These bits indicate how many inputs will be converted within a one shot or continuous mode.

#### CONT Mode

The CONTCV bit of the ADC mode register indicates if the peripheral is converting in a continuous mode (in this case the bit is high) or in a one shot mode. This bit is initialized to 0.

#### 18.2.3 Conversion Sequence

The basic sequence of operation after a reset is detailed below.

- 1. Program the ADC Mode register: the interface has to know the STARTUPTIME and the PRLVAL before receiving a start command. Determine the names of analog inputs. Indicate if conversions are made in continuous mode (CONTCV=1).
- 2. Program the ADC Control register (ADC\_CR) to enable the ADC.
- 3. Wait for READY bit in ADC\_SR. When the flag is set, the ADC is ready to start conversion. An interrupt can be generated to detect when the ADC is ready to start a conversion if the corresponding bit is enabled in the ADC\_IMR (Interrupt Mask Register). This flag can be high only after an enable command has been performed because it is this operation which starts the startup time of 4 µs. Every command (even start) before the ready flag is ignored.
- 4. Initiate conversion start by writing a start command in the ADC Control register.

If a single conversion is being done, the following applies:

- 5. The analog input voltage is sampled during 1 ADC clock cycle. The digital result is transferred after 10 ADC clock cycles.
- Conversion completes after 11 ADC clock cycles from the start command. The digital 10-bit data is latched into ADC\_DR (Convert Data Register), the EOC bit in ADC\_SR (Status Register) is set.
- 7. The PDC or the CPU can then read the digital value in ADC\_DR, which automatically clears the EOC bit in ADC\_SR. Another result can be saved before the data has been read. In this case, the OVR bit is set. When the PDC has stored all the data required, the TEND bit is set.

When multiple conversions or continuous conversions are programmed, steps 5, 6 and 7, as stated above, are repeated.

#### 18.3 Power Management

The ADC is provided with a power management block allowing optimization of power consumption. See "Power Management Block" on page 25.

#### 18.4 Example of Use

Acquisition in a single shot on eight different inputs using the PDC, with an interrupt associated to the end of conversion, is performed according to the following steps.

#### 18.4.1 Configuration Steps

- Enable the clock on the ADC peripheral by writing the ADC bit in ADC\_ECR.
- Do a software reset of the ADC to be in a known state by writing the SWRST bit in ADC\_CR.
- Configuration of ADC\_ MR: Select the ADC clock to be less than 700kHz (PRLVAL = 9 for 30MHz), STOPEN bit remains at 0 as the device is not in continuous conversion mode. Analog circuitry stabilization time is programmed using the STARTUPTIME field (0x78 for 30 MHz) and must be more than 4  $\mu$ s. The number of acquisitions is programmed in the NBRCH field; write 7 for eight conversions. The CONTCV bit selects the conversion mode and is left at 0.
- Enable the ADC by writing the ADCEN bit in ADC\_CR.
- Configuration of ADC\_ CMR: This register indicates which input will be used for the eight conversions.





• Choose the input order to be {5,7,6,2,0,1,4,3}, therefore ADC\_MR has the following programmed value:

```
((5<<0) | (7<<4) | (6<<8) | (2<<12) | (0<<16) | (1<<20) | (4<<24) | (3<<28) )
```

- Configuration of ADC \_IER: Generate an interrupt at the end of the eight conversions. This is done by writing TEND. When the PDC completes the eight conversions, an interrupt will be generated. GIC must be configured.
- Configure PDC for ADC module and set it to eight conversions of 16 bits.
- If the status READY bit is set in ADC\_SR, start conversions by writing the START bit in ADC\_CR, otherwise users should wait the READY flag, in consequence of bit ADCEN in ADC\_CR. An interrupt can be associated to this event.

### 18.4.2 Interrupt Handling

- IRQ Entry and call C function.
- Read ADC\_SR and verify the source of the interrupt.
- Clear the corresponding interrupt at peripheral level by writing in the ADC\_CSR.
- Interrupt treatment: read data conversion in the memory space programmed in the PDC. Result are between 0 to 1023 (0 to 3.3V with  $V_{REFP} = 3.3V$ ).
- IRQ Exit.

# 18.5 10-bit Analog to Digital Converter (ADC) Memory Map

Base Address ADC0: 0xFFFC0000

# Base Address ADC1: 0xFFFC4000

#### Table 18-1. ADC Memory Map

Offset	Register	Name	Access	Reset State
0x000 _ 0x04C	Reserved	-	-	-
0x050	Enable Clock Register	ADC_ECR	Write-only	_
0x054	Disable Clock Register	ADC_DCR	Write-only	_
0x058	Power Management Status Register	ADC_PMSR	Read-only	0x00000000
0x05C	Reserved	-	_	_
0x060	Control Register	ADC_CR	Write-only	_
0x064	Mode Register	ADC_MR	Read/Write	0x0000000
0x068	Conversion Mode Register	ADC_CMR	Read/Write	0x0000000
0x06C	Clear Status Register	ADC_CSR	Write-only	_
0x070	Status Register	ADC_SR	Read-only	0x00000000
0x074	Interrupt Enable Register	ADC_IER	Write-only	_
0x078	Interrupt Disable Register	ADC_IDR	Write-only	_
0x07C	Interrupt Mask Register	ADC_IMR	Read-only	0x00000000
0x080	Convert Data Register	ADC_DR	Read-only	0x00000000





# 18.6 ADC Enable Clock Register

Name: Access: Base Address:	ADC_E Write-or 0x050						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	_	-
W	W	W	W	W	W	W	W
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
W	W	W	W	W	W	W	W
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
W	W	W	W	W	W	W	W
7	6	5	4	3	2	1	0
—	—	—	_	_	_	ADC	-

# 18.7 ADC Disable Clock Register

Name: Access: Base Address:	ADC_D Write-or	ADC_DCR Write-only 0x054						
31	30	29	28	27	26	25	24	
-	_	-	-	-	-	-	-	
23	22	21	20	19	18	17	16	
-	-	-	-	-	-	-	-	
15	14	13	12	11	10	9	8	
-	-	-	-	-	-	-	-	
7	6	5	4	3	2	1	0	
-	_	_	_	_	_	ADC	_	

# 18.8 ADC Power Management Status Register

Name: Access: Base Address:	ADC_P Read-o 0x058		-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	_	_	_	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	—	-	_	_	_	ADC	_

### • ADC: ADC Clock Status

0: ADC clock disabled.

1: ADC clock enabled.





# **18.9 ADC Control Register**

Name: Access: Base Address:	ADC_C Write-or 0x060	R					
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	_	_	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	—	-
7	6	5	4	3	2	1	0
-	—	-	STOP	START	ADCDIS	ADCEN	SWRST

## • SWRST: ADC Software Reset

0: No effect.

1: Reset of the ADC peripheral.

When a software reset is performed, all the registers of the peripheral are reset.

#### • ADCEN: ADC Enable

0: No effect.

1: ADC is enabled for conversion.

### ADCDIS: ADC Disable

0: No effect.

1: ADC is disabled (Standby Mode).

#### • START: Start Conversion

0: No analog to digital conversion to be started.

1: Begin analog to digital conversion, clears EOC bit.

**Note:** Before starting conversions, users should ensure that the ADC is ready for conversion (READY bit is set to logical one in ADC\_SR).

### STOP: Stop Conversion in Continuous Conversion

0: No effect.

1: Stop the continuous conversion.

	mode negio						
Name: Access: Base Address	ADC_M Read/W : 0x064						
31	30	29	28	27	26	25	24
-	-	_	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	_	-	CONTCV		NBRCH[2:0]	
15	14	13	12	11	10	9	8
			STARTUR	PTIME[7:0]			
7	6	5	4	3	2	1	0
_	STOPEN	_	PRVAL[4:0]				

# 18.10 ADC Mode Register

#### • PRLVAL[4:0]: Preload Value

A division of the Master clock (MCLK) determines ADC\_clk. The preload value is chosen by the user to adapt the Master clock to the ADC peripheral as well as possible. It is the start value of the down counter. The LSB is fixed to 0 because this value has to be even parity to guaranty a duty cycle of ½, so the user has only to initialize the 5 MSB.

 $ADC_clk = CORECLK/(4xPRLVAL[4:0]).$ 

Note: The clock rate to the ADC must not exceed 700 kHz.

#### • STOPEN: Stop Enable

0: TEND cannot stop conversion when the device is in continuous mode.

1: TEND stops conversion when the device is in continuous conversion mode.

This bit is initialized to 0.

### • STARTUPTIME[7:0]: Startup Time

This value indicates the number of master clock periods necessary to make 4  $\mu$ s.

This time is the stabilization time of the analog circuitry.

For example, if CORECLK = 30 MHz, 120 periods of CORECLK are needed to obtain the absolute time of 4  $\mu$ s. So, the STARTUPTIME value of the mode register is 120 (0x78 in hexadecimal code).

### • NBRCH[2:0]: Number of Conversions

NBRCH[2:0]	Number of Conversions
000	1
001	2
010	3
011	4
100	5





NBRCH[2:0]	Number of Conversions
101	6
110	7
111	8

Note: Even in one shot mode, ADC will run multiple conversions if NBRCH[2:0] is greater than 000.

### CONTCV: Continuous Conversion

0: one shot mode. ADC converts as much inputs as specified by the NBRCH[2:0] in the order specified in the ADC\_CMR, and stops.

1: continuous mode. ADC converts as much inputs as specified by the NBRCH[2:0] in the order specified in the ADC\_CMR, and repeats. This bit is initialized to 0.

# 18.11 DC Conversion Mode Register

Name: Access: Base Address:	ADC_0 Read/V 0x068						
31	30	29	28	27	26	25	24
_	CV8[2:0]			—	CV7[2:0]		
23	22	21	20	19	18	17	16
_	CV6[2:0] – CV5[2:0]						
15	14	13	12	11	10	9	8
_	CV4[2:0] – CV3[2:0]						
7	6	5	4	3	2	1	0
-		CV2[2:0]		-		CV1[2:0]	

# • CVx[2:0]: Input Selection

 $x = \{1, 2, 3, 4, 5, 6, 7, 8\}$  is the conversion number.

CVx[2:0]	Input Selected
000	In0
001	In1
010	In2
011	In3
100	In4
101	In5
110	In6
111	In7





# 18.12 ADC Clear Status Register

Name: Access: Base Address:	ADC_C Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	_	-	-	_	-
15	14	13	12	11	10	9	8
-	_	-	_	_	-	_	-
7	6	5	4	3	2	1	0
_	_	-	-	TEND	OVR	-	-

# • OVR: Overrun Interrupt

0: No effect.

1: Clear OVR interrupt.

# • TEND: End of PDC Transfer Interrupt

0: No effect.

1: Clear TEND interrupt.

# 18.13 ADC Status Register

	Status negi						
Name: Access: Base Address:	ADC_C Read-o : 0x070						
31	30	29	28	27	26	25	24
-	—	-	_	_	_	_	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	—	CTVS	ADCENS
7	6	5	4	3	2	1	0
-	_	_	-	TEND	OVR	READY	EOR

#### • EOC: End of Conversion

0: Conversion not complete or inactive.

1: Conversion complete, data in ADC\_DR is valid.

This bit is cleared when the ADC\_DR is read.

### • READY: ADC Ready for Conversion

0: ADC ignores start or stop command. It is not ready for conversion or is converting data.

1: ADC is ready to start a conversion.

To explain more completely "ready\_flag", define "working" as the event high when ADC is converting data and "analog\_ready" the event low when the analog part is disabled or in the initializing phase.

analog_ready	working	ready_flag
0	0	0
0	1	0
1	0	1
1	1	0

### • OVR: Overrun

0: No data has been transferred by ADC from the last ADC\_DR read.

1: At least one data has been transferred by ADC since the last ADC\_DR read.

### • TEND: End of Total Transfer of PDC

0: Transfer of all data not complete.

1: PDC transfer complete.

This bit is set when the transfer of all the data by the PDC is complete. When we are in continuous mode, it stops the conversion only if the STOPEN bit of the mode register is active.

#### ADCENS: ADC Enable Status





0: ADC is disabled.

1: ADC is enabled.

### • CTCVS: Continuous Mode Status

0: One shot mode with help of microprocessor.

1: Continuous mode, the peripheral is stand-alone. This bit is initialized to 0 and changes when there is a change of mode. This bit never generates an interrupt.

# 18.14 ADC Interrupt Enable Register

Name: Access: Base Address:	ADC_IE Write-or 0x074						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	—	-	-	TEND	OVR	READY	EOR

# 18.15 ADC Interrupt Disable Register

Name: Access: Base Address:	ADC_ID Write-or 0x078						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	—	—	-	TEND	OVR	READY	EOR





# 18.16 ADC Interrupt Mask Register

Name: Access: Base Address:	ADC_IN Read-ou 0x07C						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	—	_	_	_	_	_	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	—	TEND	OVR	READY	EOR

### • EOC: End of Conversion

0: EOC interrupt is disabled.

1: EOC interrupt is enabled.

### • READY: ADC Ready for Conversion

0: READY interrupt is disabled.

1: READY interrupt is enabled.

#### • OVR: Overrun

0: OVR interrupt is disabled.

1: OVR interrupt is enabled.

### • TEND: End of PDC Transfer

0: TEND interrupt is disabled.

1: TEND interrupt is enabled.

# 18.17 ADC Convert Data Register

Name: Access: Base Address:	ADC_C Read-or 0x080							
31	30	29	28	27	26	25	24	
_	_	_	-	_	—	_	-	
23	22	21	20	19	18	17	16	
-	-	-	-	-	-	_	-	
15	14	13	12	11	10	9	8	
-	_	-	-	-	-	DATA[9:8]		
7	6	5	4	3	2	1	0	
			DAT	A[7:0]				

### • DATA[9:0]: Converted Data

The result data from an analog to digital conversion is latched into this register at the end of a conversion and remains valid until a new conversion is completed.

When this register is read, the EOC bit in the ADC\_SR register is cleared.

Note: When debugging, to avoid clearing the EOC bit, users should use ghost registers (see "Ghost Registers" on page 9).





# 19. Universal Synchronous/ Asynchronous Receiver/Transmitter (USART)

### 19.1 Overview

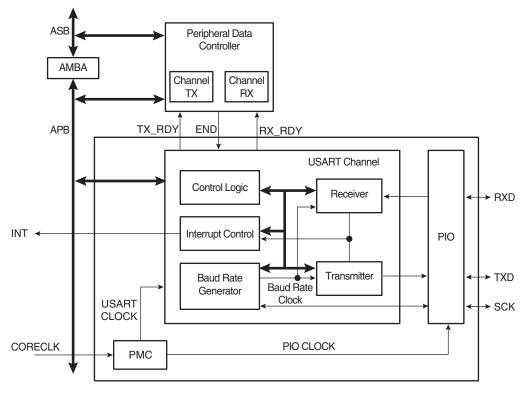
The AT91SAM7A2 includes two USARTs. Each transmitter and receiver module is connected to the Peripheral Data Controller.

The main features are:

- Programmable baud rate generator
- · Parity, framing and overrun error detection
- Supports Hardware LIN protocol (specification 1.2)
- Idle flag for J1587 protocol
- · Line break generation and detection
- · Automatic echo, local loopback and remote loopback channel modes
- Multi-drop mode: address detection and generation
- Interrupt generation
- Two dedicated Peripheral Data Controller channels per USART
- 5 to 9-bit character length

### 19.2 Block Diagram

Figure 19-1. USART Block Diagram



### 19.3 Baud Rate Generator

The baud rate generator provides the bit period clock, named the baud rate clock, to both the receiver and the transmitter.

The baud rate generator can select between external and internal clock sources. The external clock source is SCK. The internal clock sources can be either CORECLK or CORECLK divided by 8 (CORECLK/8).

**Note:** In all cases, if an external clock is used, the duration of each of its levels must be longer than the CORECLK period. The external clock frequency must be less than 40% of CORECLK frequency.

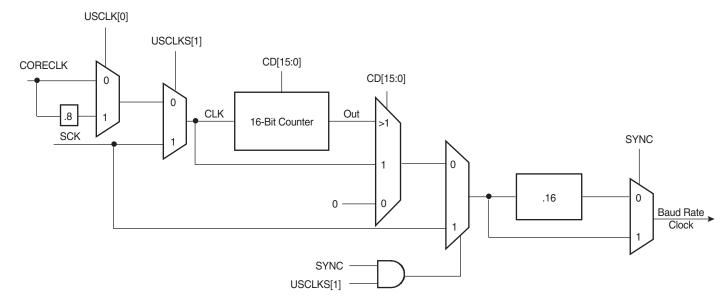
When the USART is programmed to operate in asynchronous mode (SYNC = 0 in the Mode Register US\_MR), the selected clock is divided by 16 times the value (CD) written in US\_BRGR (Baud Rate Generator Register). If US\_BRGR is set to 0, the baud rate clock is disabled.

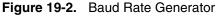
Baud rate = Selected Clock/16 x CD when the selected clock is either CORECLK, CORECLK/8 or SCK.

When the USART is programmed to operate in synchronous mode (SYNC = 1) and the selected clock is internal (USCLKS[1] = 0 in the Mode Register US\_MR), the Baud Rate Clock is the internal selected clock divided by the value written in US\_BRGR. If US\_BRGR is set to 0, the Baud Rate Clock is disabled.

#### Baud Rate = Selected Clock/CD

In synchronous mode with external clock selected (USCLKS[1] = 1), the clock is provided directly by the signal on the SCK pin. No division is active. The value written in US\_BRGR has no effect.





### 19.4 Receivers

The USART is configured with two receiver operating modes, one for asynchronous operations and the other for synchronous operations.

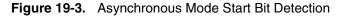


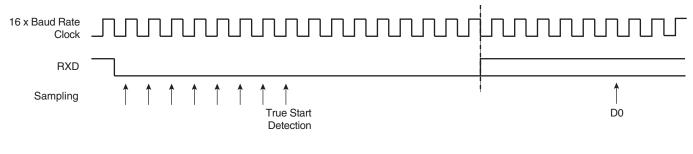


#### 19.4.1 Asynchronous Receiver

The USART is configured for asynchronous operation when SYNC = 0 (bit 7 of US\_MR). In asynchronous mode, the USART detects the start of a received character by sampling the RXD signal until it detects a valid start bit. A low level (space) on RXD is interpreted as a valid start bit if it is detected for more than 7 cycles of the sampling clock, which is 16 times the baud rate. Hence a space which is longer than 7/16 of the bit period is detected as a valid start bit. A space that is 7/16 of a bit period or shorter is ignored and the receiver continues to wait for a valid start bit.

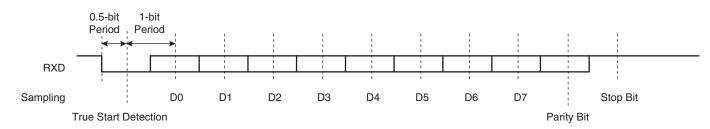
When a valid start bit has been detected, the receiver samples RXD at the theoretical mid-point of each bit. It is assumed that each bit lasts 16 cycles of the sampling clock (one bit period) so the sampling point is 8 cycles (0.5 bit periods) after the start of the bit. The first sampling point is therefore 24 cycles (1.5 bit periods) after the falling edge of the start bit was detected. Each subsequent bit is sampled 16 cycles (1 bit period) after the previous one.





#### Figure 19-4. Asynchronous Mode Character Reception

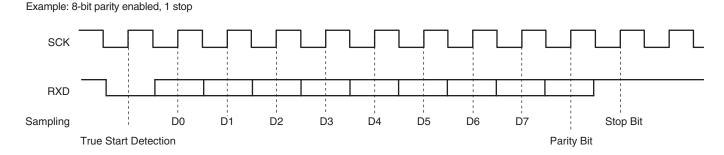
Example: 8-bit parity enabled, 1 stop



#### 19.4.2 Synchronous Receiver

When configured for synchronous operation (SYNC = 1), the receiver samples the RXD signal on each rising edge of SCK. If a low level is detected, it is considered as a start. Data bits, parity bit and stop bit are sampled and the receiver waits for the next start bit. See the example below.

### Figure 19-5. Synchronous Mode Character Reception



#### 19.4.3 Receiver Ready

When a complete character is received, it is transferred to the US\_RHR and the RXRDY status bit in US\_SR is set. The RXRDY is set after the last stop bit.

If US\_RHR has not been read since the last transfer, the USOVRE status bit in US\_SR is set.

#### 19.4.4 Parity Error

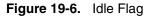
Each time a character is received, the receiver calculates the parity of the received data bits, in accordance with the PAR field in US\_MR. It then compares the result with the received parity bit. If different, the parity error bit PARE in US\_SR is set.

#### 19.4.5 Framing Error

If a character is received with a stop bit at low level and with at least one data bit at high level, a framing error is generated. This sets FRAME in US\_SR.

#### 19.4.6 Idle Flag for J1587 Protocol Frame

The idle flag turns low when USART receives a start bit and turns high at the end of a J1587 protocol frame (after 10 stop bits). An interrupt can be generated on the rising edge of the idle flag.





#### 19.4.7 Time Out

The time out function allows an idle condition on the RXD line to be detected. The maximum delay for which the USART should wait for a new character to arrive while the RXD line is inactive (high level) is programmed in US\_RTOR (Receiver Time Out Register). When this register is set to 0, no time out is detected.

Otherwise, the receiver waits for a first character and then initializes a counter which is decremented at each bit period and reloaded at each byte reception. When the counter reaches 0, the TIMEOUT bit in US\_SR is set. The user starts (or restarts) the wait for a first character by setting the STTTO (start time out) bit in US\_CR.





To start a time out, the following conditions must be met:

- US\_RTOR must not be equal to 0.
- The time out must be started by setting STTTO to a logical 1 in the US\_CR register.
- One character must be received.

Calculation of time out duration in asynchronous mode:

Duration = Value x 4 x Bit period.

## 19.5 Transmitter

The transmitter has the same behavior in both synchronous and asynchronous operating modes. Start bit, data bits, parity bit and stop bits are serially shifted, least significant bit first, on the falling edge of the serial clock.

The number of data bits is selected in the CHRL field in US\_MR.

The parity bit is set according to the PAR field in US\_MR.

The number of stop bits is selected in the NBSTOP field in US\_MR.

When a character is written to US\_THR (Transmit Holding Register), it is transferred to the Shift Register as soon as it is empty.

When the transfer occurs, the TXRDY bit in US\_SR is set until a new character is written to US\_THR. If the Transmit Shift Register and US\_THR are both empty, the TXEMPTY bit in US\_SR is set (after the last stop bit of the last transfer).

#### 19.5.1 Time Guard

The Time Guard function allows the transmitter to insert an idle state on the TXD line between two characters. The duration of the idle state is programmed in US\_TTGR (Transmitter Time Guard Register). When this register is set to zero, no time-guard is generated. Otherwise, the transmitter holds a high level on TXD after each transmitted byte during the number of bit periods programmed in US\_TTGR.

#### 19.5.2 Multi Drop Mode

When the PAR field in US\_MR equals  $11X_b$ , the USART is configured to run in multi drop mode. In this case, the parity error bit (PARE in US\_SR) is set when data is detected with a parity bit set to identify an address byte. PARE is cleared with the Reset Status Bits Command (RSTSTA) in US\_CR. If the parity bit is detected low, identifying a data byte, PARE is not set.

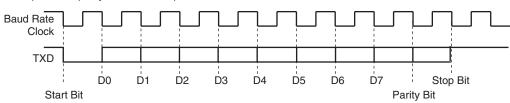
The transmitter sends an address byte (parity bit set) when a Send Address Command (SENDA) is written to US\_CR.

In this case, the next byte written to US\_THR will be transmitted as an address. After this any byte transmitted will have the parity bit cleared.

Idle state duration between two characters = Time Guard Value x Bit Period



Example: 8-bit, parity enabled, 1 stop



## 19.6 Break Condition

#### 19.6.1 Transmit Break

The transmitter can generate a break condition on the TXD line when the STTBRK command is set in US\_CR (Control Register). In this case, the characters present in US\_THR and in the Transmit Shift Register are completed before the line is held low.

To remove this break condition on the TXD line, the STPBRK command in US\_CR must be set. The USART generates a minimum break duration of one character length.

The TXD line then returns to high level (idle state) for at least 12 bit periods to ensure that the end of break is correctly detected. Then the transmitter resumes normal operation.

#### 19.6.2 Receive Break

The break condition is detected by the receiver when all data, parity and stop bits are low (1 character length set to low). At the moment of the low stop bit detection, the receiver asserts the RXBRK bit in US\_SR.

End of receive break is detected by a high level for at least 2/16 of the bit period in asynchronous operating mode or at least one sample in synchronous operating mode. RXBRK is also set after end of break has been detected.

#### 19.6.3 Interrupt Generation

Each status bit in US\_SR has a corresponding bit in US\_IER (Interrupt Enable Register) and US\_IDR (Interrupt Disable Register) which controls the generation of interrupts by asserting the USART interrupt line connected to the Generic Interrupt Controller. US\_IMR (Interrupt Mask Register) indicates the status of the corresponding bits.

When a bit is set in US\_SR and the same bit is set in US\_IMR, the interrupt line is asserted.

#### 19.6.4 Channel Modes

The USART can be programmed to operate in three different test modes, using the CHMODE field in US\_MR.

Automatic Echo Mode provides bit by bit retransmission. When a bit is received on the RXD line, it is sent to the TXD line. Programming the transmitter has no effect

Local Loopback Mode enables the transmitted characters to be received. TXD and RXD pins are not used and the output of the transmitter is internally connected to the input of the receiver. The RXD pin level has no effect and the TXD pin is held high, as in idle state.

Remote Loopback Mode directly connects the RXD pin to the TXD pin. The Transmitter and the Receiver are disabled and have no effect. This mode provides bit by bit retransmission.





## 19.7 LIN Protocol

To ensure that the checksum is sent correctly in LIN transmission, the following procedure should be followed: Use the PDC instead of US\_DFWR0, US\_DFWR1 registers and the STRESP bit to transmit data frame + check byte. The user must develop his own checksum compute function, and transmits this checksum with the PDC module. Do not use the STRESP bit in the CR that starts data transmission with the content of US\_DFWR0, US\_DFWR1.

This section describes the LIN protocol supported by the USART.

It supports Hardware LIN protocol revision 1.2.

If the LIN is set in the Mode Register, the USART works as a "Master Control Unit" (as it is defined in the LIN Protocol Specification).

It generates the "HEADER FRAME" automatically if the STHEADER bit is set on the Controller Register and the end of header is signaled through the ENDHEADER bit on the Status Register. The header content concerning the "IDENTIFIER" is defined by the Identifier Register (on Read/Write access). The parity is automatically calculated. In the header part, the SYNC\_BREAK LENGTH is configurable through the Sync Break Length Register (the SYNC\_BREAK can be set from 8 Tbits up to 31Tbits.

Because bit transfer in LIN Protocol depends on the bit rate and because of the large difference between Slave and Master bit rates in LIN specifications, the Tbit represents bit time based on the Master time base. Hence, 1 Tbit = the Master bit rate.

To send a "RESPONSE" (see the LIN Protocol Specification), the STRESP bit is set in the Controller Register, then a part or the totality of Data Field Write 0 Register content and the Data Field Write1 Register content is sent (depending on the value of ID5 and ID4). The CHECKSUM FIELD is also automatically calculated and sent. It is also possible to fill the message response to the THR register. Then if the software wants to use the PDC, it has to wait for the END HEADER interrupt and to start the PDC.

The "MESSAGE RESPONSE" content which has been received by the USART is available on the Data Field Read Register content and the Data Field Read 1 Register even if the USART has filed the "MESSAGE RESPONSE".

The USART is able to detect and to signal the end of "LIN MESSAGE" through the ENDMESS bit in the Status Register. Moreover the USART detects the Bit-Error, the Identifier-Parity-Error, Not-Responding-Error, the Checksum-Error and the Wakeup.

It is not possible to write in the LIN Identifier Register (US\_LIR) and in the Data Field Write Register (US\_DFWR) during a Message Transfer.

It is also advised not to write on the Transmitter Holding Register (US\_THR) during the Header.

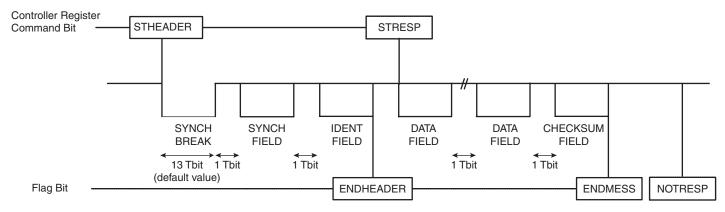
The interbyte space is configurable through the TIME GUARD register, the default interbyte space is one Tbit.

## **19.8 Line Configuration in LIN Mode**

To work in LIN mode, the USART has to be set in normal mode (see "USART Mode Register" on page 168). Moreover, when using USART's LIN function, the module must have its transmitter and its receiver enabled.

## 19.9 Message Characteristics





In Figure 19-8 above, unless the master has to file the "Response", no bits are to be set on the Control Register to wait for the slave's data field.

If the "Response" is not correctly ended, a time out flag rises 91, 119 or 175 Tbits (depending if Ndata = 2, 4 or 8 bytes) after the HEADER's beginning.

## 19.10 Smart Card Protocol (ISO7816-3)

The USARTs are ISO7816-3 compliant and allow character repetition or error signaling on parity errors.

Following below is a description of the functions available if the SMCARDPT bit is set on the US\_MR register.

#### 19.10.1 Character Transmission to Smart Card

To perform character transmission to a smart card, the transmitter and the receiver have to be enabled.

If the SMCARDPT bit is set in the US\_MR register, the USART is able to detect that the smart card has not correctly received the last transmitted byte (parity error signaled by the card).

When the card generates the error signal, the last transmitted byte is retransmitted again as many times as determined in the SCREPEAT[1:0] bits of the US\_MR register until the error signal is no longer generated by the smart card.

When the error signal is detected by the USART, the PARE and FRAME error flags are set in the US\_SR register.

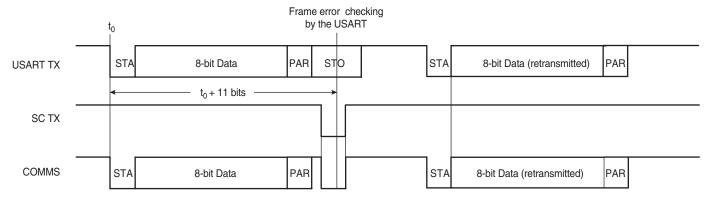
The error signal is checked by the USART at t0 + 11 bits where  $t_0$  is the falling edge of the start bit (i.e. between the two stop bits).

In the example shown Figure 19-9 on page 150, a parity error is detected by the smart card. The smart card generates the error signal on the COMMS line. The error signal is detected by the USART which retransmits the last character.





## Figure 19-9. Smart Card Transmission Error



If a PDC transfer is used to send a data byte to the card, the PDC counter will not be decremented and the PDC memory pointer will not be incremented until the card has received a correct byte or the maximum repetition time has been reached.

#### 19.10.2 Character Reception From Smart Card

To obtain character reception from a smart card, the receiver has to be enabled and the transmitter has to be disabled.

If the SMCARDPT bit is set in the US\_MR register, the USART is able to generate the error signal (see ISO7816-3 protocol) when the last byte received has a parity error.

When a parity error is detected by the USART, the USART transmission line is driven low for 1.0625 bit period starting at  $t_0 + 10.625 + [0:0.0625]$  (i.e. during the two stop bits) to indicate to the smart card that a bad reception has occurred on the USART. With T = 0 protocol type smart cards, the smart card must resend the last character.

In that case, the USART signals only a parity error by setting the PARE bit in the US\_SR register.

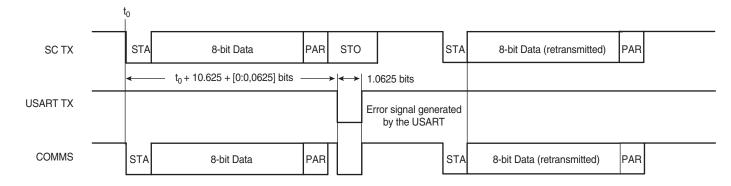


Figure 19-10. Error Signaling on Reception

If a PDC transfer is used to receive a data byte from the card, the PDC counter will be decremented and the PDC memory pointer will be incremented even when a parity error is detected. The user must reconfigure the PDC memory pointer (decrement by 1) and counter (increment by 1) in order to receive all the remaining bytes.

## 19.11 Line Configuration in Smart Card Mode

To work in Smart Card mode, the USART has to be set in normal mode (see the "USART Mode Register" on page 168).

## **19.12 PIO Controller**

Each USART has three programmable I/O lines. These lines are multiplexed with signals (TXD, RXD, SCK) of the USART to optimize the use of the available package pins. These lines are controlled by the USART PIO controller.

### 19.13 Power Management

The USART is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).

#### 19.14 Example

Example USART usage with LIN functionality:

Send a Header Frame and send the Message Frame with an identification of 0x30 with a message of 2x16 bit at 19600 bps using interrupt.

#### 19.14.1 Configuration

- Enable the clock on USART and PIO peripheral by writing the USART bit in US\_ECR.
- Do a software reset of the USART peripheral to be in a known state by writing the SWRST bit in US\_CR.
- Disable PIO pins RXD, TXD, SCK in US\_PDR.
- Configuration of US\_BRGR: For 19600 bps field CD = Core clock/(16\*19600).
- Enable Transmitter and Receiver bits, RXEN and TXEN, in US\_CR.
- Configuration of US\_MR: The standard configuration is preferred.
  - No parity, 8 bits, 1 stop bit, normal mode, LIN supported, USART clock is Coreclk
- Configuration of US\_SBLR: Configure the syncbreak to 0x13.
- Enter the data to send in DFWR0 and DFWR1.
- Configuration of US\_IER: Generate an interrupt at the end of the header and message transmission. This is done by writing the ENDHEADER, ENDMESS. The user can associate interrupts to errors; such as BIT error, parity error, checksum error, etc, (GIC must be configured).
- Configure the identifier by writing 0x30 in US\_LIR.
- Enable the LIN header transmission by writing the STHEADER bit in US\_CR, this will send the header. Before going to the next step, the user should be informed by the interrupt that the header has been sent.
- Enable the LIN response transmission by writing the STRESP bit in US\_CR, this will send the message. Then an interrupt is generated at the end of the transmission.

#### 19.14.2 Interrupt Handling

- IRQ Entry and call C function.
- Read US\_SR and verify the source of the interrupt. This register is read and cleared for certain status fields. Care should be taken to keep status information so as to be able to





proceed in all cases. Fields that are present in US\_CSR are not read and cleared and should be cleared in the next step.

- Clear the corresponding interrupt at peripheral level by writing in the ST\_CSR.
- Interrupt treatment: Inform background software that header or message is transmitted.
- IRQ Exit.

## 19.15 USART Memory Map

Base Address USART0: 0xFFFA8000

Base Address USART1: 0xFFFAC000

Table 19-1.USART Memory Map

Offset	Register	Name	Access	Reset State
0x000	PIO Enable Register	US_PER	Write-only	_
0x004	PIO Disable Register	US_PDR	Write-only	-
0x008	PIO Status Register	US_PSR	Read-only	0x00070000
0x00C	Reserved	_	-	-
0x010	Output Enable Register	US_OER	Write-only	_
0x014	Output Disable Register	US_ODR	Write-only	_
0x018	Output Status Register	US_OSR	Read-only	0x00000000
0x01C _ 0x02C	Reserved	-	_	_
0x030	Set Output Data Register	US_SODR	Write-only	_
0x034	Clear Output Data Register	US_CODR	Write-only	_
0x038	Output Data Status Register	US_ODSR	Read-only	0x0000000
0x03C	Pin Data Status Register	US_PDSR	Read-only	0x000X0000
0x040	Multi-Driver Enable Register	US_MDER	Write-only	_
0x044	Multi-Driver Disable Register	US_MDDR	Write-only	_
0x048	Multi-Driver Status Register	US_MDSR	Read-only	0x0000000
0x04C	Reserved	-	_	_
0x050	Enable Clock Register	US_ECR	Write-only	_
0x054	Disable Clock Register	US_DCR	Write-only	_
0x058	Power Management Status Register	US_PMSR	Read-only	0x0000000
0x05C	Reserved	-	-	_
0x060	Control Register	US_CR	Write-only	0x00000000
0x064	Mode Register	US_MR	Read/Write	0x0000000
0x068	Reserved	-	_	_
0x06C	Clear Status Register	US_CSR	Write-only	0x00000000
0x070	Status Register	US_SR	Read-only	0x0000800
0x074	Interrupt Enable Register	US_IER	Write-only	_
0x078	Interrupt Disable Register	US_IDR	Write-only	_
0x07C	Interrupt Mask Register	US_IMR	Read-only	0x0000000
0x080	Receiver Holding Register	US_RHR	Read-only	0x0000000
0x084	Transmitter Holding Register	US_THR	Write-only	_
0x088	Baud Rate Generator Register	US_BRGR	Read/Write	0x0000000





## Table 19-1. USART Memory Map

Offset	Register	Name	Access	Reset State
0x08C	Receiver Time-out Register	US_RTOR	Read/Write	0x00000000
0x090	Transmitter Time-guard Register	US_TTGR	Read/Write	0x00000000
0x094	LIN Identifier Register	US_LIR	Read/Write	0x00000000
0x098	Data Field Write0 Register	US_DFWR0	Read/Write	0x00000000
0x09C	Data Field Write 1 Register	US_DFWR1	Read/Write	0x00000000
0x0A0	Data Field Read 0 Register	US_DFRR0	Read-only	0x00000000
0x0A4	Data Field Read 1 Register	US_DFRR1	Read-only	0x00000000
0x0A8	Sync Break Length Register	US_SBLR	Read/Write	0x000000D

## 19.16 USART PIO Enable Register

Name: Access: Base Address:	US_PE Write-or	R					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	—	_	_	_	-

## 19.17 USART PIO Disable Register

Name: Access: Base Address:	US_PDI Write-or 0x004	R					
31	30	29	28	27	26	25	24
_	_	-	-	-	-	_	_
23	22	21	20	19	18	17	16
-	-	_	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	-	-	-	-	_





## 19.18 USART PIO Status Register

Name: Access: Base Address:	US_PSI Read-or 0x008						
31	30	29	28	27	26	25	24
-	_	_	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	-	-	-	-

## • SCK: SCK Pin

0: PIO is inactive on the SCK pin.

1: PIO is active on the SCK pin.

## • TXD: TXD Pin

0: PIO is inactive on the TXD pin.

1: PIO is active on the TXD pin.

#### • RXD: RXD Pin

0: PIO is inactive on the RXD pin.

1: PIO is active on the RXD pin.

## 19.19 USART PIO Output Enable Register

Name: Access: Base Address:	US_OE Write-or 0x010		5				
31	30	29	28	27	26	25	24
-	_	-	-	_	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	-	-	-	-

## 19.20 USART PIO Output Disable Register

Name: Access: Base Address:	US_OD Write-or 0x014	R					
31	30	29	28	27	26	25	24
_	-	_	-	-	_	—	_
23	22	21	20	19	18	17	16
-	-	_	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	-	_	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	_	-	_	_	_





## 19.21 USART PIO Output Status Register

Name: Access: Base Address:	US_OS Read-or 0x018		-				
31	30	29	28	27	26	25	24
-	—	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	_	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	—	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	-	-	-	-	-

## • SCK: SCK Pin

0: PIO is an input on the SCK pin.

1: PIO is an output on the SCK pin.

## • TXD: TXD Pin

0: PIO is an input on the TXD pin.

1: PIO is an output on the TXD pin.

#### • RXD: RXD Pin

0: PIO is an input on the RXD pin.

1: PIO is an output on the RXD pin.

# 19.22 USART PIO Set Output Data Register

Name: Access: Base Address:	US_SO Write-oi						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	—	-	-	—	—	-	-
7	6	5	4	3	2	1	0
_	-	_	-	_	-	—	-

## 19.23 USART PIO Clear Output Data Register

Name: Access: Base Address:	US_CODR Write-only 0x034									
31	30	29	28	27	26	25	24			
-	_	_	-	_	-	_	-			
23	22	21	20	19	18	17	16			
-	—	_	-	-	RXD	TXD	SCK			
15	14	13	12	11	10	9	8			
-	_	_	-	-	-	-	-			
7	6	5	4	3	2	1	0			
-	_	_	-	-	_	_	-			





## 19.24 USART PIO Output Data Status Register

Name: Access: Base Address:	US_OD Read-or 0x038		-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	_	-	_	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	-	-	-	-	-	-

### • SCK: SCK Pin

0: The output data for the SCK pin is programmed to 0.

1: The output data for the SCK pin is programmed to 1.

## • TXD: TXD Pin

0: The output data for the TXD pin is programmed to 0.

1: The output data for the TXD pin is programmed to 1.

#### • RXD: RXD Pin

0: The output data for the RXD pin is programmed to 0.

1: The output data for the RXD pin is programmed to 1.

## 19.25 USART PIO Pin Data Status Register

Name: Access: Base Address:	US_PD Read-o 0x03C						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	-	-	—	-	-	-
7	6	5	4	3	2	1	0
—	—	_	_	_	_	—	-

## • SCK: SCK Pin

0: The output data for the SCK pin is programmed to 0.

1: The output data for the SCK pin is programmed to 1.

## • TXD: TXD Pin

0: The output data for the TXD pin is programmed to 0.

1: The output data for the TXD pin is programmed to 1.

## • RXD: RXD Pin

0: The output data for the RXD pin is programmed to 0.

1: The output data for the RXD pin is programmed to 1.





## 19.26 USART PIO Multi Drive Enable Register

Name: Access: Base Address:	US_MD Write-oi 0x040		-				
31	30	29	28	27	26	25	24
-	_	_	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	—	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	—	—	-	-	-	-

## 19.27 USART PIO Multi Drive Disable Register

Name: Access: Base Address:	US_MD Write-or 0x044						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	-	_	-	_	-

## 19.28 USART PIO Multi Drive Status Register

Name: Access: Base Address:	US_MD Read-o 0x048						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18 RXD	17 TXD	16 SCK
	_	_	_		TIXD	TAD	301
15	14	13	12	11	10	9	8
—	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	-	_	—	—	—	-

### • SCK: SCK Pin

0: The SCK pin is not configured as an open drain.

1: The SCK pin is configured as an open drain.

## • TXD: TXD Pin

0: The TXD pin is not configured as an open drain.

1: The TXD pin is configured as an open drain.

#### • RXD: RXD Pin

0: The RXD pin is not configured as an open drain.

1: The RXD pin is configured as an open drain.





## 19.29 USART Enable Clock Register

Name: Access: Base Address:	US_EC Write-or 0x050						
31	30	29	28	27	26	25	24
_	_	_	_	_	—	_	—
23	22	21	20	19	18	17	16
-	_	-	-	-	-	—	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	-	-	-	-	USART	PIO

## 19.30 USART Disable Clock Register

Name: Access: Base Address:	US_DCI Write-or 0x054	R C					
31	30	29	28	27	26	25	24
-	-	_	-	-	_	-	-
23	22	21	20	19	18	17	16
-	-	_	-	-	_	-	-
15	14	13	12	11	10	9	8
-	-	_	-	-	-	-	-
7	6	5	4	3	2	1	0
_	-	_	-	-	-	USART	PIO

## 19.31 USART Power Management Status Register

Name: Access: Base Address:	US_PM Read-o 0x058		Ū				
31	30	29	28	27	26	25	24
-	_	-	-	_	-	-	-
23	22	21	20	19	18	17	16
-	_	-	_	_	-	_	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	-	-	-	-	-	USART	PIO

## • PIO: PIO Clock Status

0: PIO clock disabled.

1: PIO clock enabled.

## • USART: USART Clock Status

0: USART clock disabled.

1: USART clock enabled.

Note: The US\_PMSR register is not reset by software reset.





## 19.32 USART Control Register

Name: Access: Base Address:	US_CR Write-oi 0x060						
31	30	29	28	27	26	25	24
-	—	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	_	_	STRESP	STHEADER
15	14	13	12	11	10	9	8
-	_	-	SENDA	STTO	STPBRK	STTBRK	RSTSTA
7	6	5	4	3	2	1	0
TXDIS	TXEN	RXDIS	RXEN	RSTTX	RSTRX	-	SWRST

## • SWRST: Software Reset

0: No effect.

1: Reset the USART.

A software triggered hardware reset of the USART is performed. It resets all the registers, including PIO registers (except US\_PMSR).

### • RSTRX: Reset Receiver

0: No effect.

1: The receiver logic is reset.

### • RSTTX: Reset Transmitter

0: No effect.

1: The transmitter logic is reset.

#### • RXEN: Receiver Enable

0: No effect.

1: The receiver is enabled if RXDIS is 0.

## • RXDIS: Receiver Disable

0: No effect.

1: The receiver is disabled.

## • TXEN: Transmitter Enable

0: No effect.

1: The transmitter is enabled if TXDIS is 0.

## • TXDIS: Transmitter Disable

0: No effect.

1: The transmitter is disabled.

166 AT91SAM7A2

## • RSTSTA: Reset Status Bit

0: No effect.

1: Resets the status bits PARE, FRAME, USOVRE and RXBRK in US\_SR.

#### • STTBRK: Start Break

0: No effect.

1: If break is not being transmitted, start transmission of a break after the characters present in US\_THR and the Transmit Shift Register have been transmitted.

### • STPBRK: Stop Break

0: No effect.

1: If a break is being transmitted, stop transmission of the break after a minimum of one character length and transmit a high level during 12 bit periods.

### • STTTO: Start Time-out

0: No effect.

1: Start waiting for a character before clocking the time-out counter.

### • SENDA: Send Address

0: No effect.

1: In Multi-drop Mode only, the next character written to the US\_THR is sent with the address bit set.

#### • STHEADER: Start Header

0: No effect.

1: If the LIN bit is set on the Mode Register, send the LIN 's Header Frame.

## • STRESP: Start Response

0: No effect.

1: Send a part or the Data Field 0 Register content and Data Field 1 Register content.





## 19.33 USART Mode Register

Name: Access: Base Address:	US_MR Read/Writ 0x064	te					
31	30	29	28	27	26	25	24
-	-	_	-	-	_	_	-
23	22	21	20	19	18	17	16
-	-	_	-	-	CLKO	MODE9	SMCARDPT
15	14	13	12	11	10	9	8
CHMOD	E[1:0]	NBST	OP[1:0]		PAR[2:0]		SYNC
7	6	5	4	3	2	1	0
CHRL[	1:0]	USCI	_KS[1:0]	SENDT	IME[1:0]	_	LIN

## LIN: Local Interconnect Network Mode

0: USART does not support LIN protocol.

1: USART supports LIN protocol.

## • SENDTIME [1:0]: Send Time

Indicates the maximum number of times that the USART has to send data in case of a bad reception of the smart card linked to this same USART.

SENDTIME[1:0]		Number of Time
0	0	0
0	1	1
1	0	2
1	1	3

## • .USCLKS[1:0]: Clock Selection (Baud Rate Generator Input Clock)

USCLKS[1:0]		Selected Clock
0	0	CORECLK
0	1	CORECLK/8
1	Х	External (SCK)

## • CHRL[1:0]: Character Length

Start, stop and parity bits are added to the character length.

CHR	L[1:0]	Character Length
0	0	5 bits
0	1	6 bits
1	0	7 bits
1	1	8 bits

### • SYNC: Synchronous Mode Select

0: USART operates in Asynchronous Mode.

1: USART operates in Synchronous Mode.

### • PAR[2:0]: Parity Type

	PAR[2:0]	Parity Type	
0	0	0	Even Parity
0	0	1	Odd Parity
0	1	0	Parity forced to 0 (Space)
0	1	1	Parity forced to 1 (Mark)
1	0	х	No parity
1	1	х	Multi-drop mode

## • NBSTOP[1:0]: Number of Stop Bits

The interpretation of the number of stop bits depends on SYNC.

NBSTO	DP[1:0]	Asynchronous (SYNC = 0)	Synchronous (SYNC = 1)		
0	0	1 stop bit	1 stop bit		
0	1	1.5 stop bits	Reserved		
1	0	2 stop bits	2 stop bits		
1	1	Reserved	Reserved		

## • CHMODE[1:0]: Channel Mode

СНМО	DE[1:0]	Mode Description			
0	0 0 Normal Mode: The USART Channel operates as a Rx/Tx USART				
0	1	Automatic Echo: Receiver Data Input is connected to TXD pin			
1	0	Local Loopback: Transmitter Output Signal is connected to Receiver Input Signal			
1 1 Remote Loopback: RXD pin is internally connected to TXD pin					





## • SMCARDPT: Smart Card Protocol

0: The USART is not in smart card protocol.

1: The USART is in smart card protocol.

## • MODE9: 9-bit Character Length

0: CHRL defines character length.

1: 9-Bit character length.

### • CLKO: Clock Output Select

- 0: The USART does not drive the SCK pin.
- 1: The USART drives the SCK pin if CLKS[1] is 0.

## 19.34 USART Clear Status Register

Name: Access: Base Address:	US_CS Write-or 0x06C						
31	30	29	28	27	26	25	24
—	WAKEUP	CHECKSUM	IPERROR	BITERROR	NOTRESP	ENDMESS	ENDHEADER
23	22	21	20	19	18	17	16
15	14	13	12	11	10	9	8
-	_	_	_	_	_	_	_
7	6	5	4	3	2	1	0
-	_	_	_	_	_	_	-

## • ENDHEADER: End of Header

0: No effect.

1: Clear ENDHEADER interrupt.

### • ENDMESS: End of Message

0: No effect.

1: Clear ENDMESS interrupt.

#### • NOTRESP: Not Responding

0: No effect.

1: Clear NOTRESP interrupt.

### • BITERROR: Bit Error

0: No effect.

1: Clear BITERROR interrupt.

### • IPERROR: Identity Parity Error

0: No effect.

1: Clear IPERROR interrupt.

## • CHECKSUM: Check Sum

0: No effect.

1: Clear CHECKSUM interrupt.

#### • WAKEUP: Wake Up

0: No effect.

1: Clear WAKEUP interrupt.





## 19.35 USART Status Register

Name: Access: Base Address:	US_SR Read-oi 0x070						
31	30	29	28	27	26	25	24
-	WAKEUP	CHECKSUM	IPERROR	BITERROR	NOTRESP	ENDMESS	ENDHEADER
23	22	21	20	19	18	17	16
-	_	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	-	-	IDLEFLAG	IDLE	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	USOVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

Note: This register is a "read-active" register, which means that reading it can affect the state of some bits. When reading US\_SR register, the following bits are cleared if set: IDLE, ENDRX, ENDTX, SCK, TXD and RXD. When debugging, to avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

## • RXRDY: Receiver Ready

0: No complete character has been received since the last read of the US\_RHR or the receiver is disabled.

1: At least one complete character has been received and the US\_RHR has not yet been read.

### • TXRDY: Transmitter Ready

0: A character is in US\_THR waiting to be transferred to the Transmit Shift Register or the transmitter is disabled.

1: There is no character in US\_THR.

Equal to zero when the USART is disabled or at reset. Transmitter Enable command (in US\_CR) sets this bit to one.

## RXBRK: Break Received/End

0: No Break Received and no End of Break has been detected since the last "Reset Status Bits" command in the Control Register.

1: Break Received or End of Break has been detected since the last "Reset Status Bits" command in the Control Register.

## • ENDRX: End of PDC Receiver Transfer

0: The End of Transfer signal from the Peripheral Data Controller channel dedicated to the receiver is inactive.

1: The End of Transfer signal from the Peripheral Data Controller channel dedicated to the receiver is active.

#### • ENDTX: End of PDC Transmitter Transfer

0: The End of Transfer signal from the Peripheral Data Controller channel dedicated to the transmitter is inactive.

1: The End of Transfer signal from the Peripheral Data Controller channel dedicated to the transmitter is active.

#### • USOVRE: Overrun Error

0: No byte has been transferred from the Receive Shift Register to the US\_RHR when RxRDY was asserted since the last "Reset Status Bits" command.

AT91SAM7A2

1: At least one byte has been transferred from the Receive Shift Register to the US\_RHR when RxRDY was asserted since the last "Reset Status Bits" command.

### • FRAME: Framing Error

0: No stop bit has been detected low since the last "Reset Status Bits" command.

1: At least one stop bit has been detected low since the last "Reset Status Bits" command.

#### • PARE: Parity Error

0: No parity bit has been detected false (or a parity bit high in multi-drop mode) since the last "Reset Status Bits" command.

1: At least one parity bit has been detected false (or a parity bit high in multi-drop mode) since the last "Reset Status Bits" command.

## • TIMEOUT: Receiver Time Out

0: There has not been a time-out since the last "Start Time-out" command or the Time-out Register is 0.

1: There has been a time-out since the last "Start Time-out" command.

### • TXEMPTY: Transmitter Empty

0: There are characters in either US\_THR or the Transmit Shift Register.

1: There are no characters in either US\_THR or the Transmit Shift Register.

Equal to zero when the USART is disabled or at reset. Transmitter Enable command (in US\_CR) sets this bit to one.

#### • IDLE: Idle Interrupt

0: No end of J1587 protocol frame since last read of the US\_SR register.

1: An end of J1587 protocol frame occurred since last read of the US\_SR register.

#### • IDLEFLAG: Idle Flag

0: A frame is being received by the USART.

1: No frame is being received by the USART.

This bit indicates a frame transmission in J1587 protocol. It's turned low when a reception starts and turned high when a reception is followed by at least 10 stop bits (10 bits at high level).

#### • SCK, TXD, RXD: PIO Interrupt Status

These bits indicate for each pin when an input logic value change has been detected (rising or falling edge). This is valid whether the PIO is selected for the pin or not.

These bits are reset to zero following a read and at reset.

0: No input change has been detected on the corresponding pin since the register was last read.

1: At least one input change has been detected on the corresponding pin since the register was last read.

#### • ENDHEADER: End of Header

0: No end of Header has occurred on a LIN Frame.

1: An end of Header has occurred on a LIN Frame.

#### • ENDMESS: End of Message

0: No end of Message occurred on a LIN Frame.





1: An end of Message has occurred on a LIN Frame.

### • NOTRESP: Not Responding

- 0: No Slave-not-responding-error has been detected LIN Frame.
- 1: A Slave-not-responding-error has been detected LIN Frame.

### • BITERROR: Bit Error

- 0: No Bit-error has been detected on a LIN Frame.
- 1: A Bit-error has been detected on a LIN Frame.

### • IPERROR: Identity Parity Error

- 0: No Identity-parity-Error has been detected on a LIN Frame.
- 1: An Identity-parity-error has been detected on a LIN Frame.

### • CHECKSUM: Check Sum

- 0: No Checksum-error has been detected LIN Frame.
- 1: A Checksum-error has been detected LIN Frame.

### • WAKEUP: Wake Up

- 0: No Wakeup has been detected.
- 1: A Wakeup has been detected.

## 19.36 USART Interrupt Enable Register

Name: Access: Base Address:	US_IEF Write-or 0x074						
31	30	29	28	27	26	25	24
-	WAKEUP	CHECKSUM	IPERROR	BITERROR	NOTRESP	ENDMESS	ENDHEADER
23	22	21	20	19	18	17	16
-	—	—	—	—	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	—	-	-	—	IDLE	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	USOVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

## 19.37 USART Interrupt Disable Register

Name: Access: Base Address:	US_IDF Write-or 0x078						
31	30	29	28	27	26	25	24
-	WAKEUP	CHECKSUM	IPERROR	BITERROR	NOTRESP	ENDMESS	ENDHEADER
23	22	21	20	19	18	17	16
-	_	—	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	_	—	-	-	IDLE	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	USOVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY





## 19.38 USART Interrupt Mask Register

Name: Access: Base Address:	US_IMF Read-or 0x07C						
31	30	29	28	27	26	25	24
-	WAKEUP	CHECKSUM	IPERROR	BITERROR	NOTRESP	ENDMESS	ENDHEADER
23	22	21	20	19	18	17	16
-	-	-	-	-	RXD	TXD	SCK
15	14	13	12	11	10	9	8
-	-	-	-	IDLEFLAG	IDLE	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	USOVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

### • RXRDY: Mask RXRDY Interrupt

0: RXRDY Interrupt is Disabled.

1: RXRDY Interrupt is Enabled.

## • TXRDY: Mask TXRDY Interrupt

0: TXRDY Interrupt is Disabled.

1: TXRDY Interrupt is Enabled.

#### • RXBRK: Mask Receiver Break Interrupt

0: RXBRK Interrupt is Disabled.

1: RXBRK Interrupt is Enabled.

## • ENDRX: Mask End of PDC Receive Transfer Interrupt

0: ENDRX Interrupt is Disabled.

1: ENDRX Interrupt is Enabled.

#### • ENDTX: Mask End of PDC Transmit Transfer Interrupt

0: ENDTX Interrupt is Disabled.

1: ENDTX Interrupt is Enabled.

## USOVRE: Mask Overrun Error Interrupt

0: USOVRE Interrupt is Disabled.

1: USOVRE Interrupt is Enabled.

#### • FRAME: Mask Framing Error Interrupt

0: FRAME Interrupt is Disabled.

1: FRAME Interrupt is Enabled.

#### • PARE: Mask Parity Error Interrupt

0: PARE Interrupt is Disabled.

176 **AT91SAM7A2** 

1: PARE Interrupt is Enabled.

## • TIMEOUT: Mask Time Out Interrupt

0: TIMEOUT Interrupt is Disabled.

1: TIMEOUT Interrupt is Enabled.

## • TXEMPTY: Mask TXEMPTY Interrupt

0: TXEMPTY Interrupt is Disabled.

1: TXEMPTY Interrupt is Enabled.

#### • IDLE: Mask IDLE Interrupt

0: IDLE Interrupt is Disabled.

1: IDLE Interrupt is Enabled.

#### • SCK, TXD, RXD: PIO Interrupt Mask

These bits show which pins have interrupts enabled. They are updated by writing to US\_IER or US\_IDR.

0: Interrupt is not enabled on the corresponding input pin.

1: Interrupts is enabled on the corresponding input pin.

## • ENDHEADER: Enable ENDHEADER Interrupt

0: No effect.

1: Enables ENDHEADER interrupt.

#### • ENDMESS: Enable ENDMESS Interrupt

0: No effect.

1: Enables ENDMESS Interrupt.

## • NOTRESP: Enable NOTRESP Interrupt

0: No effect.

1: Enables NOTRESP Interrupt.

## • BITERROR: Enable BITERRROR Interrupt

0: No effect.

1: Enables BITERROR Interrupt.

### • IPERROR: Enable IPERROR Interrupt

- 0: No effect.
- 1: Enables IPERROR Interrupt.

## CHECKSUM: Enable CHECKSUM Interrupt

- 0: No effect.
- 1: Enables CHECKSUM Interrupt.
- WAKEUP: Enable WAKEUP Interrupt

0: No effect.





1: Enables WAKEUP Interrupt.

## 19.39 USART Receiver Holding Register

Name: Access: Base Address:	US_RH Read-or 0x080						
31	30	29	28	27	26	25	24
-	_	—	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	—	-	—	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	—	-	-	RXCHR[8]
7	6	5	4 BXCh	3 IR[7:0]	2	1	0
			ПЛОГ	n ([ <i>1</i> .0]			

## • RXCHR[8:0]: Received Character

Last character received if RXRDY is set. When the number of data bits is less than 9 bits, the bits are right aligned.

- Notes: 1. When reading this register, RXRDY bit is clear in the US\_RHR.
  - 2. When debugging, to avoid clearing RXRDY bit, users should use ghost registers (see "Ghost Registers" on page 9).

## 19.40 USART Transmit Holding Register

Name: Access: Base Address:	US_THR Write-onl 0x084	l					
31	30	29	28	27	26	25	24
-	_	_	-	-	-	_	-
23	22	21	20	19	18	17	16
-	-	-	_	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	TXCHR[8]
7	6	5	4	3	2	1	0
			TXCH	IR[7:0]			

## • TXCHR[8:0]: Character to be Transmitted

Next character to be transmitted after the current character if TXRDY is not set. When the number of data bits is less than 9 bits, the bits are right aligned.

## 19.41 USART Baud Rate Generator Register

Name: Access: Base Address:	US_BR Read/W 0x088	GR							
31	30	29	28	27	26	25	24		
-	-	-	-	-	-	-	-		
23	22	21	20	19	18	17	16		
-	-	_	-	-	-	—	-		
15	14	13	12	11	10	9	8		
CD[15:8]									
7	6	5	4	3	2	1	0		
	CD[7:0]								

## • CD[15:0]: Clock divisor

This register has no effect if synchronous mode is selected with an external clock. <sup>(1)</sup> <sup>(2)</sup>

CD[15:0]	Action
0	Disables clock
1	Clock divider bypassed
2 to 65535	Baud Rate (Asynchronous Mode) = Selected clock/(16 x CD) Baud Rate (Synchronous Mode) = Selected clock/CD

Notes: 1. In synchronous mode, the value programmed must be even to ensure a 50:50 mark/space ratio.

2. CD = 1 must not be used when internal clock (CORECLK) is selected (i.e USCLKS[1:0] = 00b).





## 19.42 USART Receiver Time Out Register

Name: Access: Base Address:	US_RT Read/W 0x08C		-							
31	30	29	28	27	26	25	24			
—	-	—	—	-	-	—	-			
23	22	21	20	19	18	17	16			
-	-	-	-	-	-	-	-			
15	14	13	12	11	10	9	8			
			TO	15:8]						
7	6	5	4	3	2	1	0			
	TO[7:0]									

## • TO[15:0]: Time Out Value

O[7:0]	Action
0	Disables the RX Time-out function.
1-65535	The Time-out counter is loaded with TO[15:0] when the Start Time-out Command is given or when each new data character is received (after reception has started).

In asynchronous mode:

Time out duration =  $TO[15:0] \times Bit$  period.

In synchronous mode:

Time out duration =  $TO[15:0] \times 4 \times Bit$  period.

When the receiver is disabled by setting the RXDIS bit in the US\_CR register, the time out is stopped. If the receiver is reenabled by setting the RXEN bit in the US\_CR register, the time out restarts where it left off (i.e. it is not reset).

# 19.43 USART Transmit Time Guard Register

Name: Access: Base Address:	US_TT Read/W 0x090		5				
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	—	_	_	_	_	_	_
15	14	13	12	. 11	10	9	8
-	—	-	-	-	—	-	-
7	6	5	4	3	2	1	0
			TO	[7:0]			

# • TG[7:0]: Time Guard Value

TG[7:0]	Action
0	Disables the TX Time Guard function.
1-255	TXD is inactive high after the transmission of each character for the Time Guard duration.

Time Guard duration =  $TG[7:0] \times Bit$  period.

# 19.44 USART LIN Identifier Register

Name: Access: Base Address	US_LIF Read/W : 0x094						
31	30	29	28	27	26	25	24
-	—	-	-	—	—	—	—
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_			IDENTIF	FIER[5:0]		

### • IDENTIFIER [5:0]: LIN 'S IDENTIFIER

Indicates the LIN 's Identifier Content to be transmitted on the next "HEADER MESSAGE".





# 19.45 USART Data Field Write 0 Register

Name: Access: Base Address:	US_DF\ Read/W 0x098		-								
31	30	29	28	27	26	25	24				
	DATA3[7:0]										
23	22	21	20	19	18	17	16				
			DATA	2[7:0]							
15	14	13	12	11	10	9	8				
DATA1[7:0]											
7	6	5	4	3	2	1	0				
			DATA	.0[7:0]							

### • DATAx [7:0]: LIN 's BYTE FIELD to be Transmitted

Data to be transmitted on the "RESPONSE 's LIN MESSAGE" when STRESP is set on the Controller Register.

19.40 USARI		a write i Reg	gister				
Name: Access: Base Address:	US_DF\ Read/W 0x09C						
31	30	29	28	27	26	25	24
			DATA	7[7:0]			
23	22	21	20	19	18	17	16
			DATA	6[7:0]			
15	14	13	12	11	10	9	8
			DATA	.5[7:0]			
7	6	5	4	3	2	1	0
			DATA	4[7:0]			

# 19.46 USART Data Field Write 1 Register

### • DATAx [7:0]: LIN 's BYTE FIELD to be Transmitted

Data to be transmitted on the "RESPONSE 's LIN MESSAGE" when STRESP is set on the Controller Register.

# 19.47 USART Data Field Read 0 Register

Name: Access: Base Address:	US_DFR Read/Wr 0x0A0	R0	<b>.</b>						
31	30	29	28	27	26	25	24		
			DATA	3[7:0]					
23	22	21	20	19	18	17	16		
			DATA	2[7:0]					
15	14	13	12	11	10	9	8		
DATA1[7:0]									
7	6	5	4	3	2	1	0		
			DATA	0[7:0]					

• DATAx [7:0]: LIN 's BYTE FIELD to be Received

Data read on the latest "RESPONSE 's LIN MESSAGE".

# 19.48 USART Data Field Read 1 Register

Name: Access: Base Address:	US_DFRI Read/Wri 0x0A4		-							
31	30	29	28	27	26	25	24			
	DATA7[7:0]									
23	22	21	20	19	18	17	16			
			DATA	.6[7:0]						
15	14	13	12	11	10	9	8			
	DATA5[7:0]									
7	6	5	4	3	2	1	0			
			DATA	.4[7:0]						

### • DATAx [7:0]: LIN 's BYTE FIELD to be Received

Data read on the latest "RESPONSE 's LIN MESSAGE".





Name: Access: Base Address:	US_SBL Read/W 0x0A8						
31	30	29	28	27	26	25	24
-	-	-	-	_	_	_	-
23	22	21	20	19	18	17	16
-	-	_	-	-	-	_	-
15	14	13	12	11	10	9	8
-	-	_	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	—	SYNCH_BRK				

-1: 1 after reset

### 19.49 USART Sync Break Length Register

• SYNC\_BRK [4:0]: Sync Break Length

R: Read

Configures sync break field length.

W: Write

If SYNC\_BRK is less than 8, it remains in previous state.

-0: 0 after reset

-U: undefined after reset

# 20. Capture (CAPT)

### 20.1 Overview

The AT91SAM7A2 provides a Capture module that serves as a frame analyzer. It stores the duration period (high and low level) in the CAPTURE Data Register (CAPT\_DR); these durations are described as a number of counter cycles (CAPTCLK). The Capture peripheral provides data transfer with the PDC.

Capture can detect all frame variations (with an error of one CAPTCLK period maximum) if the input signal has a frequency less than CAPTCLK/2.

It is possible to choose among three modes of measurement:

- Duration between both edges (positive and negative).
- Duration between positive edges.
- Duration between negative edges.

It is important to check that the frame to be watched by the capture module is not too fast, otherwise the PDC can monopolize the ASB bus as the PDC, in this case, will often have something to read on the Capture Data Register (CAPT\_DR).

If an overrun occurs, it is possible to choose to overwrite the data stored in the Data Register (CAPT\_DR) or to stop the data acquisition through the mode register (CAPT\_MR).

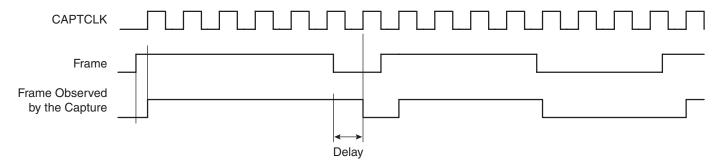
After a read of the data register, the DATACAPT bit (in the CAPT\_SR register) is automatically cleared (i.e. set to a logical 0).

It is recommended to disable the capture module after every modification of the CAPT\_MR register, otherwise the first measurement can be false.

When Capture is disabled, the capture counter is reset.

### 20.2 Example of Use

Figure 20-1. Capture Pin Resynchronization



#### 20.2.1 Example Use of CAPT

Capture of one period of a signal with a core clock of 30 MHz using the PDC and the associated interrupt. The signal frequency should be below 7.5 MHz and above 228 Hz. The period measurement will be between two positive edges.





### 20.2.2 Configuration

- Enable the clock on CAPT peripheral by writing the CAP bit in CAPT\_ECR.
- Do a software reset of the capture peripheral to be in a known state by writing the SWRST bit in CAPT\_CR.
- Configuration of CAPT\_MR: Select the CAPT clock to be 15 MHz with the PRESCALAR field set to 0, OVERMODE is ignored here as one capture is done and ONESHOT is set. The measurement is described by MEASMODE and the value 0x2 is chosen for this field (between two positive edges).
- Enable the CAPT by writing the CAPEN bit in CAPT\_CR.
- Configuration of CAPT\_IER: An interrupt should be generated at the end of the capture. This
  is done by writing PDCEND. When the PDC finishes the capture, an interrupt will be
  generated. GIC must be configured.
- Configure PDC for CAPT module and set it to 1 capture of 16 bits.
- If the status bit CAPENS is set in CAPT\_SR, start the capture by writing the STARTCAPT bit in CAPT\_CR, otherwise users should wait the CAPENS flag (consequence of the CAPEN bit in CAPT\_CR).

### 20.2.3 Interrupt Handling

- IRQ Entry and call C function.
- Read CAPT \_SR and verify the source of the interrupt.
- Clear the corresponding interrupt at peripheral level by writing in CAPT\_CSR.
- Interrupt treatment: Read the duration between the two positive edges in the received memory space programmed in the PDC. The duration is expressed by the number of capture clocks (duration/15 MHz).
- IRQ Exit.

### 20.3 Capture Limits

To prevent the capture from missing frame edges it is important to check that:

Frame frequency = CAPTCLK/2

Moreover, the capture detects each frame edge with a delay equal to the CAPTCLK period.

The CAPTCLK frequency can take a value between 15 MHz and 916 Hz (if 30 MHz is the CORECLK frequency).

#### Table 20-1. Capture Delay and Error (Example)

CAPTCLK Frequency	Fattest Observable Frame	Delay Max for Edge Detection	Error Max in Level Duration Value	
F	F/2 if F< CORECLK frequency F/4 if F= CORECLK frequency	1/F	2/F	
30 MHz	7.5 MHz	33 ns	66 ns	
916 Hz	458 Hz	1.1 ms	2.2 ms	

### 20.4 Power Management

Capture is provided with a power management block allowing optimization of power consumption. See "Power Management Block" on page 25.

# 20.5 Capture (CAPT) Memory Map

Base Address CAPT0: 0xFFFDC000

Base Address CAPT1: 0xFFFE0000

Table 20-2. CAPT Memory Map

Offset	Register	Name	Access	Reset State
0x000 - 0x04C	Reserved	-	-	_
0x050	Enable Clock Register	CAP_ECR	Write-only	_
0x054	Disable Clock Register	CAP_DCR	Write-only	_
0x058	Power Management Status Register	CAP_PMSR	Read-only	0x0000000
0x05C	Reserved	_	_	-
0x060	Control Register	CAP_CR	Write-only	-
0x064	Mode Register	CAP_MR	Read/Write	0x0000000
0x068	Reserved	_	_	-
0x06C	Clear Status Register	CAP_CSR	Write-only	-
0x070	Status Register	CAP_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	CAP_IER	Write-only	-
0x078	Interrupt Disable Register	CAP_IDR	Write-only	-
0x07C	Interrupt Mask Register	CAP_IMR	Read-only	0x0000000
0x080	Data Register	CAP_DR	Read-only	0x0000000





# 20.6 CAPTURE Enable Clock Register

Name: Access: Base Address	Write-or	CAPT_ECR Write-only 0x050									
31	30	29	28	27	26	25	24				
-	-	-	-	-	-	-	-				
23	22	21	20	19	18	17	16				
-	-	-	-	-	-	-	-				
15	14	13	12	11	10	9	8				
-	-	—	-	-	-	-	-				
7	6	5	4	3	2	1	0				
-	_	_	-	-	-	CAP	-				

### 20.7 CAPTURE Disable Clock Register

Name: Access: Base Address:	CAPT_I Write-or 0x054						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	-	_	-	CAP	-

# 20.8 CAPTURE Power Management Status Register

Name: Access: Base Address:	CAPT_F Read-or 0x058			-			
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	_	-
15	14	13	12	11	10	9	8
-	_	_	_	-	—	-	-
7	6	5	4	3	2	1	0
-	_	-	-	-	—	CAP	-

### • CAP: CAPTURE Clock

0: Capture clock disabled.

1: Capture clock enabled.

Note: The CAPT\_PMSR register is not reset by a software reset.





# 20.9 CAPTURE Control Register

Name: Access: Base Address:	CAPT_C Write-on 0x060						
31	30	29	28	27	26	25	24
-	_	_	-	-	_	-	_
W	W	W	W	W	W	W	W
23	22	21	20	19	18	17	16
-	_	_	-	-	_	-	_
W	W	W	W	W	W	W	W
15	14	13	12	11	10	9	8
-	_	_	-	-	_	-	_
W	W	W	W	W	W	W	W
7	6	5	4	3	2	1	0
-	-	-	-	STARTCAP	CAPDIS	CAPEN	SWRST

### SWRST: Capture Software Reset

0: No effect.

1: Resets the Capture.

A software reset of the Capture is performed. It resets all the registers (except CAPT\_PMSR).

### • CAPEN: Capture Enable

0: No effect.

1: Enables the Capture.

### • CAPDIS: Capture Disable

0: No effect.

### 1: Disables the Capture.

If both CAPEN and CAPDIS are equal to one when the control register is written, the CAPTURE will be disabled.

#### • STARTCAP: Start Capture

0: No effect.

1: The Capture starts a new Capture.

# 20.10 CAPTURE Mode Register

Name: Access: Base Address	CAPT_N Read/W : 0x064	٨R					
31	30	29	28	27	26	25	24
-	-	-	-	-	_	-	-
23	22	21	20	19	18	17	16
_	_	_	_	_	_	_	-
15	14	13	12	11	10	9	8
_	_	_	_	_	_	_	-
7	6	5	4	3	2	1	0
ONESHOT	OVERMODE	MEASM	IODE[1:0]		PRESCI	LAR[3:0]	

### • PRESCALR[3:0]: Counter CLock Prescalar

CAPTCLK Frequency = 2<sup>PRESCALAR + 1</sup> MEASMODE[1:0]: Measurement Mode

MEASM	ODE[1:0]	Measure
0	Х	Measure between each edge (positive and negative).
1	0	Measure between positive edges.
0	1	Measure between negative edges.

### OVERMODE: Overrun Mode

0: In case of an overrun the capture stops writing on the Data Register. If DATACAPT bit is enabled and the CAPTURE module receives new data, the data register will not be refreshed.

1: In case of an overrun the capture does not stop writing on the Data Register.

#### • ONESHOT: One Shot

0: The capture still captures a frame variation.

1: The module captures a frame variation and stops. To ask for another capture, the STARTCAPT bit has to be set at 1 in CAPT\_CR.





# 20.11 CAPTURE Clear Status Register

Name: Access: Base Address:	CAPT_0 Write-or 0x06C						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	_	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	Ι	-
7	6	5	4	3	2	1	0
-	-	_	-	-	CAPDIS	CAPEN	SWRST

### Clear PDCEND Interrupt

0: No effect.

1: Clears the PDCEND interrupt.

# • OVERRUN: Clear Overrun Interrupt

0: No effect.

1: Clears the OVERRUN interrupt.

# OVERFLOW: Clear Overflow Interrupt

0: No effect.

1: Clears the OVERFLOW interrupt.

# 20.12 CAPTURE Status Register

Name: Access: Base Address	CAPT_S Read-o : 0x070						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	-	-	-	_	—	-	-
15	14	13	12	11	10	9	8
_	-	-	-	_	_	—	CAPENS
7	6	5	4	3	2	1	0
_	—	—	-	DATACAPT	OVERFLOW	OVERRUN	PDCEND

### • PDCEND: PDC end

0: No effect.

1: The PDC has finished the data transfer.

### • OVERRUN: Overrun

0: No effect.

1: An overrun has occurred.

Overrun indicates a valid data was not read when an overwrite occurred.

### • OVERFLOW: Overflow

- 0: No effect.
- 1: An overflow has occurred.

Overflow indicates that the counter of the duration has been saturated.

### • DATACAPT: Data Captured

0: No effect.

1: Data in CAP\_DR has to be read.

This bit is cleared by reading the CAP\_DR register.

### • CAPENS: Capture Enable Status

0: Capture is disabled.

1: Capture is enabled.





# 20.13 CAPTURE Interrupt Enable Register

Name: Access: Base Address:	CAPT_I Write-or 0x074						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	DATACAPT	OVERFLOW	OVERRUN	PDCEND

# 20.14 CAPTURE Interrupt Disable Register

Name: Access: Base Address:	CAPT_I Write-or 0x078		-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
—	_	—	—	DATACAPT	OVERFLOW	OVERRUN	PDCEND

# 20.15 CAPTURE Interrupt Mask Register

Name: Access: Base Address:	CAPT_I Write-or 0x07C	MR	5				
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	. 11	10	9	8
-	_	-	-	-	_	_	_
7	6	5	4	3	2	1	0
-	—	_	_	DATACAPT	OVERFLOW	OVERRUN	PDCEND

### PDCEND: PDC End Interrupt Mask

0: PDCEND Interrupt is disabled.

1: PDCEND Interrupt is enabled.

### • OVERRUN: Overrun Interrupt Mask

0: OVERRUN Interrupt is disabled.

1: OVERRUN interrupt is enabled.

#### • OVERFLOW: Overflow Interrupt Mask

0: OVERFLOW Interrupt is disabled.

1: OVERFLOW interrupt is enabled.

### • DATACAPT: Data Capture Interrupt Mask

0: DATACAPT interrupt is disabled.

1: DATACAPT interrupt is enabled.





# 20.16 CAPTURE Data Register

Name: Access: Base Address:	CAPT_ Read/W						
31	30	29	28	27	26	25	24
-	_	-	-	_	-	_	-
23	22	21	20	19	18	17	16
-	—	-	-	-	_	-	-
15	14	13	12	11	10	9	8
LEVEL		DURATION[14:8]					
7	6	5	4	3	2	1	0
			DURAT	ION[7:0]			

Notes: 1. When reading this register, the DATACAPT bit is clear in CAPT\_SR

2. When debugging, to avoid clearing the DATACAPT bit, users should use ghost registers (see "Ghost Registers" on page 9).

### • DURATION[14:0]: Capture Duration

Numbers of CAPTCLK cycles during which the output is at the level indicated by the LEVEL bit, or during a frame period, depending of MEASMODE in CAP\_MR.

### • LEVEL: Level measured

0: the duration concerns a low level.

1: the duration concerns a high level.

Note that if the MEASMODE[1:0] (CAPT\_MR) equals 1X, the LEVEL bit is used as a duration bit.

# 21. Simple Timer (ST)

### 21.1 Overview

The AT91SAM7A2 microcontroller includes two 16-bit Simple Timers (ST0 and ST1) with two channels per simple timer.

Each simple timer channel provides basic functions for timing calculation including two cascaded dividers and a 16-bit counter.

The prescalar defines the clock frequency of the channel counter.

For each channel it is possible to select the divider clock between the core clock (CORECLK) and the low frequency clock (LFCLK).

The 16-bit counter starts down-counting when a value different from zero is loaded. An interrupt is generated when the counter reaches 0x0000.

When a value is loaded in the LOAD field of the STx\_CTz register and the channel is started, the counter starts down-counting at channel clock frequency until the counter reaches zero. The delay between the load and the interrupt is the counter value multiplied by the clock period.

The counter value is the value loaded in the counter. The clock period is the period of the channel clock (divided by the prescalar).

The precision is one clock period (from 0 to 1 channel clock period lost).

Note: When enabling or disabling the Simple Timer, software must wait for enabled or disabled interrupt (or status) to be sure that the counter is really enabled or disabled (it is asynchronously clocked).

It is not possible to change the Channel x Counter Register content when the Simple Timer is enabled.

If a channel is disabled before it reaches the end of the down-counting, the counter value is held. If no write has occurred on the Counter Register before it is re-enabled, the down-counting restarts from the latest counter value.

When the core clock (CORECLK) is selected on the Prescalar Register, the clock is divided twice to obtain the counter clock: firstly by a divider driven by the SYSCAL bits of the Prescalar Register and finally by a divider driven by the prescalar bits of the Prescalar Register.

However, if the low frequency clock (LFCLK) is selected, the clock is divided just once to obtain the counter clock by a divider driven by the prescalar bits of the Prescalar Register.

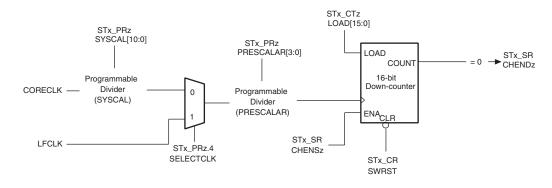
The simple timer also integrates an automatic reload function if the AUTOREL bit is set on the corresponding STx\_PRz register. When the counter reaches 0x0000, it is automatically reloaded with the value written in the LOAD[15:0] bits of the STx\_CTz registers (x for ST0 or ST1 and z for channel 0 or channel 1).

The current value of the counter can be read in the corresponding ST\_CCVx register.





### 21.2 Block Diagram



### Figure 21-1. Simple Timer Block Diagram

### 21.3 Peripheral Structure

#### 21.3.1 Interrupt Description

For each channel there are three types of interrupt:

- A CHDISz interrupt which indicates that the channel has been disabled.
- A CHLOADz interrupt which indicates that the channel has started to down count and loaded the data.
- A CHENDz interrupt which indicates that the channel has reach the end of the downcounting. This interrupt rises 3 CORECLK cycles after the down counter reaches the value 0x0000.

where z = is the channel number.

#### 21.4 Power Management

Each Simple Timer (ST0 and ST1) is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25). The PMC controls the clock inputs and the clock divider blocks.

When the timer is stopped, the clock is immediately stopped. When the clock is re-enabled, the timer resumes action where it left off.

### 21.5 Example of Use

This section gives an example of the use of the Simple Timer, explaining how to generate a tick (usually used in RTOS) of 10 ms with auto-reload and an interrupt.

Core clock is 30 MHz.

#### 21.5.1 Configuration

- Enable the clock on the Simple Timer (ST) by writing bit ST in ST\_ECR.
- Do a software reset of the peripheral so that it is in a known state by writing SWRST bit in ST\_CR. Wait about four LFCLK period for the circuitry to become stable.
- Configuration of ST\_PRX: A tick should occur every 10 ms, thus giving a frequency of 100 Hz; e.g., the counter frequency is 15 kHz and is decremented by 150. To produce a counter frequency of 15 kHz, calculate

# 198 **AT91SAM7A2**

Divider clock =  $\frac{\text{CORECLK}}{2 \text{ SYSCAL}} + 1$ 

with SYSCAL = 0 and SELCETCLK = 0, divider clock = 15 MHz. Then with a PRESCALAR of 10,

Counter clock =  $\frac{\text{dividerclock}}{2^{PRESCALAR}}$ 

or about 14648 Hz. Users can adjust using SYSCAL and PRESCALAR to match the nearest wanted value. To enable the auto-reload functionality, the bit AUTOREL must be set.

- Configuration of ST\_CTX: As shown previously, to obtain a 10 ms tick timer, the counter load value must be 150. This is the value to be written in the LOAD field. An interrupt can be generated when this value is really loaded. Users can verify that CHLDX bit in ST\_SR is set before going to the next step.
- Configuration of ST\_IER: Generation occurs when the counter reaches 0. This is done by writing the CHENDX bit. GIC must be configured.
- Start the simple timer by writing the CHENX bit in ST\_CR. Users can verify that CHENSX bit in ST\_SR is set.

### 21.5.2 Interrupt Handling

- IRQ Entry and call C function.
- Read ST\_SR and verify the source of the interrupt.
- Clear the corresponding interrupt at peripheral level by writing in the ST\_CSR.
- Interrupt treatment.
- IRQ Exit.





# 21.6 Simple Timer (ST) Memory Map

Base Address ST0: 0xFFFE4000

Base Address ST1: 0xFFFE8000

Table 21-1. ST Memory Map

Offset	Register	Name	Access	Reset State
0x000 - 0x04C	Reserved	-	-	-
0x050	Enable Clock Register	ST_ECR	Write-only	-
0x054	Disable Clock Register	ST_DCR	Write-only	-
0x058	Power Management Status Register	ST_PMSR	Read-only	0x0000000
0x05C	Reserved	-	_	_
0x060	Control Register	ST_CR	Write-only	_
0x064	Reserved	-	-	-
0x068	Reserved	-	-	_
0x06C	Clear Status Register	ST_CSR	Write-only	_
0x070	Status Register	ST_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	ST_IER	Write-only	_
0x078	Interrupt Disable Register	ST_IDR	Write-only	_
0x07C	Interrupt Mask Register	ST_IMR	Read-only	0x0000000
0x080	Channel 0 Prescalar	ST_PR0	Read/Write	0x00000000
0x084	Channel 0 Counter	ST_CT0	Read/Write	0x00000000
0x088	Channel 1 Prescalar	ST_PR1	Read/Write	0x0000000
0x08C	Channel 1 Counter	ST_CT1	Read/Write	0x00000000
0x200	Current Counter Value 0	ST_CCV0	Read	0x00000000
0x204	Current Counter Value 1	ST_CCV1	Read	0x00000000

# 21.7 ST Enable Clock Register

Name:	ST_ECR
Access:	Write-only
Base Address:	0x050

# 21.8 ST Disable Clock Register

Name:	ST_DCR
Access:	Write-only
Base Address:	0x054

31	30	29	28	27	26	25	24
_	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	-	-	—	—	—	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	—	—	—	—	—	ST	—

### • ST: Simple Timer Clock Status

0: Simple Timer clock disabled.

1: Simple Timer clock enabled.





# 21.9 ST Power Management Status Register

Name: Access: Base Address:	ST_PM Read-or 0x058		-				
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	_	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	—	-	-	-	—	-
7	6	5	4	3	2	1	0
_	_	_	-	-	-	ST	-

# • ST: Simple Timer Clock Status

0: Simple Timer clock disabled.

1: Simple Timer clock enabled.

Note: The ST\_PMSR register is not reset by a software reset.

# 21.10 ST Control Register

Name: Access: Base Address:	ST_CR Write-or						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	_	_	_	-
15	14	13	12	11	10	9	8
-	_	-	-	-	_	_	_
7	6	5	4	3	2	1	0
-	_	-	CHDIS1	CHEN1	CHDIS0	CHEN0	SWRST

### • SWRST: ST Software Reset

0: No effect.

1: Resets the Simple Timer.

A software reset of the ST is performed. It resets all the registers.

### • CHENX: Simple Timer Channel Enable

0: No effect.

1: Enables the Simple Timer Channel X.

### • CHDISX: Simple Timer Channel Disable

0: No effect.

1: Disables the Simple Timer Channel X.

If both CHENX and CHDISX are equal to one when the control register is written, the Simple Timer Channel X will be disabled.





# 21.11 ST Clear Status Register

Name: Access: Base Address:	ST_CSF Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	. 11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	-	CHDIS1	CHEN1	CHDIS0	CHEN0	SWRST

### • CHENDX: Clear Channel End Interrupt

0: No effect.

1: Clears CHENDX interrupt.

### CHDISX: Clear Channel Disable Interrupt

0: No effect.

1: Clears CHDISX interrupt.

### • CHLDX: Clear Channel Load Interrupt

0: No effect.

1: Clears CHLDX interrupt.

### 21.12 ST Status Register

Name: Access: Base Address:	ST_SR Read-or 0x070						
31	30	29	28	27	26	25	24
-	_	_	_	_	_	CHENS1	CHENS0
23	22	21	20	19	18	17	16
15	14	13	12	11	10	9	8
_	_	_	-	_	_	_	-
7	6	5	4	3	2	1	0
_	_	CHLD1	CHDIS1	CHEND1	CHLD0	CHDIS0	CHEND0

### • CHENDX: Channel End Status

0: No end of counting on Channel X.

1: End of counting on Channel X.

### • CHDISX: Channel Disable Status

0: No effect.

1: Channel X divider has been reset.

CHDISX indicates that the divider module has been reset (a delay due to asynchronous clock is introduced).

### • CHLDX: Channel Load Status

0: No effect.

1: Channel X down-counter is loaded.

Note: CHLDX indicates also that the counter (divider) has been enabled (a delay due to asynchronous clock is introduced).

#### • CHENSX: Channel Enable Status

0: Channel X is disabled.

1: Channel X is enabled.





# 21.13 ST Interrupt Enable Register

Name: Access: Base Address:	ST_IER Write-or 0x074						
31	30	29	28	27	26	25	24
-	_	_	-	_	—	_	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	CHLD1	CHDIS1	CHEND1	CHLD0	CHDIS0	CHEND0

### 21.14 ST Interrupt Disable Register

Name: Access: Base Address:	ST_IDR Write-or	ST_IDR Write-only 0x078								
31	30	29	28	27	26	25	24			
-	_	-	-	-	-	-	-			
23	22	21	20	19	18	17	16			
-	_	_	-	_	_	_	-			
15	14	13	12	11	10	9	8			
-	_	-	-	-	-	-	-			
7	6	5	4	3	2	1	0			
-	_	CHLD1	CHDIS1	CHEND1	CHLD0	CHDIS0	CHEND0			

# 21.15 ST Interrupt Mask Register

Name: Access: Base Address:	ST_IMR Read-or						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	_	_	_	-	_	_	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	CHLD1	CHDIS1	CHEND1	CHLD0	CHDIS0	CHEND0

### • CHENDX: Channel End Interrupt Mask

0: Channel X end interrupt is disabled.

1: Channel X end interrupt is enabled.

### CHDISX: Channel Disable Interrupt Mask

0: Channel X disable interrupt is disabled.

1: Channel X disable interrupt is enabled.

#### CHLDX: Channel Load Interrupt Mask

0: Channel X load interrupt is disabled.

1: Channel X load interrupt is enabled.





		soulai negi	5101						
Name:	ST_PR0	I							
Access:	Read/Write								
Base Address:	0x080								
31	30	29	28	27	26	25	24		
-	-	-	-	-	_	—	-		
23	22	21	20	19	18	17	16		
-	-	_	-			SYSCAL			
15	14	13	12	11	10	9	8		
			SYSC	CAL					
7	6	5	4	3	2	1	0		
-	_	AUTOREL	SELECTCLK		PRESC	CALAR			

# 21.16 ST Channel 0 Prescalar Register

### • PRESCALAR: Channel 0 Prescalar

Channel 0 counter\_clock\_frequency = divider\_clock\_0/  $2^{PRESCALAR}$ 

### • SELECTCLK: Select Clock

0: The divider\_clock\_0 is generated by divider driven by the SYSCAL bits.

1: The divider\_clock\_0 is connected to the low frequency clock.

### • AUTOREL: Auto Reload

0: The counter is not automatically reloaded with the COUNTER[15:0] value when it reaches 0x0000.

1: The counter is automatically reloaded with the COUNTER[15:0] value when it reaches 0x0000.

### • SYSCAL: System Clock Prescalar Value

This prescalar is used to divide the CORECLK when the system clock is selected (SELECTCLK = 0).

 $livider_clock_0 = CORECLK / (2 \times (SYSCAL + 1))$ 

When modifying the SYSCAL value, if the new value is smaller than the previous one, the counter may be locked up during at max 2048 core clock cycles + 1 cycle of counter clock before starting to decrement. Once this delay is passed, the counter decrements properly.

A write in this register can only be done if CHENS0 = 0 (i.e., channel 0 is disabled).

Name: Access: Base Address:	ST_CTC Read/W 0x084	)					
31	30	29	28	27	26	25	24
-	—	-	—	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	_	-	_	_	-
15	14	13	12	11	10	9	8
			LC	DAD			
7	6	5	4	3	2	1	0
			LC	DAD			

# 21.17 ST Channel 0 Counter Register

### • LOAD: Channel 0 Preload Value

Gives the instruction to the timer; i.e., from which counter value the Simple Timer has to decrement.





Name: Access: Base Address:	ST_PR1 Read/Wr 0x088						
31	30	29	28	27	26	25	24
-	—	_	-	_	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	_		SYSCAL	
15	14	13	12	11	10	9	8
			SYS	CAL			
7	6	5	4	3	2	1	0
	_	AUTOREL	SELECTCLK		PRES	CALAR	

# 21.18 ST Channel 1 Prescalar Register

### • PRESCALAR: Channel 1 Prescalar

Channel 1 counter\_clock\_frequency = divider\_clock\_0/  $2^{PRESCALAR}$ 

### • SELECTCLK: Select Clock

- 0: The divider\_clock\_0 is generated by divider driven by the SYSCAL bits.
- 1: The divider\_clock\_0 is connected to the low frequency clock.

### • AUTOREL: Auto Reload

- 0: The counter is not automatically reloaded with the COUNTER[15:0] value when it reaches 0x0000.
- 1: The counter is automatically reloaded with the COUNTER[15:0] value when it reaches 0x0000.

### • SYSCAL: System Clock Prescalar Value

This prescalar is used to divide the CORECLK when the system clock is selected (SELECTCLK = 0).

 $livider_clock_0 = CORECLK / (2 \times (SYSCAL + 1))$ 

A write in this register can only be done if CHENS0 = 0 (i.e. channel 0 is disabled).

Name:	ST_CT1								
Access:	Read/W	/rite							
Base Address:	0x08C								
31	30	29	28	27	26	25	24		
-	-	-	_	-	_	-	-		
23	22	21	20	19	18	17	16		
-	—	-	-	-	-	-	-		
15	14	13	12	11	10	9	8		
LOAD									
7	6	5	4	3	2	1	0		
			LC	DAD					

# 21.19 ST Channel 1 Counter Register

### • LOAD: Channel 1 Preload Value

Gives the instruction to the timer; i.e., from which counter value the Simple Timer has to decrement.





Name:	ST_CC\		logicite:						
Access: Base Address:	Read-or 0x0200								
31	30	29	28	27	26	25	24		
-	_	_	-	_	-	_	-		
23	22	21	20	19	18	17	16		
-	_	-	-	-	-				
15	14	13	12	11	10	9	8		
COUNT									
7	6	5	4	3	2	1	0		
			CO	UNT					

# 21.20 ST Current Counter Value 0 Register

### • COUNT[15:0]: Current Counter Value 0 Register

This register gives the current value of the Channel 0 down counter. In case a ST register is accessed concurrently to a rising edge of the counter clock, this register is not updated, and the value read in this register is the previous value. This is made to prevent asynchronous reading error.

Name: Access: Base Address:	ST_CC\ Read-or 0x0204		-					
31	30	29	28	27	26	25	24	
-	_	_	-	_	-	—	-	
23	22	21	20	19	18	17	16	
-	-	_	-	-	-	_	-	
15	14	13	12	11	10	9	8	
COUNT								
7	6	5	4	3	2	1	0	
			CO	UNT				

# 21.21 ST Current Counter Value 1 Register

### • COUNT[15:0]: Current Counter Value 1 Register

This register gives the current value of the Channel 1 down counter. this register is the previous value. This is made to prevent asynchronous reading error.

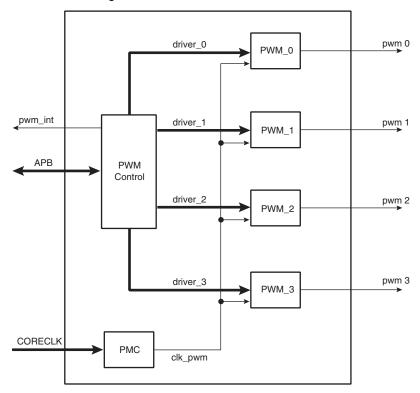
# 22. Pulse Width Modulator (PWM)

### 22.1 Overview

The AT91SAM7A2 includes a four-channel PWM that generates pulses. The width and the frequency of the pulses can be controlled independently on each channel, thus making it possible to generate four different waveforms at the same time.

### 22.2 Block Diagram

#### Figure 22-1. PWM Block Diagram



### 22.3 Pin Description

There are four output pins: PWM\_0, PWM\_1, PWM\_2, PWM\_3. They provide four digital signals that can be used to control, for example, a voltage level.

### 22.4 PWM Parameters

There are three parameters for each PWM: counter frequency, delay and pulse width.

A fourth parameter allows users to change polarity of the pulse and of the delay levels. By default, the pulse is at a logical 1 level and the delay is at a logical 0 level.

### 22.4.1 Counter Frequency

The PWM module is a counter. It counts delay width cycles, setting the PWMX output inactive, then it counts pulse width cycles, setting the PWMX output active. This operation is repeated until the PWM channel is stopped.

The user can control the frequency of this counter with a prescalar.





#### 22.4.2 Delay Width

Delay width is the number of counter cycles during which the output is inactive. PWM starts with a delay when enabled.

#### 22.4.3 Pulse Width

Pulse width is the number of counter cycles during which the output is active.

#### 22.4.4 PWM Output Level When Disabled

When the PWM module goes from enable to disable state, its outputs are set to the delay level defined in the mode register by the PLX bits.

If PLX is set to 0, when the PWM is disabled, the output will drive a logical 1.

If PLX is set to 1, when the PWM is disabled, the output will drive a logical 0.

### 22.5 Power Management

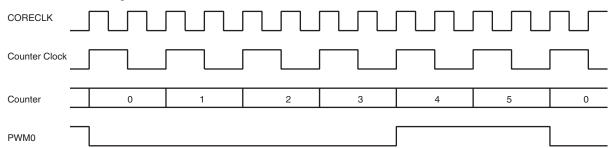
The PWM is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).

### 22.6 Example of Use

This section gives an example of use of the PWM, explaining how to generate a pulse with a frequency of 5 KHz with a duty cycle of 50%, core clock = 30 MHz.

If the counter frequency is set to CORECLK/2, delay width is equal to four cycles and pulse width to two cycles. See Figure 22-2.

#### Figure 22-2. Waveform Diagram



#### 22.6.1 Configuration

- Enable the clock on PWM peripheral by writing bit PWM in PWM\_ECR.
- Do a software reset of the PWM peripheral to be in a known state by writing bit SWRST in PWM\_CR.
- Configuration of PWM\_ MR: Use a PWM clock of 15 MHz with a PRESCAL of 0 and the pulse at a logical state 1. Users can choose one of the four PWMs available.
- Configuration of PWM\_ DLYX: This register indicates the number of PWM clocks needed to form the delay (level 0), e.g. 1500, that should result in half a period of 0.1 ms.
- Configuration of PWM\_ PULX: This register indicates the number of PWM clocks needed to form the pulse (level 1), e.g. 1500, that should result in half a period of 0.1ms.

• Start the selected PWM by writing the bit PWMENX in PWM\_CR. This should supply a pulse of 5 kHz on the selected PWM with a duty cycle of 50%.

### 22.7 Pulse Width Modulator (PWM) Memory Map

Base Address:0xFFFD0000

#### Table 22-1. PWM Memory Map

Offset	Register	Name	Access	Reset State
0x000 - 0x04C	Reserved	-	-	-
0x050	Enable Clock Register	PWM_ECR	Write-only	_
0x054	Disable Clock Register	PWM_DCR	Write-only	_
0x058	Power Management Status Register	PWM_PMSR	Read-only	0x0000000
0x05C	Reserved	-	-	_
0x060	Control Register	PWM_CR	Write-only	_
0x064	Mode Register	PWM_MR	Read/Write	0x10101010
0x068	Reserved	_	_	_
0x06C	Clear Status Register	PWM_CSR	Write-only	-
0x070	Status Register	PWM_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	PWM_IER	Write-only	_
0x078	Interrupt Disable Register	PWM_IDR	Write-only	_
0x07C	Interrupt Mask Register	PWM_IMR	Read-only	0x0000000
0x080	Delay Register 0	PWM_DLY_0	Read/Write	0x0000000
0x084	Pulse Register 0	PWM_PUL_0	Read/Write	0x0000000
0x088	Delay Register 1	PWM_DLY_1	Read/Write	0x0000000
0x08C	Pulse Register 1	PWM_PUL_1	Read/Write	0x0000000
0x090	Delay Register 2	PWM_DLY_2	Read/Write	0x0000000
0x094	Pulse Register 2	PWM_PUL_2	Read/Write	0x0000000
0x098	Delay Register 3	PWM_DLY_3	Read/Write	0x0000000
0x09C	Pulse Register 3	PWM_PUL_3	Read/Write	0x0000000





# 22.8 PWM Enable Clock Register

Name: Access: Base Address:	PWM_E Write-or 0x050						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	—	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	-	-	-	PWM	-

### • PWM: PWM Clock

0: PWM clock disabled.

1: PWM clock enabled.

The PWM\_PMSR register is not reset by software reset.

# 22.9 PWM Disable Clock Register

Name: Access: Base Address:	PWM_D Write-or 0x054						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	—	-	-	_	-	-
15	14	13	12	11	10	9	8
-	—	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	-	-	-	_	PWM	-

#### • PWM: PWM Clock

0: PWM clock disabled.

1: PWM clock enabled.

Note: The PWM\_PMSR register is not reset by software reset.

### 22.10 PWM Power Management Status Register

Name: Access: Base Address:	PWM_F Read-or 0x058						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	—	_	_	_	_	PWM	_

#### • PWM: PWM Clock

0: PWM clock disabled.

1: PWM clock enabled.

Note: The PWM\_PMSR register is not reset by software reset.





### 22.11 PWM Control Register

Name: Access: Base Address:	PWM_C Write-or 0x060	R					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	_	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	_	PWMDIS3
7	6	5	4	3	2	1	0
PWMEN3	PWMDIS2	PWMEN2	PWMDIS1	PWMEN1	PWMDIS0	PWMEN0	SWRST

#### • SWRST: PWM Software Reset

0: No effect

1: Resets the PWM.

A software-triggered hardware reset of the PWM is performed. It resets all the registers except the PWM\_PMSR register.

#### • PWMENX: PWM Enable Channel Number X

0: No effect.

1: Enables the PWM.

#### • PWMDISX: PWM Disable Channel Number X

0: No effect.

1: Disables the PWM.

If both PWMENX and PWMDISX are equal to one, the corresponding PWM channel is disabled.

### 22.12 PWM Mode Register

Name: Access: Base Address:	PWM_M Read/W 0x064	<i>I</i> R						
31	30	29	28	27	26	25	24	
-	-	-	PL3		PRES	SCAL3		
23	22	21	20	19	18	17	16	
-	-	-	PL2	PRESCAL2				
15	14	13	12	11	10	9	8	
-	_	_	PL1		PRES	SCAL1		
7	6	5	4	3	2	1	0	
			PL0	PRESCAL0				

#### • PRESCALX[3:0]: Counter Clock Prescalar for PWM Channel X

Counter Clock Frequency = CORECLK/  $2^{PRESCALAR1}$ 

Note: "X" indicates that the bit concerns the PWMX.

#### • PLX: Pulse Level for PWM Channel X

0: The pulses are at logic level 0. During the delay, the output is at logic level 1.

1: The pulses are at logic level 1. During the delay, the output is at logic level 0.

Note: When pulse level change, a pulse start or a pulse end is detected, and the corresponding bit is set in the status register (generating an interrupt if enabled).





### 22.13 PWM Clear Status Register

Name: Access: Base Address:	PWM_C Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	—
15	14	13	12	11	10	9	8
-	_	-	_	_	_	-	_
7	6	5	4	3	2	1	0
PEND3	PSTA3	PEND2	PSTA2	PEND1	PSTA1	PEND0	PSTA0

#### • PSTAX: Pulse Start Interrupt

0: No effect.

1: Clears the PSTA interrupt.

### • PENDX: Pulse End Interrupt

0: No effect.

1: Clears the PEND interrupt.

### 22.14 PWM Status Register

Name: Access: Base Address:	PWM_S Read-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	_	-	_	_	_
15	14	13	12	11	10	9	8
—	_	-	-	PWMENS3	PWMENS2	PWMENS1	PWMENS0
7	6	5	4	3	2	1	0
PEND3	PSTA3	PEND2	PSTA2	PEND1	PSTA1	PEND0	PSTA0

#### • PSTAX: Pulse Start

0: No pulse started since the last read of PWM\_SR.

1: A pulse started since the last read of PWM\_SR.

#### • PENDX: Pulse End

0: No pulse ended since the last read of PWM\_SR.

1: A pulse ended since the last read of PWM\_SR.

#### • PWMENSX: PWM Enable Status of Channel X

0: PWM is disabled.

1: PWM is enabled.





### 22.15 PWM Interrupt Enable Register

Name:PWM\_IERAccess:Write-onlyBase Address:0x074

### 22.16 PWM Interrupt Disable Register

Name: Access: Base Address:	PWM_II Write-or 0x078						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	-	_	_	-
7	6	5	4	3	2	1	0
PEND3	PSTA3	PEND2	PSTA2	PEND1	PSTA1	PEND0	PSTA0

#### • PSTAX: Pulse Start Interrupt

0: Pulse Start Interrupt is disabled.

1: Pulse Start Interrupt is enabled.

#### • PENDX: Pulse End Interrupt

0: Pulse End Interrupt is disabled.

1: Pulse End Interrupt is enabled.

### 22.17 PWM Interrupt Mask Register

Name: Access: Base Address:	Read-or	PWM_IMR Read-only 0x07C								
31	30	29	28	27	26	25	24			
-	_	-	-	-	-	-	-			
23	22	21	20	19	18	17	16			
-	_	_	_	_	_	_	_			
15	14	13	12	11	10	9	8			
-	_	_	-	-	-	-	-			
7	6	5	4	3	2	1	0			
PEND3	PSTA3	PEND2	PSTA2	PEND1	PSTA1	PEND0	PSTA0			

#### • PSTAX: Pulse Start Interrupt

0: Pulse Start Interrupt is disabled.

1: Pulse Start Interrupt is enabled.

#### • PENDX: Pulse End Interrupt

0: Pulse End Interrupt is disabled.

1: Pulse End Interrupt is enabled.





### 22.18 PWM Delay Register x [x = 0..3]

Name: Access: Address:	PWM_D Read/W 0x080	)LY_x /rite					
31	30	29	28	27	26	25	24
-	-	-	-	-	-	_	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
			DE	LAY			
7	6	5	4	3	2	1	0
			DE	LAY			

### • DELAY[15:0]: PWM Delay on Channel X

Number of counter cycles during which the output is inactive.

Note: "X" indicates that the bit concerns the PWMX.

Change of this value is immediately taken into account. This may cause a bad delay on the current pulse.

22.19 PWM	Pulse Regis	ster x $[x = 0]$	.3]				
Name: Access: Address:	PWM_F Read/W 0x084	/rite					
31	30	29	28	27	26	25	24
_	-	-	-	—	—	—	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
			PU	LSE			
7	6	5	4	3	2	1	0
			PU	LSE			

### 22.19 PWM Pulse Register x [x = 0..3]

### • PULSE[15:0]: Pulse Width on Channel X

Number of counter cycles during which the output is active.

Note: "X" indicates that the bit concerns the PWMX.

Change of this value is immediately taken into account. This may cause a bad delay on the current pulse.





## 23. Serial Peripheral Interface (SPI)

#### 23.1 Overview

The AT91SAM7A2 includes a Serial Peripheral Interface (SPI) that provides communication with external devices in Master or Slave Mode.

Seven pins are associated with the SPI interface. When not needed for the SPI function, each of these pins can be configured as a PIO, using PIO Controller functions.

After a hardware reset, the SPI pins are not enabled by default. The user must configure the PIO Controller to enable the corresponding pins to their PIO function.

To configure a SPI pin as open-drain to support external drivers, the software can set the corresponding bits in the multi-driver registers. The NPCS0 pin can function as a peripheral chip select output or slave select input, or as a single PIO.

#### 23.2 Master Mode

In Master Mode, the SPI controls data transfers to and from the slave(s) connected to the SPI bus. The SPI drives the chip select(s) to the slave(s) and the serial clock (SPCK). After enabling the SPI, a data transfer begins when the ARM core writes to the Transmit Data Register (SPI\_TDR).

Transmit and Receive buffers maintain the data flow at a constant rate with a reduced requirement for high-priority interrupt servicing. As long as new data is available in the Transmit Data Register (SPI\_TDR), the SPI continues to transfer data. If the Receive Data Register (SPI\_RDR) has not been read before new data is received, the overrun error (SPIOVRE) flag is set.

The delay between the activation of the chip select and the start of the data transfer (DLYBS) as well as the delay between each data transfer (DLYBCT) can be programmed for each of the four external chip selects. All data transfer characteristics including the two timing values are programmed in registers SPI\_CSR0 to SPI\_CSR3 (Chip Select Registers).

In Master Mode, the peripheral selection can be defined in two different ways:

• Fixed peripheral select: SPI exchanges data with only one peripheral

• Variable peripheral select: Data can be exchanged with more than one peripheral

Note: The SPI master should never be configured as described below:

- SPI master clock equals CORECLK by setting the bit DIV32 to a logical 0 in the SPI mode register (SPI\_MR).

- Pin NPCS0/NSS configured as an open drain by setting bit NPCS0 to a logical 1 in the SPI Multi-Driver Enable Register (SPI\_MDER).

These two options set in the same configuration may cause a mode fault due to the frequency of the CORECLK used to sample the state of the NPCS0/NSS signal after each transmission.

#### 23.2.1 Fixed Peripheral Select

This mode is used for transferring memory blocks without the extra overhead in the Transmit Data Register to determine the peripheral.

Fixed peripheral select is activated by setting bit PS to a logical 0 in the SPI Mode Register (SPI\_MR).

The peripheral is defined by the PCS field in SPI\_MR.

This option is only available when the SPI is programmed in Master Mode.

#### 23.2.2 Variable Peripheral Select

Variable peripheral select is activated by setting bit PS to a logical 1 in the SPI Mode Register (SPI\_MR). The PCS field in Transmit Data Register (SPI\_TDR) is used to select the destination peripheral. The data transfer characteristics are changed when the selected peripheral changes, according to the associated chip select register.

The PCS field in the SPI\_MR has no effect.

This option is only available when the SPI is programmed in Master Mode.

#### 23.2.3 Chip Selects

The Chip Select lines are driven by the SPI only if it is programmed in Master Mode. These lines are used to select the destination peripheral.

The PCSDEC field in SPI\_MR (Mode Register) is used to select one to four peripherals or up to 16 peripherals:

- PCSDEC = 0, each NPCSx acts as a chip select line so up to four devices can be selected.
- PCSDEC = 1, the NPCS[3:0] signals act as an address bus selecting up to 16 peripherals.

If variable peripheral select is active (PS = 1 in SPI\_MR), the chip select signals are defined for each transfer in the PCS field in SPI\_TDR. Chip select signals can thus be defined independently for each transfer.

If fixed peripheral select is active (PS = 0 in SPI\_MR), chip select signals are defined for all transfers by the field PCS in SPI\_MR. If a transfer with a new peripheral is necessary, the software must wait until the current transfer is completed, then change the value of PCS in SPI\_MR before writing new data in SPI\_TDR.

The value on the NPCS pins at the end of each transfer can be read in the Receive Data Register (SPI\_RDR). By default, all NPCS signals are high (equal to one) before and after each transfer.

#### 23.2.4 Mode Fault Detection

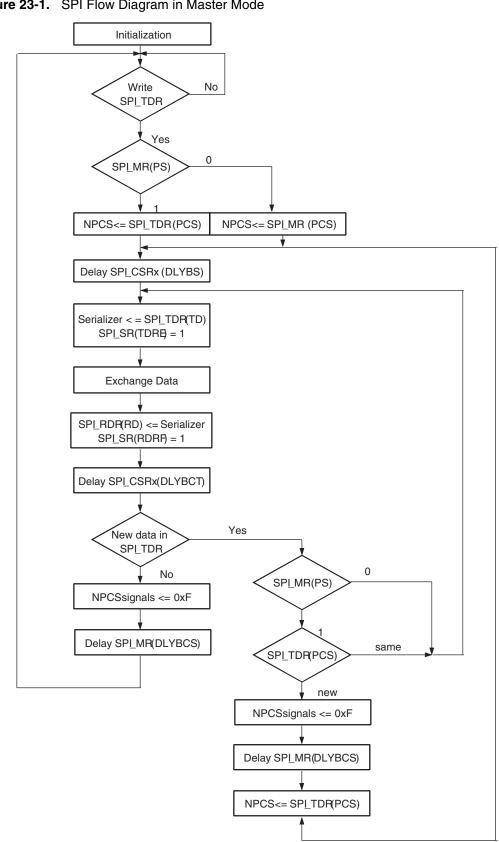
A mode fault is detected when the SPI is programmed in Master Mode and a low level is driven by an external master on the NPCS0 signal.

When a mode fault is detected, the MODF bit in the SPI\_SR is set until the SPI\_SR is read and the SPI is disabled until re-enabled by bit SPIEN in the SPI\_CR (Control Register).

A mode fault is not detected when the master NPCS0 signal is configured in PIO and driven low.



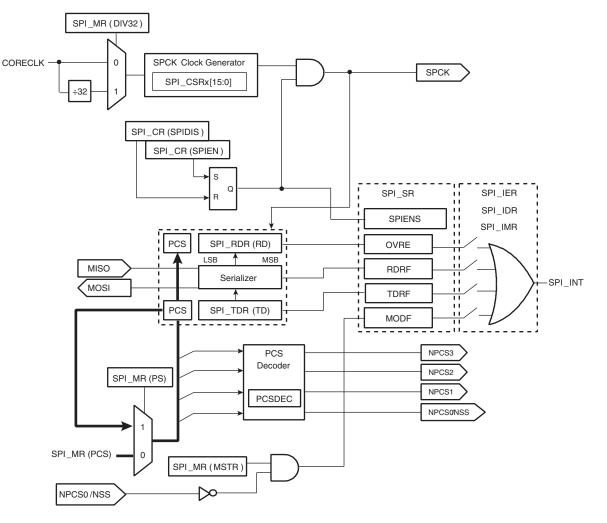






#### 23.2.5 SPI in Master Mode







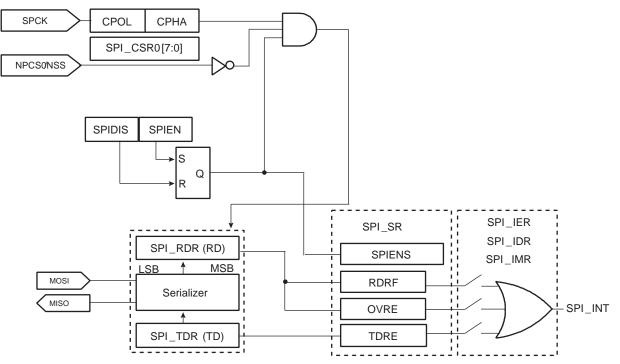


### 23.3 Slave Mode

In Slave Mode, the SPI waits for NSS to go active low before receiving the serial clock from an external master. In Slave Mode, CPOL, NCPHA and BITS fields of SPI\_CSR0 are used to define the transfer characteristics. The other fields in SPI\_CSR0 and the other Chip Select Registers are not used in Slave Mode.

Note: The SPI in Slave Mode can not be used with the PDC for data transmission or reception.





#### 23.4 PIO Controller

The SPI has 7 programmable I/O lines. These I/O lines are multiplexed with signals (MISO, MOSI, SPCK, NPCS[3:0]) of the SPI to optimize the use of available package pins. These lines are controlled by the SPI PIO controller.

#### 23.5 Power Management

The SPI is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).

### 23.6 Serial Peripheral Interface (SPI) Memory Map

Base Address: 0xFFFB4000

### Table 23-1.SPI Memory Map

Offset	Register	Name	Access	Reset State
0x000	PIO Enable Register	SPI_PER	Write-only	-
0x004	PIO Disable Register	SPI_PDR	Write-only	-
0x008	PIO Status Register	SPI_PSR	Read-only	0x007F0000
0x00C	Reserved	_	_	_
0x010	Output Enable Register	SPI_OER	Write-only	-
0x014	Output Disable Register	SPI_ODR	Write-only	_
0x018	Output Status Register	SPI_OSR	Read-only	0x0000000
0x01C - 0x02C	Reserved	_	_	_
0x030	Set Output Data Register	SPI_SODR	Write-only	_
0x034	Clear Output Data Register	SPI_CODR	Write-only	_
0x038	Output Data Status Register	SPI_ODSR	Read-only	0x0000000
0x03C	Pin Data Status Register	SPI_PDSR	Read-only	0x00XX0000
0x040	Multi-Driver Enable Register	SPI_MDER	Write-only	_
0x044	Multi-Driver Disable Register	SPI_MDDR	Write-only	_
0x048	Multi-Driver Status Register	SPI_MDSR	Read-only	0x0000000
0x04C	Reserved	_	_	_
0x050	Enable Clock Register	SPI_ECR	Write-only	_
0x054	Disable Clock Register	SPI_DCR	Write-only	_
0x058	Power Management Status Register	SPI_PMSR	Read-only	0x0000000
0x05C	Reserved	_	_	_
0x060	Control Register	SPI_CR	Write-only	0x0000000
0x064	Mode Register	SPI_MR	Read/Write	0x0000000
0x068 - 0x06C	Reserved	_	_	_
0x070	Status Register	SPI_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	SPI_IER	Write-only	_
0x078	Interrupt Disable Register	SPI_IDR	Write-only	_
0x07C	Interrupt Mask Register	SPI_IMR	Read-only	0x0000000
0x080	Receive Data Register	SPI_RDR	Read-only	_
0x084	Transmit Data Register	SPI_TDR	Write-only	0x00000000
0x088 - 0x08C	Reserved	-	_	_
0x090	Chip Select Register 0	SPI_CSR0	Read/Write	0x0000000
0x094	Chip Select Register 1	SPI_CSR1	Read/Write	0x0000000
0x098	Chip Select Register 2	SPI_CSR2	Read/Write	0x00000000
0x09C	Chip Select Register 3	SPI_CSR3	Read/Write	0x00000000





### 23.7 SPI PIO Enable Register

Name:	SPI_PER
Access:	Write-only
Base Address:	0x000

### 23.8 SPI PIO Disable Register

Name: Access: Base Address:	SPI_PD Write-or : 0x004						
31	30	29	28	27	26	25	24
_	_	_	_	—	_	—	-
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
_	_	_	-	_	_	-	-
7	6	5	4	3	2	1	0
_	_	_	-	_	-	-	_

#### • SPCK: SPCK Pin

0: PIO is inactive on the pin SPCK.

1: PIO is active on the pin SPCK.

#### • MISO: MISO Pin

0: PIO is inactive on the pin MISO.

1: PIO is active on the pin MISO.

#### • MOSI: MOSI Pin

0: PIO is inactive on the pin MOSI.

1: PIO is active on the pin MOSI.

#### • NPCS0: NPCS0 Pin

0: PIO is inactive on the pin NPCS0/NSS.

1: PIO is active on the pin NPCS0/NSS.

#### • NPCSx[x = 1..3]: NPCSx Pin

0: PIO is inactive on the pin NPCSx.

1: PIO is active on the pin NPCSx.

### 23.9 SPI PIO Status Register

Name: Access: Base Address:	SPI_PS Read-or	R					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	. 11	10	9	8
-	_	_	-	-	-	_	-
7	6	5	4	3	2	1	0
-	-	_	-	-	-	_	-

#### • SPCK: SPCK Pin

0: PIO is inactive on the pin SPCK.

1: PIO is active on the pin SPCK.

#### • MISO: MISO Pin

0: PIO is inactive on the pin MISO.

1: PIO is active on the pin MISO.

#### • MOSI: MOSI Pin

0: PIO is inactive on the pin MOSI.

1: PIO is active on the pin MOSI.

#### • NPCS0: NPCS0 Pin

0: PIO is inactive on the pin NPCS0/NSS.

1: PIO is active on the pin NPCS0/NSS.

#### • NPCSx[x = 1..3]: NPCSx Pin

0: PIO is inactive on the pin NPCSx.

1: PIO is active on the pin NPCSx.





### 23.10 SPI PIO Output Enable Register

Name:SPI\_OERAccess:Write-onlyBase Address:0x010

### 23.11 SPI PIO Output Disable Register

Name: Access: Base Address:	SPI_OD Write-or 0x014						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	_	_	_	_	_	_
7	6	5	4	3	2	1	0
-	_	_	_	_	_	_	_

#### • SPCK: SPCK Pin

0: PIO is input on the pin SPCK.

1: PIO is output on the pin SPCK.

#### MISO: MISO Pin

0: PIO is input on the pin MISO.

1: PIO is output on the pin MISO.

#### • MOSI: MOSI Pin

0: PIO is input on the pin MOSI.

1: PIO is output on the pin MOSI.

#### • NPCS0: NPCS0 Pin

0: PIO is input on the pin NPCS0/NSS.

1: PIO is output on the pin NPCS0/NSS.

#### • NPCSx[x = 1..3]: NPCSx Pin

0: PIO is input on the pin NPCSx.

1: PIO is output on the pin NPCSx.

### 23.12 SPI PIO Output Status Register

Name: Access: Base Address:	SPI_OS Read-or						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	-	-	_	—

#### • SPCK: SPCK Pin

0: PIO is input on the pin SPCK.

1: PIO is output on the pin SPCK.

#### • MISO: MISO Pin

0: PIO is input on the pin MISO.

1: PIO is output on the pin MISO.

#### • MOSI: MOSI Pin

0: PIO is input on the pin MOSI.

1: PIO is output on the pin MOSI.

#### • NPCS0: NPCS0 Pin

0: PIO is input on the pin NPCS0/NSS.

1: PIO is output on the pin NPCS0/NSS.

#### • NPCSx [x = 1..3]: NPCSx Pin

0: PIO is input on the pin NPCSx.

1: PIO is output on the pin NPCSx.





### 23.13 SPI PIO Set Output Data Register

Name:	SPI_SODR
Access:	Write-only
Base Address:	0x030

### 23.14 SPI PIO Clear Output Data Register

Name: Access: Base Address:	SPI_CO Write-or 0x034						
31	30	29	28	27	26	25	24
-	-	-	-	_	-	-	-
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	_	-	_	-	_	_
7	6	5	4	3	2	1	0
-	_	-	_	_	_	-	-

#### • SPCK: SPCK Pin

0: The output data for the pin SPCK is programmed to 0.

1: The output data for the pin SPCK is programmed to 1.

#### MISO: MISO Pin

0: The output data for the pin MISO is programmed to 0.

1: The output data for the pin MISO is programmed to 1.

#### • MOSI: MOSI Pin

0: The output data for the pin MOSI is programmed to 0.

1: The output data for the pin MOSI is programmed to 1.

#### • NPCS0: NPCS0 Pin

0: The output data for the pin NPCS0/NSS is programmed to 0.

1: The output data for the pin NPCS0/NSS is programmed to 1.

#### • NPCSx [x = 1..3]: NPCSx Pin

- 0: The output data for the pin NPCSx is programmed to 0.
- 1: The output data for the pin NPCSx is programmed to 1.

### 23.15 SPI PIO Output Data Status Register

Name: Access: Base Address:	SPI_OD Read-or	SPI_ODSR Read-only								
31	30	29	28	27	26	25	24			
-	_	-	-	-	-	-	-			
23	22	21	20	19	18	17	16			
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK			
15	14	13	12	11	10	9	8			
-	_	-	-	-	-	-	-			
7	6	5	4	3	2	1	0			
-	_	-	—	-	—	_	_			

#### • SPCK: SPCK Pin

0: The output data for the pin SPCK is programmed to 0.

1: The output data for the pin SPCK is programmed to 1.

#### • MISO: MISO Pin

0: The output data for the pin MISO is programmed to 0.

1: The output data for the pin MISO is programmed to 1.

#### • MOSI: MOSI Pin

0: The output data for the pin MOSI is programmed to 0.

1: The output data for the pin MOSI is programmed to 1.

#### • NPCS0: NPCS0 Pin

0: The output data for the pin NPCS0/NSS is programmed to 0.

1: The output data for the pin NPCS0/NSS is programmed to 1.

#### • NPCSx [x = 1..3]: NPCSx Pin

- 0: The output data for the pin NPCSx is programmed to 0.
- 1: The output data for the pin NPCSx is programmed to 1.





### 23.16 SPI PIO Pin Data Status Register

Name: Access: Base Address:	SPI_PD Read-or 0x03C						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22 NPCS3	21 NPCS2	20 NPCS1	19 NPCS0	18 MOSI	17 	16 SPCK
-	NPC53	NPC52	NPCST	NPC50	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	_	_	_	_	_

#### • SPCK: SPCK Pin

0: The pin SPCK is at logic 0.

1: The pin SPCK is at logic 1.

#### • MISO: MISO Pin

0: The pin MISO is at logic 0.

1: The pin MISO is at logic 1.

#### • MOSI: MOSI Pin

0: The pin MOSI is at logic 0.

1: The pin MOSI is at logic 1.

#### • NPCS0: NPCS0 Pin

0: The pin NPCS0/NSS is at logic 0.

1: The pin NPCS0/NSS is at logic 1.

#### • NPCSx [x = 1..3]: NPCSx Pin

0: The pin NPCSx is at logic 0.

1: The pin NPCSx is at logic 1.

### 23.17 SPI PIO Multi-driver Enable Register

Name:	SPI_MDER
Access:	Write-only
Base Address:	0x040

### 23.18 SPI PIO Multi-driver Disable Register

Name: Access: Base Address:	SPI_MD Write-or 0x044						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	_	_	_	_	_	_
7	6	5	4	3	2	1	0
-	_	-	-	-	_	_	-

#### • SPCK: SPCK Pin

0: The pin SPCK is not configured as an open drain.

1: The pin SPCK is configured as an open drain.

#### • MISO: MISO Pin

0: The pin MISO is not configured as an open drain.

1: The pin MISO is configured as an open drain.

#### • MOSI: MOSI Pin

0: The pin MOSI is not configured as an open drain.

1: The pin MOSI is configured as an open drain.

#### • NPCS0: NPCS0 Pin

0: The pin NPCS0/NSS is not configured as an open drain.

1: The pin NPCS0/NSS is configured as an open drain.

#### • NPCSx [x = 1..3]: NPCSx Pin

- 0: The pin NPCSx is not configured as an open drain.
- 1: The pin NPCSx is configured as an open drain.





### 23.19 SPI PIO Multi-driver Status Register

Name: Access: Base Address:	SPI_MD Read-or 0x048		-				
31	30	29	28	27	26	25	24
-	_	-	-	_	-	-	-
23	22	21	20	19	18	17	16
_	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	_	_	_	_	-

#### • SPCK: SPCK Pin

0: The pin SPCK is not configured as an open drain.

1: The pin SPCK is configured as an open drain.

#### • MISO: MISO Pin

0: The pin MISO is not configured as an open drain.

1: The pin MISO is configured as an open drain.

#### • MOSI: MOSI Pin

0: The pin MOSI is not configured as an open drain.

1: The pin MOSI is configured as an open drain.

#### • NPCS0: NPCS0 Pin

0: The pin NPCS0/NSS is not configured as an open drain.

1: The pin NPCS0/NSS is configured as an open drain.

#### • NPCSx [x = 1..3]: NPCSx Pin

0: The pin NPCSx is not configured as an open drain.

1: The pin NPCSx is configured as an open drain.

### 23.20 SPI Enable Clock Register

Name:	SPI_ECR
Access:	Write-only
Base Address:	0x050

### 23.21 SPI Disable Clock Register

Name: Access: Base Address:	SPI_DC Write-or 0x054						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	_	_	SPI	PIO

#### • PIO: PIO Clock Status

0: PIO clock disabled.

1: PIO clock enabled.

#### • SPI: SPI Clock Status

0: SPI clock disabled.

1: SPI clock enabled.





### 23.22 SPI Power Management Status Register

Name: Access: Base Address:	SPI_PN Read-or 0x058		-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	_	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	_	-
15	14	13	12	11	10	9	8
-	—	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	-	_	-	-	-	SPI	PIO

#### • PIO: PIO Clock Status

0: PIO clock disabled.

1: PIO clock enabled.

#### • SPI: SPI Clock Status

0: SPI clock disabled.

1: SPI clock enabled.

Note: The SPI\_PMSR register is not reset by software reset.

### 23.23 SPI Control Register

Name: Access: Base Address:	SPI_CR Write-or 0x060	ł					
31	30	29	28	27	26	25	24
-	_	—	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	-	-	-	SPIDIS	SPIEN	SWRST

### • SWRST: SPI Software Reset

0: No effect.

1: Resets the SPI.

A software-triggered hardware reset of the SPI is performed. It reset all the registers (except SPI\_PMSR).

#### • SPIEN: SPI Enable

0: No effect.

1: Enables the SPI.

This enables the SPI to transfer and receive data.

#### • SPIDIS: SPI Disable

0: No effect.

1: Disables the SPI.

All pins will be set to input mode and no data is received or transmitted.

In case a transfer is in progress, the transfer is finished before the SPI is disabled.

In case both SPIEN and SPIDIS are equal to one when the control register is written, the SPI is disabled.





### 23.24 SPI Mode Register

Name: Access: Base Address	SPI_MI Read/V : 0x064						
31	30	29	28	27	26	25	24
			DL	YBCS			
23	22	21	20	19	18	17	16
-	-	-	-		P	CS	
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
LLB	—	-	—	DIV32	PCSDEC	PS	MSTR

#### • MSTR: Master/Slave Mode

0: SPI is in Slave Mode. The SPI in Slave Mode can not be used with the PDC for data transmission or reception.

#### 1: SPI is in Master Mode.

MSTR configures the SPI for either master or Slave Mode operation.

#### • PS: Peripheral Select

0: Fix peripheral select.

1: Variable peripheral select.

The peripheral mode is only used in Master Mode. In case of fix peripheral select, the selected peripheral is defined in the mode register. For variable peripheral select, it is defined in the transmit data register.

#### • PCSDEC: Chip Select Decode

0: The chip selects are directly connected to a peripheral device.

1: The four chip select lines are connected to a 4-to-16 decoder.

In case PCSDEC is one, up to 16 Chip Select signals can be generated with the 4 lines using an external 4 to 16 decoder.

The Chip Select Registers define the characteristics of the 16 chip selects according to the following rules:

- SPI\_CSR0 defines peripheral chip select signals 0 to 3.
- SPI\_CSR1 defines peripheral chip select signals 4 to 7.
- SPI\_CSR2 defines peripheral chip select signals 8 to 11.
- SPI\_CSR3 defines peripheral chip select signals 12 to 15.

#### DIV32: Clock Selection

- 0: SPI Master Clock equals CORECLK.
- 1: SPI Master Clock equals CORECLK/32.

#### • LLB: Local Loopback

0: Local loopback path disabled.

1: Local loopback path enabled.

## <sup>244</sup> **AT91SAM7A2**

LLB controls the local loopback on the data serializer for testing.

#### • PCS[3:0]: Peripheral Chip Select

This field is only used if single peripheral select is active (PS = 0).

```
If PCSDEC = 0:
```

PCS[3:0] = xxx0NPCS[3:0] = 1110PCS[3:0] = xx01NPCS[3:0] = 1101PCS[3:0] = x011NPCS[3:0] = 1011PCS[3:0] = 0111NPCS[3:0] = 0111PCS[3:0] = 1111forbidden (no peripheral is selected)

where x = don't care.

If PCSDEC = 1:

NPCS[3:0] output signals = PCS[3:0]

#### • DLYBCS[7:0]: Delay Between Chip Selects

This field defines the delay from one NPCS inactive to the activation of another NPCS. The DLYBCS[7:0] time will guarantee non-overlapping chip selects and resolves bus contentions in case of peripherals with long data float times.

When DLYBCS[7:0] is less than 6, six SPI Master Clock periods are inserted by default.

Else, the following equation determines the delay:

NPCS\_to\_SPCK\_Delay = DLYBCS[7:0] × SPI\_Master\_Clock\_Period





### 23.25 SPI Status Register

Name: Access: Base Address:	SPI_SR Read-or 0x070						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	_	-
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	—	-	-	—	-	SPIENS
7	6	5	4	3	2	1	0
-	_	TEND	REND	SPIOVRE	MODF	TDRE	RDRF

Note: This register is a "read-active" register, which means that reading it can affect the state of some bits. When reading SPI\_SR register, following bits are cleared if set: MODF, SPIOVRE, REND, TEND, SPCK, MISO, MOSI, NPCS0, NPCS1, NPCS2 and NPCS3. When debugging, to avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

#### • RDRF: Receive Data Register Full

0: No data has been received since the last read of SPI\_RDR.

1: A data has been received and the receive data has been transferred from the serializer in SPI\_RDR since the last read of SPI\_RDR.

#### • TDRE: Transmit Data Register Empty

0: A data has been written in SPI\_TDR and not yet transferred to the serializer.

1: The last data written in the Transmit Data Register has been transferred to the serializer.

TDRE equals zero when the SPI is disabled or at reset. The SPI enable command sets this bit to one.

#### • MODF: Mode Fault Error

0: No Mode Fault has been detected since the last read of SPI\_SR.

1: A Mode Fault occurred since the last read of SPI\_SR.

#### • SPIOVRE: Overrun Error

0: No overrun has been detected since the last read of SPI\_SR.

1: An overrun has occurred since the last read of SPI\_SR.

An overrun occurs when SPI\_RDR is loaded at least twice from the serializer since the last read of the SPI\_RDR.

#### • REND: Reception End

0: No end of PDC reception has been detected since last read of SPI\_SR.

1: At least one end of PDC reception occurs since last read of SPI\_SR.

#### • TEND: Transfer End

0: No end of PDC transfer has been detected since last read of SPI\_SR.

1: At least one end of PDC transfer occurs since last read of SPI\_SR.

## 246 AT91SAM7A2

#### • SPIENS: SPI Enable

0: SPI is disabled.

1: SPI is enabled.

#### • SPCK, MISO, MOSI, NPCS0, NPCS1, NPCS2, NPCS3: PIO Interrupt

These bits indicate for each pin when an input logic value change has been detected (rising or falling edge). This is valid whether the PIO is selected for the pin or not.

These bits are reset to zero following a read and at reset.

0: No input change has been detected on the corresponding pin since the register was last read.

1: At least one input change has been detected on the corresponding pin since the register was last read.





### 23.26 SPI Interrupt Enable Register

Name:SPI\_IERAccess:Write-onlyBase Address:0x074

### 23.27 SPI Interrupt Disable Register

Name:	SPI_IDR	ł					
Access:	Write-on	ly					
Base Address:	0x078						
31	30	29	28	27	26	25	24
-	_	_	-	-	_	_	_
23	22	21	20	19	18	17	16
-	NPCS3	NPCS2	NPCS1	NPCS0	MOSI	MISO	SPCK
15	14	13	12	11	10	9	8
-	_	_	-	_	_	—	-
7	6	5	4	3	2	1	0
-	_	TEND	REND	SPIOVRE	MODF	TDRE	RDRF

#### • RDRF: Receive Data Register Full Interrupt Mask

0: Receive Data Register Full Interrupt is disabled.

1: Receive Data Register Full Interrupt is enabled.

#### TDRE: Transmit Data Register Empty Interrupt Mask

0: Transmit Data Register Empty Interrupt is disabled.

1: Transmit Data Register Empty Interrupt is enabled.

#### • MODF: Mode Fault Error Interrupt Mask

0: Mode Fault Interrupt is disabled.

1: Mode Fault Interrupt is enabled.

Only used in Master Mode.

#### • SPIOVRE: Overrun Error Interrupt Mask

0: Overrun Error Interrupt is disabled.

1: Overrun Error Interrupt is enabled.

#### • REND: PDC Reception End Interrupt Mask

0: Reception End Interrupt is disabled.

1: Reception End Interrupt is enabled.

#### • TEND: PDC Transfer End Interrupt Mask

0: Transfer End Interrupt is disabled.

1: Transfer End Interrupt is enabled.

248 **AT91SAM7A2** 

# AT91SAM7A2

#### • SPCK, MISO, MOSI, NPCS0, NPCS1, NPCS2, NPCS3: PIO Interrupt Mask

These bits show which pins have interrupts enabled. They are updated when interrupts are enabled or disabled by writing to SPI\_IER or SPI\_IDR.

0: Interrupt is not enabled on the corresponding input pin.

1: Interrupt is enabled on the corresponding input pin.





### 23.28 SPI Interrupt Mask Register

Name: Access: Base Address:	SPI_IMI Read-or						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22 NPCS3	21 NPCS2	20 NPCS1	19 NPCS0	18 MOSI	17 MISO	16 SPCK
_	NFC33	NF C32	NFC31	NFC30	WOSI	WI3O	SFCK
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	TEND	REND	SPIOVRE	MODF	TDRE	RDRF

#### • RDRF: Receive Data Register Full Interrupt Mask

0: Receive Data Register Full Interrupt is disabled.

1: Receive Data Register Full Interrupt is enabled.

#### • TDRE: Transmit Data Register Empty Interrupt Mask

0: Transmit Data Register Empty Interrupt is disabled.

1: Transmit Data Register Empty Interrupt is enabled.

#### • MODF: Mode Fault Error Interrupt Mask

0: Mode Fault Interrupt is disabled.

1: Mode Fault Interrupt is enabled.

Only used in Master Mode.

#### • SPIOVRE: Overrun Error Interrupt Mask

0: Overrun Error Interrupt is disabled.

1: Overrun Error Interrupt is enabled.

#### • REND: PDC Reception End Interrupt Mask

0: Reception End Interrupt is disabled.

1: Reception End Interrupt is enabled.

#### • TEND: PDC Transfer End Interrupt Mask

0: Transfer End Interrupt is disabled.

1: Transfer End Interrupt is enabled.SPCK, MISO, MOSI, NPCS0, NPCS1, NPCS2, NPCS3: PIO Interrupt Mask

These bits show which pins have interrupts enabled. They are updated when interrupts are enabled or disabled by writing to SPI\_IER or SPI\_IDR.

0: Interrupt is not enabled on the corresponding input pin.

1: Interrupts is enabled on the corresponding input pin.

#### 23.29 SPI Receive Data Register Name: SPI\_RDR Access: Read-only **Base Address:** 0x080 31 30 29 28 27 26 25 24 \_ \_ \_ \_ \_ \_ \_ \_ 23 22 21 20 19 18 17 16 \_ \_ PCS \_ \_ 9 8 15 14 13 12 11 10 RD 7 6 5 4 3 2 0 1 RD

Note: When reading this register, RDRF bit is cleared in the SPI\_RDR.

Note: When debugging, to avoid clearing RDRF bit, users should use ghost registers (see "Ghost Registers" on page 9).

#### • RD[15:0]: Receive Data

Data received by the SPI interface is stored in this register. Data stored is right-justified. Unused bits are read at zero.

#### • PCS[3:0]: Peripheral Chip Select Status

In Master Mode only, these bits indicate the value on the NPCS pins at the end of a transfer.





### 23.30 SPI Transmit Data Register

Name: Access: Base Address:	SPI_TDI Write-on 0x084						
31	30	29	28	27	26	25	24
-	-	_	-	-	-	_	-
23	22	21	20	19	18	17	16
-	-	_	-		PC	S	
15	14	13	12	11	10	9	8
			Т	D			
7	6	5	4	3	2	1	0
			Т	D			

#### • TD[15:0]: Transmit Data

Data that is to be transmitted by the SPI is stored in this register. Information to be transmitted must be written to the transmit data register in a right-justified format.

#### • PCS[3:0]: Peripheral Chip Select

This field is only used if multiple peripheral select is active (PS = 1).

If PCSDEC = 0:

PCS[3:0] = xxx0	NPCS[3:0] = 1110
PCS[3:0] = xx01	NPCS[3:0] = 1101
PCS[3:0] = x011	NPCS[3:0] = 1011
PCS[3:0] = 0111	NPCS[3:0] = 0111
PCS[3:0] = 1111	forbidden (no peripheral is selected)
are x = den't eare	

where x = don't care.

If PCSDEC = 1:

NPCS[3:0] output signals = PCS[3:0].

## 23.31 SPI Chip Select Register 0..3

Name: Access: Base Address:							
31	30	29	28	27	26	25	24
			DLY	BCT			
23	22	21	20	19	18	17	16
			DL	/BS			
15	14	13	12	11	10	9	8
			SC	BR			
7	6	5	4	3	2	1	0
	BI	TS		—	_	NCPHA	CPOL

#### • CPOL: Clock Polarity

0: The inactive state value of SPCK is logic level zero.

1: The inactive state value of SPCK is logic level one.

CPOL is used to determine the inactive state value of the serial clock (SPCK). It is used with NCPHA to produce a desired clock/data relationship between master and the slave devices.

#### • NCPHA: Clock Phase

0: Data is changed on the leading edge of SPCK and captured on the following edge of SPCK.

1: Data is captured on the leading edge of SPCK and changed on the following edge of SPCK.

NCPHA determines which edge of SPCK causes data to change and which edge causes data to be captured.

NCPHA is used with CPOL to produce a desired clock/data relationship between master and slave devices.

#### • BITS[3:0]: Bits Per Transfer

The BITS field determines the number of data bits transferred. Reserved values default to 8 bits.

BITS[3:0]	Bits Per Transfer	BITS[3:0]	Bits Per Transfer
0000	8	0111	15
0001	9	1000	16
0010	10	1001	Reserved
0011	11	1010	Reserved
0100	12	1011	Reserved
0101	13	11XX	Reserved
0110	14		





#### • SCBR[7:0]: Serial Clock Baud Rate

The SPI interface uses a modulus counter to derive SPCK baud rate from the SPI Master Clock, selected between CORE-CLK and CORECLK/32. Baud rate is selected by writing a value from 2 to 255 into SCBR[7:0]. The following equation determines the SPCK baud rate:

SPCK\_Baud\_Rate = SPI\_Master\_Clock\_Frequency/ (2 × SCBR[7:0])

Note: Giving SCBR[7:0] a value of zero or one disables the baud rate generator. SPCK is disabled and assumes its inactive state value. No serial transfers occur. At reset, baud rate is disabled.

#### • DLYBS[7:0]: Delay Before SPCK

This field defines the length of delay from NPCS valid to the first valid SPCK transition.

When DLYBS[7:0] equals zero, the NPCS valid to SPCK transition is one-half SPCK clock period.

Else, the following equation determines the delay:

NPCS\_to\_SPCK\_Delay = DLYBS[7:0] × SPI\_Master\_Clock\_Period

#### • DLYBCT[7:0]: Delay Between Consecutive Transfers

This field determines the delay between two consecutive transfers with the same peripheral without removing the chip select or the length of delay after transfer before chip select is deselected.

When DLYBCT[7:0] equals zero, the delay is equal to four SPI Master Clock periods.

Else, the following equation determines the delay:

 $Delay\_After\_Transfer = 32 \times DLYBCT[7:0] \times SPI\_Master\_Clock\_Period$ 

## 23.32 Timing Diagrams

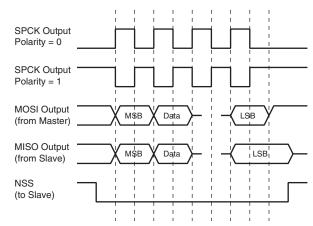
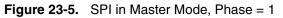


Figure 23-4. SPI in Master Mode, Phase = 0



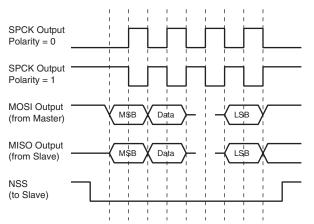
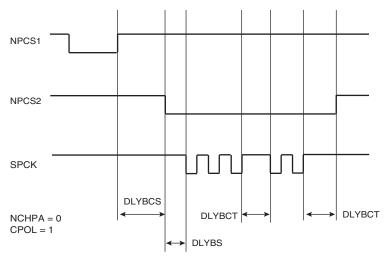


Figure 23-6. DLYBCS, DLYBS and DLYBCT







## Figure 23-7. SPI Timings

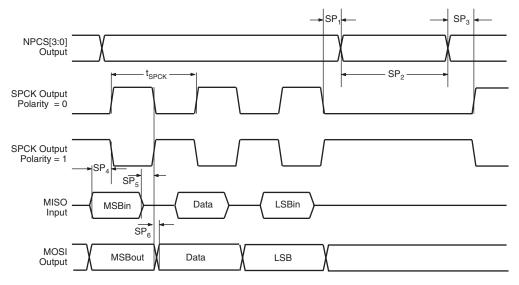


 Table 23-2.
 SPI Timing Parameters

Symbol	Parameter	Minimum	Maximum	Unit
t <sub>SPCK</sub>	SPI Operating Period	4 x (t <sub>CP</sub> )	16320 x (t <sub>CP</sub> )	ns
f <sub>SPCK</sub>	SPI Operating Frequency	1/16320 x (t <sub>CP</sub> )	1/4 x (t <sub>CP</sub> )	GHz
SP1	Delay Before NPCS[3:0]	4 x (t <sub>CP</sub> )	261120 x (t <sub>CP</sub> )	ns
SP <sub>2</sub>	Delay Between Chip Select	6 x (t <sub>CP</sub> )	8160 x (t <sub>CP</sub> )	ns
SP <sub>3</sub>	Delay Before SPCK	2 x (t <sub>CP</sub> )	8160 x (t <sub>CP</sub> )	ns
SP <sub>4</sub>	MISO/SPCK Setup Time		18	ns
SP <sub>5</sub>	MOSI/SPCK Hold Time	0		ns
SP <sub>6</sub>	MOSI valid after SPCK edge		7	ns

## 24. Unified Parallel I/O Controller (UPIO)

### 24.1 Overview

The AT91SAM7A2 microcontroller has 32 unified programmable I/O lines. These lines are controlled by a UPIO module and are not multiplexed with other module pins. The UPIO controller also provides an internal interrupt signal to the Generic Interrupt Controller.

#### 24.2 Output Selection

The user can enable each individual I/O signal as an output with the UPIO\_OER (Output Enable) register or as an input with the UPIO\_ODR (Output Disable) register. The output status of the I/O signals can be read in the register UPIO\_OSR (Output Status).

#### 24.3 I/O Levels

Each pin can be configured to be driven high or low. The level is defined in two different ways, according to the following conditions:

If a pin is defined as output, the level is programmed using the registers UPIO\_SODR (Set Output Data) and UPIO\_CODR (Clear Output Data). In this case, the programmed value can be read in UPIO\_ODSR (Output Data Status).

If a pin is not defined as output, the level is determined by the external circuit.

In all cases, the level on the pin can be read in the register UPIO\_PDSR (Pin Data Status).

#### 24.4 Interrupts

Each parallel I/O can be programmed to generate an interrupt when a level change occurs. This is controlled by the UPIO\_IER (Interrupt Enable) and UPIO\_IDR (Interrupt Disable) registers which enable/disable the I/O interrupt by setting/clearing the corresponding bit in the UPIO\_IMR.

When a change in level occurs, the corresponding bit in the UPIO\_SR (Interrupt Status) is set whether it is defined as input or output. If the corresponding interrupt in UPIO\_IMR (Interrupt Mask) is enabled, the UPIO interrupt is asserted.

When the UPIO\_SR is read, the register is automatically cleared.

#### 24.5 User Interface

Each individual I/O is associated with a bit position in the Parallel I/O user interface registers. If a parallel I/O line is not defined, writing to the corresponding bits has no effect. Undefined bits are read at zero.

## 24.6 Multi-Driver (Open Drain)

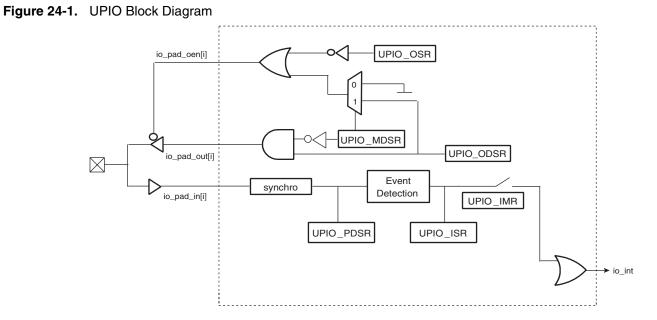
Each I/O can be programmed for multi-driver option. This means that the I/O is configured as open drain (can only drive a low level) in order to support external drivers on the same pin. An external pull-up is necessary to guarantee a logic level of one when the pin is not being driven.

Registers UPIO\_MDER (Multi-Driver Enable) and UPIO\_MDDR (Multi-Driver Disable) control this option. Multi-driver can be selected whether the I/O pin is controlled by the UPIO Controller. UPIO\_MDSR (Multi-Driver Status) indicates which pins are configured to support external drivers.



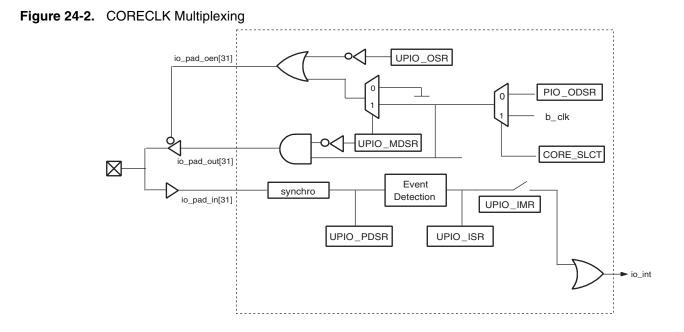


## 24.7 Block Diagram

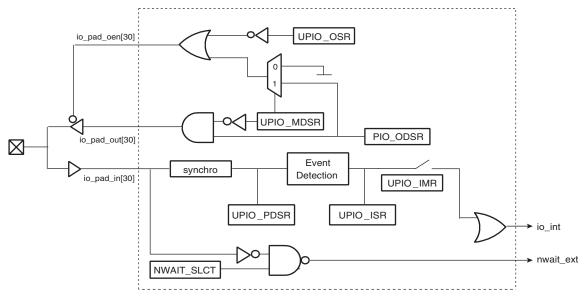


## 24.8 Special Multiplexed PIO

Two dedicated PIO pins, CORECLK and NWAIT, are multiplexed with other signals. CORECLK is multiplexed with PIO[31] and is selected by bit 31 in the PIO\_MR register. NWAIT is multiplexed with PIO[30] and is selected by bit 30 in the PIO\_MR register.



#### Figure 24-3. NWAIT Multiplexing



## 24.9 Power Management

The UPIO is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).





## 24.10 Unified Parallel I/O Controller (UPIO) Memory Map

Base Address: 0xFFFD8000

Table 24-1.	UPIO Memory Map
-------------	-----------------

Offset	Register	Name	Access	Reset State
0x000 - 0x00C	Reserved	-	_	-
0x010	Output Enable Register	UPIO_OER	Write-only	_
0x014	Output Disable Register	UPIO_ODR	Write-only	-
0x018	Output Status Register	UPIO_OSR	Read-only	0x0000000
0x01C - 0x02C	Reserved	-	_	_
0x030	Set Output Data Register	UPIO_SODR	Write-only	_
0x034	Clear Output Data Register	UPIO_CODR	Write-only	_
0x038	Output Data Status Register	UPIO_ODSR	Read-only	0x0000000
0x03C	Pin Data Status Register	UPIO_PDSR	Read-only	0xXXXXXXXX
0x040	Multi-Driver Enable Register	UPIO_MDER	Write-only	-
0x044	Multi-Driver Disable Register	UPIO_MDDR	Write-only	-
0x048	Multi-Driver Status Register	UPIO_MDSR	Read-only	0x0000000
0x04C	Reserved	_	_	-
0x050	Enable Clock Register	UPIO_ECR	Write-only	-
0x054	Disable Clock Register	UPIO_DCR	Write-only	-
0x058	Power Management Status Register	UPIO_PMSR	Read-only	0x0000000
0x05C	Reserved	_	_	-
0x060	Control Register	UPIO_CR	Write-only	-
0x064	Mode Register	UPIO_MR	Read/Write	0x0000000
0x068 - 0x06C	Reserved	_	_	-
0x070	Status Register	UPIO_SR	Read-only	0x0000000
0x074	Interrupt Enable Register	UPIO_IER	Write-only	-
0x078	Interrupt Disable Register	UPIO_IDR	Write-only	-
0x07C	Interrupt Mask Register	UPIO_IMR	Read-only	0x0000000

## 24.11 UPIO Output Enable Register

Name:	UPIO_OER
Access:	Write-only
Base Address:	0x010

## 24.12 UPIO Output Disable Register

Name:	UPIO_ODR
Access:	Write-only
Base Address:	0x014

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

#### • Px: Px Pin

0: The corresponding PIO is input on this line.

1: The corresponding PIO is output on this line.

## 24.13 UPIO Output Status Register

Name: Access: Base Address:	UPIO_OSR Read-only 0x018							
31	30	29	28	27	26	25	24	
P31	P30	P29	P28	P27	P26	P25	P24	
23	22	21	20	19	18	17	16	
P23	P22	P21	P20	P19	P18	P17	P16	
15	14	13	12	11	10	9	8	
P15	P14	P13	P12	P11	P10	P9	P8	
7	6	5	4	3	2	1	0	
P7	P6	P5	P4	P3	P2	P1	P0	

### • Px: Px Pin

0: The corresponding PIO is input on this line.

1: The corresponding PIO is output on this line.





## 24.14 UPIO Set Output Data Register

Name:	UPIO_SODR
Access:	Write-only
Base Address:	0x030

## 24.15 UPIO Clear Output Data Register

Name: Access: Base Address:	UPIO_C Write-or 0x034						
31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

#### • Px: Px Pin

0: The output data for the corresponding pin is programmed to 0.

1: The output data for the corresponding pin is programmed to 1.

## 24.16 UPIO Output Data Status Register

Name: Access: Base Address:	UPIO_ODSR Read-only 0x038							
31	30	29	28	27	26	25	24	
P31	P30	P29	P28	P27	P26	P25	P24	
23	22	21	20	19	18	17	16	
P23	P22	P21	P20	P19	P18	P17	P16	
15	14	13	12	11	10	9	8	
P15	P14	P13	P12	P11	P10	P9	P8	
7	6	5	4	3	2	1	0	
P7	P6	P5	P4	P3	P2	P1	P0	

### • Px: Px Pin

0: The output data for the corresponding pin is programmed to 0.

1: The output data for the corresponding pin is programmed to 1.

## 24.17 UPIO Pin Data Status Register

Name: Access: Base Address:	UPIO_F Read-or 0x030						
31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23 P23	22 P22	21 P21	20 P20	19 P19	18 P18	17 P17	16 P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

#### • Px: Px Pin

0: The corresponding pin is at logic 0.

1: The corresponding pin is at logic 1.





## 24.18 UPIO Multi-driver Enable Register

Name:	UPIO_MDER
Access:	Write-only
Base Address:	0x040

## 24.19 UPIO Multi-driver Disable Register

Name: Access: Base Address:		UPIO_MDDR Write-only 0x044								
31	30	29	28	27	26	25	24			
P31	P30	P29	P28	P27	P26	P25	P24			
23	22	21	20	19	18	17	16			
P23	P22	P21	P20	P19	P18	P17	P16			
15	14	13	12	. 11	10	9	8			
P15	P14	P13	P12	P11	P10	P9	P8			
7	6	5	4	3	2	1	0			
P7	P6	P5	P4	P3	P2	P1	P0			

#### • Px: Px Pin

0: PIO is not configured as an open drain.

1: PIO is configured as an open drain.

## 24.20 UPIO Multi-driver Status Register

Name: Access: Base Address:	UPIO_N Read-or 0x048						
31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

### • Px: Px Pin

0: PIO is not configured as an open drain.

1: PIO is configured as an open drain.

## 24.21 UPIO Enable Clock Register

Name:	UPIO_ECR
Access:	Write-only
Base Address:	0x050

## 24.22 UPIO Disable Clock Register

Name:	UPIO_DCR
Access:	Write-only

### Base Address: 0x054

31	30	29	28	27	26	25	24
-	—	-	—	-	_	-	—
23	22	21	20	19	18	17	16
-	-	-	-	—	-	-	-
45		10	10		10	0	0
15	14	13	12		10	9	8
-	—	-	-	—	-	-	-
7	6	5	4	3	2	1	0
-	—	-	—	-	-	—	PIO

## • PIO: PIO Clock Status

0: PIO clock disabled.

1: PIO clock enabled.

## 24.23 UPIO Power Management Status Register

Access: Base Address:	UPIO_F Read-or	MSR		•			
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	—	-	-	_	-
15	14	13	12	. 11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	_	-	-	-	PIO

## • PIO: PIO Clock Status

0: PIO clock disabled.

1: PIO clock enabled.





## 24.24 UPIO Control Register

Name: Access: Base Address:	UPIO_C Write-or						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	_	-	—	-
7	6	5	4	3	2	1	0
-	_	_	-	—	-	_	SWRST

#### • SWRST: PIO Software Reset

0: No effect.

1: Resets the PIO.

A software-triggered hardware reset of the PIO is performed. It resets all the registers (except the UPIO\_PMSR).

## 24.25 UPIO Mode Register

Name: Access: Base Address	UPIO_W Read/W : 0x064						
31	30	29	28	27	26	25	24
CLK_SLCT	NWAIT_SLCT	_	-	-	-	_	-
23	22	21	20	19	18	17	16
-	-	_	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	Ι	-	-	-	Ι	-
7	6	5	4	3	2	1	0
_	-	_	_	_	-	_	-

### • CLK\_SLCT: Core Clock Select

0: Normal function PIO select.

1: Core Clock select in output mode is connected to PIO[31].

### • NWAIT\_SLCT: NWait External Select

0: Normal function PIO select.

1: PIO[30] is connected to nWAIT external to the EBI module.

#### 24.26 UPIO Status Register UPIO\_SR Name: Access: Read-only **Base Address:** 0x070 31 30 29 28 27 26 25 24 P31 P30 P29 P28 P27 P26 P25 P24 23 22 21 20 19 18 17 16 P23 P22 P21 P20 P19 P18 P17 P16 15 14 13 12 11 10 9 8 P14 P13 P12 P10 P9 P8 P15 P11 7 4 3 2 0 6 5 1 P7 P6 P5 P4 P3 P2 P1 P0

Note: This register is a "read-active" register, thus reading it can affect the state of some bits. When read UPIO\_SR register, all bits set to logical 1 are cleared. When debugging, to avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

#### • Px: PIO Interrupt Status

These bits indicate for each pin when an input logic value change has been detected (rising or falling edge).

These bits are reset to zero following a read and at reset.

0: No input change has been detected on the corresponding pin since the register was last read.

1: At least one input change has been detected on the corresponding pin since the register was last read.





## 24.27 UPIO Interrupt Enable Register

Name:UPIO\_IERAccess:Write-onlyBase Address:0x074

## 24.28 UPIO Interrupt Disable Register

Name: Access: Base Address:	UPIO_IE Write-on 0x078						
31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

#### • Px: PIO Interrupt Mask

0: Interrupt is not enabled on the corresponding input pin.

1: Interrupt is enabled on the corresponding input pin.

## 24.29 UPIO Interrupt Mask Register

Name: Access: Base Address:	UPIO_II Read-or						
31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

## • Px: PIO Interrupt Mask

0: Interrupt is not enabled on the corresponding input pin.

1: Interrupt is enabled on the corresponding input pin.

## 25. Power Management Controller (PMC)

## 25.1 Overview

The AT91SAM7A2 Power Management Controller allows optimization of power consumption. The PMC controls the clock inputs of the ARM core and of the PDC module.

When the ARM core clock is disabled, the current instruction is finished before the clock is stopped. The clock can be re-enabled by any interrupt or by any hardware reset.

The PDC clock cannot be stopped during PDC transfers (if any TCR register is different from 0). The request is stored but the Power Management Controller will wait until the end of the transfers to stop the PDC clock (the Power Management Controller needs an authorization from the PDC before it stops the PDC clock).





## 25.2 Power Management Controller (PMC) Memory Map

Base Address: 0xFFFF4000

Table 25-1.	PMC Memory Map
-------------	----------------

Offset	Register	Name	Access	Reset State
0x000 - 0x04C	Reserved	-	-	-
0x050	Enable Clock Register	PMC_ECR	Write-only	-
0x054	Disable Clock Register	PMC_DCR	Write-only	-
0x058	Power Management Status Register	PMC_PMSR	Read-only	0x00000001

Name: Access:	—						
31	30	29	28	27	26	25	24
—	—	—	-	—	_	—	—
23	22	21	20	19	18	17	16
_	—	-	-	—	—	-	-
15	14	13	12	11	10	9	8
_	_	_	_	_	_	_	-
7	6	5	4	3	2	1	0
_	_	_	-	_	-	PDC	-

# PDC: PDC Clock

0: PDC clock is disabled, or user instructed PDC clock to be disabled during PDC transfers. The PDC clock is disabled when all TCR registers reach 0 (all transfers are finished).

1: PDC clock is enabled.

## 25.4 PMC Disable Clock Register

Name: Access: Base Address:	PMC_D Write-or						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	—	_	_	_	_	_	_
15	14	13	12	11	10	9	8
—	_	-	—	-	—	—	-
7	6	5	4	3	2	1	0
_	_	—	_	—	—	PDC	ARM

#### ARM: ARM Clock

0: ARM core clock is disabled.

1: ARM core clock is enabled.

#### PDC: PDC Clock

0: PDC clock is disabled, or user instructed PDC clock to be disabled during PDC transfers. The PDC clock is disabled when all TCR registers reach 0 (all transfers are finished).

1: PDC clock is enabled.





## 25.5 PMC Power Management Status Register

Name: Access: Base Address:	PMC_P Read-or 0x058						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
—	_	-	-	-	-	PDC	ARM

#### • ARM: ARM Clock

0: ARM core clock is disabled.

1: ARM core clock is enabled.

#### PDC: PDC Clock

0: PDC clock is disabled, or user instructed PDC clock to be disabled during PDC transfers. The PDC clock is disabled when all TCR registers reach 0 (all transfers are finished).

1: PDC clock is enabled.

## 26. Controller Area Network (CAN)

#### 26.1 Overview

The Controller Area Network (CAN) is a serial communications protocol that supports distributed real-time control with a very high level of security.

Possible applications range from high-speed networks to low-cost multiplex wiring, e.g., in automotive electronics, engine control units, sensors, anti-skid systems, etc. may be connected using CAN with bit rates up to 1 Mbit/s. At the same time, it is a cost-effective application for vehicle body electronics, such as lamp clusters, electric windows, etc. in replacement of the wiring harness otherwise required.

This section describes how to achieve compatibility between any two CAN implementations. Compatibility, however, has different aspects regarding, e.g., electrical features and the interpretation of data to be transferred.

To achieve design transparency and implementation flexibility, the CAN has been subdivided into different layers according to the ISO/OSI Reference Model:

- Data Link Layer
  - Logical Link Control (LLC) sublayer
  - Medium Access Control (MAC) sublayer
- Physical Layer

Note that in previous versions of the CAN specification, services and functions of the LLC and MAC sublayers of the Data Link Layer were described in layers denoted as 'object layer' and 'transfer layer'.

The tasks of the LLC sublayer include

- determining which messages received by the LLC sublayer are to be accepted
- providing services for data transfer and for remote data request
- providing the means for recovery management and overload notifications

There is much flexibility in defining object handling.

The tasks of the MAC sublayer concern primarily the transfer protocol, i.e., controlling framing, performing arbitration, error checking, error and fault confinement. Within the MAC sublayer, it is determined whether the bus is free to start signaling a new transmission or whether a reception is just starting. Also, some general features of bit timing are regarded as a task of the MAC sublayer. One of the constraints of the MAC sublayer is that it is not possible to make modifications.

The task of the physical layer is the actual transfer of bits between the different nodes with respect to electrical properties. Within a network, the physical layer has to be the same for all nodes. There are, however, many possible implementations of a physical layer.

In the following, the MAC sublayer and a small part of the LLC sublayer of the Data Link Layer are defined and the consequences of the CAN protocol on the surrounding layers are described.

#### 26.2 Basic Concepts

#### 26.2.1 CAN Properties

The CAN has the following properties:

• Prioritization of messages





- · Guarantee of latency times
- Configuration flexibility
- Multicast reception with time synchronization
- System wide data consistency
- Multi-master functions
- Error detection and error signaling
- · Automatic retransmission of corrupted messages as soon as the bus is idle again
- Distinction between temporary errors and permanent failures of nodes and autonomous switching off of defect nodes.

#### 26.2.2 Layered Structure of a CAN Node

The layered architecture of the CAN is compliant with the ISO/OSI reference model:

- The LLC sublayer is concerned with message filtering, overload notification and recovery management.
- The MAC sublayer represents the kernel of the CAN protocol. It presents messages received from the LLC sublayer and accepts messages to be transmitted to the LLC sublayer. The MAC sublayer is responsible for Message Framing, Arbitration, Acknowledgement, Error Detection and Signalling. The MAC sublayer is supervised by a management entity called Fault Confinement which is a self-checking mechanism for distinguishing short disturbances from permanent failures.
- The physical layer defines how signals are actually transmitted and deals with the description of bit timing, bit encoding, and synchronization. Within this description, the physical layer is not defined, as it will vary according to the requirements of individual applications (for example, transmission medium and signal level implementations).

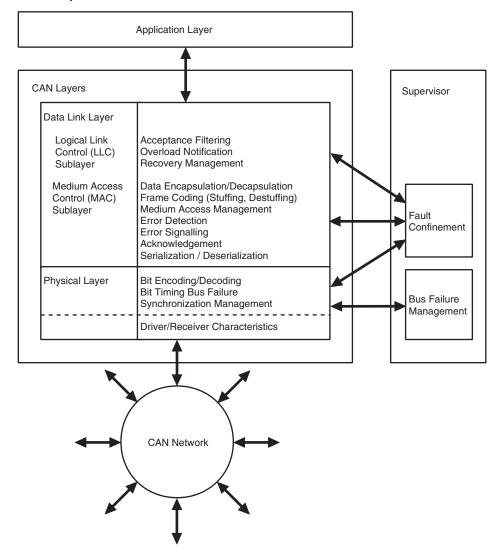


Figure 26-1. Layered Structure of a CAN Node

#### 26.2.3 Messages

Information on the bus is sent in fixed format messages of different but limited length (see "Message Transfer" on page 278). When the bus is free, any connected unit may start to transmit a new message.

#### 26.2.4 Information Routing

In CAN systems, a CAN node does not make use of any information about the system configuration (e.g., node addresses). This has several important consequences:

- System flexibility: Nodes can be added to the CAN network without requiring any change in the software or hardware of any node and application layer.
- Message routing: The content of a message is named by an identifier. The identifier does not
  indicate the destination of the message, but describes the meaning of the data, so that all
  nodes in the network are able to decide, by message filtering, whether the data is to be acted
  upon by them or not.





- Multicast: As a consequence of the message filtering, any number of nodes can receive and simultaneously act upon the same message.
- Data consistency: Within a CAN network, it is guaranteed that a message is simultaneously accepted either by all nodes or by no node. Thus, data consistency of a system is achieved by the concepts of multicast and by error handling.

#### 26.2.5 Bit Rate

The speed of the CAN may be different in different systems. However, in a given system the bitrate is uniform and fixed.

#### 26.2.6 Priorities

The identifier defines a static message priority during bus access.

#### 26.2.7 Remote Data Request

By sending a remote frame, a node requiring data may request another node to send the corresponding data frame. The data frame and the corresponding remote frame are named by the same identifier.

#### 26.2.8 Multi-master Function

When the bus is free, any unit may start transmitting a message. The unit with the message of highest priority to be transmitted gains bus access.

#### 26.2.9 Arbitration

Whenever the bus is free, any unit may start transmitting a message. If two or more units start transmitting messages at the same time, the bus access conflict is resolved by bit-wise arbitration using the identifier. The mechanism of arbitration guarantees that neither information nor time is lost. If a data frame and a remote frame with the same identifier are initiated at the same time, the data frame prevails over the remote frame. During arbitration every transmitter compares the level of the bit transmitted with the level that is monitored on the bus. If these levels are equal the unit may continue to send. When a recessive level is sent and a dominant level is monitored (see "Bus Values" on page 277), the unit has lost arbitration and must withdraw without sending one more bit.

#### 26.2.10 Data Transfer Security

In order to achieve the utmost security of data transfer, powerful measures for error detection, signalling and self-checking are implemented in every CAN node.

#### 26.2.10.1 Error Detection

For detecting errors, the following methods are used:

- Monitoring (transmitters compare the bit levels to be transmitted with the bit levels detected on bus)
- Cyclic Redundancy Check (CRC)
- Bit stuffing
- Message frame check

#### 26.2.10.2 Performance of Error Detection

The error detection mechanisms have the following properties:

• All global errors are detected by means of monitoring.

## 276 AT91SAM7A2 —

- All local errors at transmitters are detected by means of monitoring.
- Up to 5 randomly distributed errors in a message are detected by means of CRC.
- Burst errors of length less than 15 in a message are detected by means of CRC.
- Errors of any odd number in a message are detected by means of CRC.

Total residual error probability for undetected corrupted messages is less than

Message Error Rate  $\times$  4.7  $\times$  10<sup>11</sup>

#### 26.2.11 Error Signaling and Recovery Time

Corrupted messages are flagged by any node detecting an error. Such messages are aborted and will be retransmitted automatically. The recovery time from detecting an error until the start of the next message is at most 31 bit times if there is no further error.

#### 26.2.12 Fault Confinement

CAN nodes are able to distinguish short disturbances from permanent failures. Defective nodes are switched off.

#### 26.2.13 Connections

The CAN serial communication link is a bus to which a number of units may be connected. This number has no theoretical limit. In practice, the total number of units is limited by delay times and/or electrical loads on the bus line.

#### 26.2.14 Single Channel

The bus consists of a single bidirectional channel that carries bits. From this data, resynchronization information can be derived. The way in which this channel is implemented is not fixed in this document, e.g., single wire (plus ground), two differential wires, optical fibers, etc.

#### 26.2.15 Bus Values

The bus can have one of two complementary logical values: dominant or recessive. During simultaneous transmission of dominant and recessive bits, the resulting bus value will be dominant. For example, in case of a wired-AND implementation of the bus, the dominant level is represented by a logical 0 and the recessive level by a logical 1. Physical states (e.g., electrical voltage, light) that represent the logical levels are not given in this document.

#### 26.2.16 Acknowledgement

All receivers check the consistency of the message being received and acknowledge a consistent message and flag an inconsistent message.

#### 26.3 Channel Overview

The CAN module has a set of buffers, also called channels or mailboxes. Table 26-1 presents the number of channels of each CAN module present in the AT91SAM7A2 microcontroller.

Module	Number of Channels
CAN0	16 channels
CAN1	16 channels
CAN2	32 channels
CAN3	16 channels

Table 26-1. Number of Channels





Each mailbox is assigned an identifier and can be set to transmit or receive.

It is possible to reconfigure mailboxes dynamically.

When the CAN module receives a message, it checks the mailboxes in order to see if there is a receive mailbox with the same identifier as the message. If such a mailbox is found, the message is stored in it. If several mailboxes are configured with the same identifier, only the smaller channel store the message. If no mailbox is found, the message is discarded.

When the CAN module has to transmit a message, the message length, data and identifier are written to a mailbox set in transmission. If several messages in different mailboxes are waiting to be transmitted, the mailbox sending the highest priority message (smaller identifier) sends its message first.

If a remote frame is received, the CAN module checks the remote identifier against the mailbox identifier. If a match is found, and if the mailbox is set to automatically reply to the remote frame, the CAN module automatically sends a message with the identifier and data contained in that mailbox.

#### 26.4 Message Transfer

#### 26.4.1 Definition of Transmitter/Receiver

A node originating a message is called the transmitter of that message. The node stays transmitter until the bus is idle or the node loses arbitration.

A node is called receiver of a message if it is not the transmitter of that message and the bus is not idle.

#### 26.4.2 Frame Formats

There are two different formats that differ in the length of the identifier field:

Frame Format	Identifier Length
Standard frame	11 bits
Extended frame	29 bits

Table 26-2.	Identifier Length within Standard and Extended Frames

#### 26.4.3 Frame Types

Message transfer is manifested and controlled by four different frame types:

- A data frame carries data from a transmitter to the receivers.
- A remote frame is transmitted by a bus unit to request the transmission of the data frame with the same identifier.
- An error frame is transmitted by any unit on detecting a bus error.
- An overload frame is used to provide for an extra delay between the preceding and the succeeding data or remote frame.

Data frames and remote frames can be used both in standard frame format and extended frame format. They are separated from preceding frames by an interframe space.

#### 26.4.3.1 Data Frame

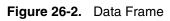
A data frame is composed of seven different bit fields:

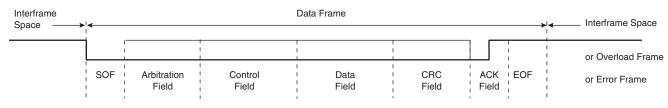
• Start Of Frame (SOF)

## 278 AT91SAM7A2

- Arbitration field
- Control field
- Data field
- CRC field
- ACK field
- End Of Frame (EOF)

The data field can be of length zero.





## Start Of Frame (Standard and Extended Format)

The Start of Frame (SOF) marks the beginning of data frames and remote frames. It consists of a single dominant bit.

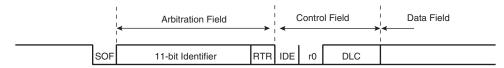
A node is only allowed to start transmission when the bus is idle. All nodes have to synchronize to the leading edge caused by start of frame of the node starting transmission first.

## Arbitration Field

The format of the arbitration field is different for standard format and extended format frames.

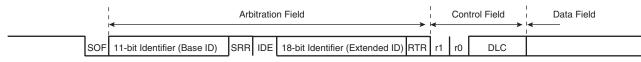
In standard format, the arbitration field consists of the 11-bit identifier and the RTR bit. The identifier bits are denoted ID[10:0].

## Figure 26-3. Standard Format



In extended format, the arbitration field consists of the 29-bit identifier, the SRR bit, the IDE bit, and the RTR bit. The identifier bits are denoted ID[28:0]. The base identifier ID[10:0] are sent first, and extended identifier ID[28:11] later.

## Figure 26-4. Extended Format



In order to distinguish between standard format and extended format, the reserved bit r1 in previous CAN specifications version 1.0-1.2 now is denoted as IDE bit.

• Identifier - Standard Format





The identifier's length is 11 bits and corresponds to the base ID in extended format. These bits are transmitted in the order from ID10 to ID0. The least significant bit is ID0. The 7 most significant bits ID[10:4] must not be all recessive.

• Identifier - Extended Format

In contrast to the standard format, the extended format consists of 29 bits. The format comprises two sections:

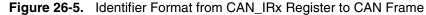
- Base ID: the base ID consists of 11 bits. It is transmitted in the order from ID10 to ID0. It is equivalent to format of the Standard Identifier. The base ID defines the Extended Frame's base priority.
- Extended ID: the Extended ID consists of 18 bits. It is transmitted in the order of ID28 to ID11.

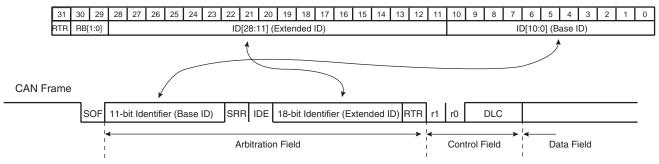
In extended format, the identifier has to be written to the CAN\_IRx register in a specific format:

CAN\_IRx = ((id & 0x1FFC0000) >> 18) | ((id & 0x3FFF) << 11)

where (id & 0x1FFC0000) is the 11-bit base ID and (id & 0x3FFFF) is the 18-bit extended ID.

Figure 26-5 shows how the identifier written in the CAN\_IRx register is transmitted to the CAN network.





• RTR Bit (Standard and Extended Formats)

Table 26-3 shows the RTR bit logical value depending on frame type.

**Table 26-3.** RTR Bit Logical Value Depending on Frame Type

Frame Type	RTR Bit Logical Value	
Data frame	Dominant	
Remote frame	Recessive	

In an extended frame, the base ID is transmitted first, followed by the IDE bit and the SRR bit. The extended ID is transmitted after the SRR bit.

SRR Bit (Extended Format)

The SRR bit (Substitute Remote Request) is a recessive bit. It is transmitted in Extended Frames at the position of the RTR bit in standard frames and so substitutes the RTR bit in the standard frame.

Therefore, collisions of a standard frame and an extended frame, the base ID (see Extended Identifier below) of which is the same as the standard frame's identifier, are resolved in such a way that the standard frame prevails the extended frame.

#### • IDE Bit (Extended Format)

Table 26-4 shows the IDE bit logical value and membership according to frame format type.

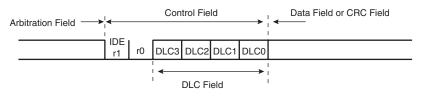
Frame Format Type	IDE Bit Belongs to	IDE Bit Logical Value	
Standard frame format	Control field	Dominant	
Extended frame format	Arbitration field	Recessive	

 Table 26-4.
 IDE Bit Logical Value and Membership Depending on Format Type

#### Control Field (Standard and Extended Format)

The control field consists of six bits. The format of the control field is different for standard format and extended format. Frames in standard format include the data length code, the IDE bit, which is transmitted dominant, and the reserved bit r0. Frames in extended format include the data length code and two reserved bits, r1 and r0. The reserved bits have to be sent dominant, but receivers accept dominant and recessive bits in all combinations.

#### Figure 26-6. Control Field



#### • Data Length Code (Standard and Extended Format)

The number of bytes in the data field is indicated by the data length code. The DLC field is four bits wide and is transmitted within the control field. Coding of the number of data bytes by the DLC field is described in Table 26-5.

Number of Data		Data Len	gth Code	
(Bytes)	DLC3	DLC2	DLC1	DLC0
0	D	D	D	D
1	D	D	D	R
2	D	D	R	D
3	D	D	R	R
4	D	R	D	D
5	D	R	D	R
6	D	R	R	D
7	D	R	R	R
8	R	D	D	D

Table 26-5.	Data Length Code (D = dominant, R = recessive)

Data Frame (Standard and Extended Format)

The admissible numbers of data bytes is 0 to 8. Other values may not be used.



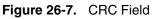


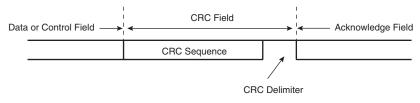
#### Data Field (Standard and Extended Format)

The data field consists of the data to be transferred within a data frame. It can contain from 0 to 8 bytes, which each contain 8 bits which are transferred MSB first.

## CRC Field (Standard and Extended Format)

The CRC (Cyclic Redundancy Check) field contains the CRC sequence followed by a CRC delimiter.





• CRC Sequence (Standard and Extended Format)

The frame check sequence is derived from a cyclic redundancy code best suited for frames with bit counts less than 127 bits (BCH code).

In order to carry out the CRC calculation, the polynomial to be divided is defined as the polynomial, the coefficients of which are given by the de-stuffed bit stream consisting of start of frame, arbitration field, control field, data field (if present) and, for the 15 lowest coefficients, by 0. This polynomial is divided (the coefficients are calculated modulo 2) by the generator-polynomial:

X15 + X14 + X10 + X8 + X7 + X4 + X3 + 1

The remainder of this polynomial division is the CRC sequence transmitted over the bus. In order to implement this function, a 15-bit shift register CRC\_RG(14:0) can be used. If NXTBIT denotes the next bit of the bit stream, given by the de-stuffed bit sequence from start of frame until the end of the data field, the CRC sequence is calculated as follows:

```
CRC_RG = 0// initialize shift register
REPEAT
CRCNXT = NXTBIT EXOR CRC_RG[14]
CRC_RG[14:1] = CRC_RG[13:0]// shift left by
CRC_RG[0] = 0// 1 position
IF CRCNXT THEN
CRC_RG[14:0] = CRC_RG[14:0] EXOR 0x4599
ENDIF
UNTIL (CRC sequence starts or there is an error condition)
```

After the transmission/reception of the last bit of the data field, CRC\_RG contains the CRC sequence.

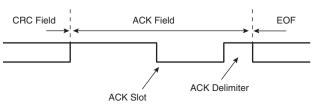
• CRC Delimiter (Standard and Extended Format)

The CRC sequence is followed by the CRC delimiter which consists of a single recessive bit.

## ACK Field (Standard and Extended Format)

The ACK (acknowledgement) field is two bits long and contains the ACK slot and the ACK delimiter. In the ACK field, the transmitting node sends two recessive bits. A receiver that has received a valid message correctly reports this to the transmitter by sending a dominant bit during the ACK slot (it sends ACK).

#### Figure 26-8. ACK Field



#### ACK Slot

All stations having received the matching CRC sequence report this within the ACK slot by overwriting the recessive bit of the transmitter by a dominant bit.

• ACK Delimiter

The ACK delimiter is the second bit of the ACK field and has to be a recessive bit. As a consequence, the ACK slot is surrounded by two recessive bits (CRC delimiter, ACK delimiter).

#### End Of Frame (Standard and Extended Format)

Each data frame and remote frame is delimited by seven recessive bits. These bits form the end of frame sequence (EOF).

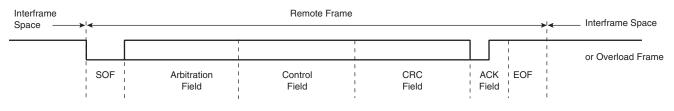
#### 26.4.3.2 Remote Frame

A station acting as a receiver for certain data can initiate the transmission of the respective data by its source node by sending a remote frame. A remote frame exists both in standard format and in extended format. In both cases, it is composed of six different bit fields:

- Start Of Frame (SOF)
- Arbitration field
- Control field
- CRC field
- ACK field
- End Of Frame

Contrary to data frames, the RTR bit of remote frames is recessive. There is no data field, independent of the values of the data length code which may be signed any value within the admissible range from 0 to 8. The value is the data length code of the corresponding data frame.

#### Figure 26-9. Remote Frame



The polarity of the RTR bit (CAN\_IRX register) indicates whether a transmitted frame is a data frame (RTR bit dominant) or a remote frame (RTR bit recessive).





If a mailbox is set to reply automatically to the remote frame (RPLYV bit set to logical 1 in CAN\_CRx register), the CAN module automatically sends a message with the identifier and data contained in that mailbox after receiving the remote frame.

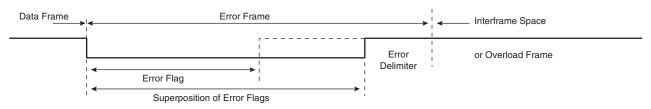
To allow a channel to receive a remote frame, the RTR bit in CAN\_IRx register has to be recessive (i.e., at logical 1), or masked (MRTR bit of CAN\_MSKx register set to a logical 1).

#### 26.4.3.3 Error Frame

The error frame consists of two different fields. The first field is given by the superposition of error flags contributed from different stations. The second field is the error delimiter.

In order to terminate an error frame correctly, an error passive node may need the bus to be bus idle for at least three bit times if there is a local error at an error passive receiver. Therefore, the bus should not be loaded to 100%.

#### Figure 26-10. Error Frame



#### Error Flag

There are two forms of error flag: an active error flag and a passive error flag.

- The active error flag consists of six consecutive dominant bits.
- The passive error flag consists of six consecutive recessive bits unless it is overwritten by dominant bits from other nodes.

An error active node detecting an error condition signals this by transmission of an active error flag. The error flag's form violates the law of bit stuffing (see "Coding" on page 287) applied to all fields from start of frame to CRC delimiter or destroys the fixed form ACK field or end of frame field. As a consequence, all other nodes detect an error condition and start transmission of an error flag. Thus the sequence of dominant bits that can actually be monitored on the bus results from a superposition of different error flags transmitted by individual nodes. The total length of this sequence varies between a minimum of six and a maximum of twelve bits.

An error passive node detecting an error condition tries to signal this by transmission of a passive error flag. The error passive node waits for six consecutive bits of equal polarity, beginning at the start of the passive error flag. The passive error flag is complete when these six equal bits have been detected.

#### Error Delimiter

The error delimiter consists of eight recessive bits.

After transmission of an error flag, each node sends recessive bits and monitors the bus until it detects a recessive bit. Afterwards, it starts transmitting seven more recessive bits.

#### 26.4.3.4 Overload Frame

The overload frame (this emission can be enabled/disabled in CAN\_CR register) contains the two bit fields, overload flag and overload delimiter. There are two kinds of overload conditions that lead to the transmission of an overload flag:

## 284 **AT91SAM7A2**

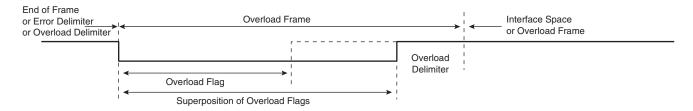
- 1. The internal conditions of a receiver, requiring a delay of the next data frame or remote frame.
- 2. Detection of a dominant bit at the first and second bit of intermission.

If a CAN node samples a dominant bit at the eighth bit (the last one) of an error delimiter or overload delimiter, it starts transmitting an overload frame (not an error frame). The error counters are incremented.

The start of an overload frame due to the first overload condition is only allowed to be started at the first bit time of an expected intermission, whereas overload frames due to the second overload condition and condition 3 start one bit after detecting the dominant bit.

At most two overload frames may be generated to delay the next data or remote frame.

#### Figure 26-11. Overload Frame



#### Overload Flag

The overload flag consists of six dominant bits. The overall form corresponds to that of the active error flag.

The overload flag's form destroys the fixed form of the intermission field. As a consequence, all other nodes also detect an overload condition and start transmission of an overload flag. If there is a dominant bit detected during the 3rd bit of intermission, then it interpret this bit as start of frame.

Note: Controllers based on the CAN Specification version 1.0 and 1.1 have another interpretation of the 3rd bit of intermission: if a dominant bit was detected locally at some node, the other nodes do not interpret the overload flag correctly, but interpret the first of these six dominant bits as start of frame; the sixth dominant bit violates the rule of bit stuffing causing an error condition.

#### **Overload Delimiter**

The overload delimiter consists of eight recessive bits.

The overload delimiter is of the same form as the error delimiter. After transmission of an overload flag, the node monitors the bus until it detects a transition from a dominant to a recessive bit. At this time, every bus node has finished sending its overload flag and all nodes start transmission of seven more recessive bits simultaneously.

#### 26.4.3.5 Interframe Spacing

Data frames and remote frames are separated from preceding frames whatever type they are (data frame, remote frame, error frame, overload frame) by a bit field called interframe space. In contrast, overload frames and error frames are not preceded by an interframe space and multiple overload frames are not separated by an interframe space.

#### Interframe Space

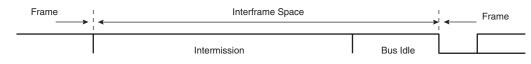
The interframe space contains the bit field intermission and bus idle and, for error passive nodes that have been transmitter of the previous message, suspend transmission.





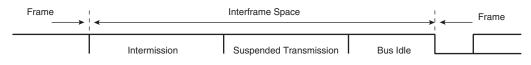
For nodes that are not error passive or have been receiver of the previous message, see Figure 26-12.

#### Figure 26-12. Interframe Space for Receiver



For error passive nodes which have been transmitter of the previous message, see Figure 26-13.

#### Figure 26-13. Interframe Space for Transmitter



#### Intermission

Intermission consists of three recessive bits.

During intermission, the only action to be taken is signalling an overload condition. No node is allowed to actively start a transmission of a data frame or remote frame.

Note: If a CAN node has a message waiting for transmission and it samples a dominant bit at the third bit of intermission, it will interpret this as a start of frame bit, and, with the next bit, start transmitting its message with the first bit of its identifier without first transmitting a start of frame bit and without becoming receiver.

#### Bus Idle

The period of bus idle may be of arbitrary length. The bus is recognized to be free and any node having something to transmit can access the bus. A message, which is pending for transmission during the transmission of another message, is started in the first bit following intermission.

The detection of a dominant bit on the bus is interpreted as start of frame.

#### Suspend Transmission

After an error passive node has transmitted a message, it sends eight recessive bits following intermission, before starting to transmit a further message or recognizing the bus to be idle. If a transmission (caused by another node) starts in the meantime, the node becomes receiver of this message.

#### 26.4.4 Conformance with Frame Formats

The Standard Format is equivalent to the Data/Remote Frame Format as described in the CAN Specification 1.2. In contrast, the Extended Format is a new feature of the CAN protocol. In order to allow the design of relatively simple controllers, the implementation of the Extended Format to its full extent is not required (e.g., send messages or accept data from messages in Extended Format), whereas the Standard Format must be supported without restriction.

New controllers are considered to be in conformance with this CAN Specification if they have at least the following properties with respect to the frame formats defined in "Definition of Transmitter/Receiver" on page 278 and "Frame Types" on page 278.

• Every new controller supports the standard format.

## 286 **AT91SAM7A2**

• Every new controller can receive messages of the extended format. This requires that extended frames are not destroyed just because of their format. However, it is not required that the extended format must be supported by new controllers.

#### 26.5 Message Filtering

Message filtering is based upon the whole identifier. Mask registers that allow any identifier bit to be set "don't care" for message filtering may be used to select groups of identifiers to be mapped into the attached received buffers.

If mask registers are implemented, every bit of the mask registers must be programmable, i.e., they can be enabled or disabled for message filtering. The length of the mask register can comprise the whole identifier or only part of it.

#### 26.6 Message Validation

The point in time at which a message is taken to be valid is different for the transmitter and the receiver of the message.

• Transmitter

The message is valid for the transmitter if there is no error until the end of end of frame. If a message is corrupted, retransmission will follow automatically and according to prioritization. In order to be able to compete for bus access with other messages, retransmission has to start as soon as the bus is idle.

Receivers

The message is valid for the receivers if there is no error until the first-to-last bit of end of frame. The value of the last bit of EOF is treated as "don't care", a dominant value does not lead to a FORM error (see "Error Detection" on page 287).

## 26.7 Coding

In bit stream coding, the frame segments start of frame, arbitration field, control field, data field and CRC sequence are coded by bit stuffing. Whenever a transmitter detects five consecutive bits of identical value in the bit stream to be transmitted, it automatically inserts a complementary bit in the currently transmitted bit stream.

The remaining bit fields of the data frame or remote frame (CRC delimiter, ACK field and end of frame) are of fixed form and not stuffed. The error frame and the overload frame are of fixed form as well and not coded via bit stuffing.

The bit stream in a message is coded according to the Non-Return-to-Zero (NRZ) method. This means that during the total bit time the generated bit level is either dominant or recessive.

## 26.8 Error Handling

#### 26.8.1 Error Detection

There are 5 different error types (not mutually exclusive):

• Bit error (BUS bit in CAN\_SRX): A unit that is sending a bit on the bus also monitors the bus. A bit error has to be detected when the bit value that is monitored is different from the bit value that is sent. An exception is the sending of a recessive bit during the stuffed bit stream of the arbitration field or during the ACK slot. In this case, no bit error occurs when a





dominant bit is monitored. A transmitter sending a passive error flag and detecting a dominant bit does not interpret this as a bit error.

- Stuff error (STUFF bit in CAN\_SRX): A stuff error is detected at the bit time of the 6th consecutive equal bit level in a message field that should be coded by the method of bit stuffing.
- CRC error (CRC bit in CAN\_SRX): The CRC sequence consists of the result of the CRC calculation by the transmitter. The receivers calculate the CRC in the same way as the transmitter. A CRC error has to be detected if the calculated result is not the same as that received in the CRC sequence.
- Form error (FRAME bit in CAN\_SRX): A form error has to be detected when a fixed-form bit field contains one or more illegal bits. (Note that for a receiver a dominant bit during the last bit of EOF is not treated as a form error).
- Acknowledgement error (ACK bit in CAN\_SRX): An acknowledgement error is detected by a transmitter whenever it does not monitor a dominant bit during ACK slot.

#### 26.8.2 Error Signalling

A node detecting an error condition signals this by transmitting an error flag. For an error active node, it is an active error flag; for an error passive node, it is a passive error flag.

Whenever a bit error, a stuff error, a form error or an acknowledgement error is detected by any node, transmission of an error flag is started at the respective node at the next bit.

Whenever a CRC error is detected, transmission of an error flag starts at the bit following the ACK delimiter, unless an error flag for another error condition has already been started.

#### 26.9 Fault Confinement

Regarding fault confinement, a unit may be in one of three states:

- · error active
- error passive
- bus off

An error active unit can normally take part in bus communication and sends an active error flag when an error has been detected.

An error passive unit must not send an active error flag. It takes part in bus communication, but when an error has been detected, only a passive error flag is sent. After a transmission, an error passive unit will wait before initiating a further transmission (see "Suspend Transmission" on page 286).

A bus off unit is not allowed to have any influence on the bus. (e.g., output drivers switched off).

For fault confinement, two counts are implemented in every bus unit:

- transmit error count
- receive error count

These counts are modified according to the following rules (note that more than one rule may apply during a given message transfer):

• When a receiver detects an error, the receive error count will be increased by 1, except when the detected error was a bit error during the sending of an active error flag or an overload flag.

- When a receiver detects a dominant bit as the first bit after sending an error flag, the receive error count will be increased by 8.
- When a transmitter sends an error flag, the transmit error count is increased by 8 except:
  - if the transmitter is error passive and detects an acknowledgement error because of not detecting a dominant ACK. It does not detect a dominant bit while sending its passive error flag.
  - if the transmitter sends an error flag because a stuff error occurred during arbitration, and should never been recessive, and has been sent as recessive but monitored as dominant

In case of one of these exceptions, the transmit error count is not changed.

- If a transmitter detects a bit error while sending an active error flag or an overload flag, the transmit error count is increased by 8.
- If a receiver detects a bit error while sending an active error flag or an overload flag, the receive error count is increased by 8.
- Any node tolerates up to 7 consecutive dominant bits after sending an active error flag, passive error flag or overload flag. After detecting the 14th consecutive dominant bit (in case of an active error flag or an overload flag) or after detecting the 8th consecutive dominant bit following a passive error flag, and after each sequence of additional eight consecutive dominant bits, every transmitter increases its transmit error count by 8 and every receiver increases its receive error count by 8.
- After the successful transmission of a message (getting ACK and no error until end of frame is finished), the transmit error count is decreased by 1 unless it was already 0.
- After the successful reception of a message (reception without error up to the ACK slot and the successful sending of the ACK bit), the receive error count is decreased by 1, if it was between 1 and 127. If the receive error count was 0, it stays 0, and if it was greater than 127, then it will be set to a value between 119 and 127.
- A node is error passive when the transmit error count equals or exceeds 128, or when the receive error count equals or exceeds 128. An error condition letting a node become error passive causes the node to send an active error flag.
- A node is bus off when the transmit error count is greater than or equal to 256.
- An error passive node becomes error active again when both the transmit error count and the receive error count are less than or equal to 127.
- An node that is bus off is permitted to become error active (no longer bus off) with its error counters both set to 0 after 128 occurrences of 11 consecutive recessive bits have been monitored on the bus.
- Note: An error count value greater than 96 indicates a heavily disturbed bus. It may be useful to provide means to test for this condition.
- Note: Start-up/Wake-up: If during system start-up only one node is online, and if this node transmits a message, it gets no acknowledgement, detects an error and repeats the message. It can become error passive but not bus off due to this reason.

### 26.10 Oscillator Tolerance

A maximum oscillator tolerance of 1.58% is given and therefore a ceramic resonator can be used at a bus speed of up to 125 Kbits/s as a rule of thumb. For a more precise evaluation, refer to "Impact of Bit Representation on Transport Capacity and Clock Accuracy in Serial Data





Streams" by S. Dais and M. Chapman, SAE Technical Paper Series 890532, Multiplexing in Automobile SP-773, March 1989.

For the full bus speed range of the CAN protocol, a quartz oscillator is required.

The chip of the CAN network with the highest requirement for its oscillator accuracy determines the oscillator accuracy which is required from all the other nodes.

Note: CAN controllers compliant with this CAN Specification and controllers compliant with the previous versions 1.0 and 1.1, used in one and the same network, must all be equipped with a quartz oscillator. Thus ceramic resonators can only be used in a network with all the nodes of the network according to the CAN Protocol Specification versions 1.2 or later.

# 26.11 Bit Timing Requirements

#### 26.11.1 Nominal Bit Rate

The nominal bit rate is the number of bits per second transmitted in the absence of resynchronization by an ideal transmitter.

#### 26.11.2 Nominal Bit Time

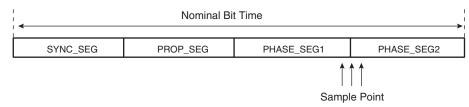
The nominal bit time of the network is uniform throughout the network and is given by:

Nominal Bit Time = 1/ Nominal Bit Rate

The nominal bit time can be thought of as being divided into separate non-overlapping time segments. These segments are:

- synchronization segment (SYNC\_SEG)
- propagation time segment (PROP\_SEG)
- phase buffer segment 1 (PHASE\_SEG1)
- phase buffer segment 2 (PHASE\_SEG2)

### Figure 26-14. Partition of the Bit Time



The duration of each segment is set in the mode register and is expressed as a multiple of time quantum  $(t_{CAN})$ .

### 26.11.3 Time Quantum

Each segment (SYNC\_SEG, PROP\_SEG, PHASE\_SEG1 and PHASE\_SEG2) is an integer multiple of a unit of time called a Time Quantum ( $t_{CAN}$ ). The duration of a Time Quantum is equal to the period of the CAN system clock ( $t_{CAN}$ ), which is derived from the microcontroller system clock (CORECLK) by way of a programmable prescalar, called the Baud Rate Prescalar (in the CAN\_MR register).

The formula relating  $t_{CAN}$ , CORECLK and BD[5:0] is the following:

$$t_{CAN} = \frac{(BD[5:0]+1)}{CORECLK}$$

290 **AT91SAM7A2** 

### 26.11.4 Synchronization Segment

This part of the bit time is used to synchronize the various nodes on the bus. An edge is expected to lie within this segment.

The duration of the synchronization segment, SYNC\_SEG, is not programmable and is fixed at one time quantum.

### 26.11.5 Propagation Segment

This part of the bit time is used to compensate for the physical delay times within the network. It is twice the sum of the signal's propagation time on the bus line, the input comparator delay, and the output driver delay.

The duration of the propagation segment, PROP\_SEG, may be between 1 and 8 time quanta. It is programmable using the PROP bits in the CAN\_MR, and is equal to:

$$t_{PRS} = t_{CAN} \times (PROP[2:0] + 1)$$

### 26.11.6 PHASE\_SEG1, PHASE\_SEG2

The phase buffer segments are used to compensate for edge phase errors. The segments can be lengthened or shortened by resynchronization.

The duration of the segment PHASE\_SEG1 may be between 1 and 8 time quanta.

The duration of segment PHASE\_SEG2 is the maximum of PHASE\_SEG1 and the information processing time (See "Information Processing Time (IPT)" on page 291).

PHASE\_SEG1 and PHASE\_SEG2 are programmable using the PHSEG1[2:0] and PHSEG1[2:0] bits in the MR, respectively. They respect the following equations:

 $t_{PHS1} = t_{CAN} \times (PHSEG1[2:0] + 1)$  $t_{PHS2} = t_{CAN} \times (PHSEG2[2:0] + 1)$ 

Setting the duration of segment PHASE\_SEG2 to one time quanta is forbidden (PHASE-SEG[2:0] field in CAN\_MR must be set to 1 at minimum).

#### 26.11.7 Sample Point

The sample point is the point of time at which the bus level is read and interpreted as the value of that respective bit. Its location is at the end of PHASE\_SEG1. Two methods can be used: the incoming stream is sampled once at a sample point, or sampled 3 times with a period of half a CAN clock period, centered at one sample point.

#### 26.11.8 Information Processing Time (IPT)

The information processing time is the time segment starting with the sample point reserved for calculation of the subsequent bit level.

In other words, the time after the sample point that is needed to calculate the next bit to be sent (e.g., data bit, CRC bit, stuff bit, error flag, or idle) is called the Information Processing Time (IPT).

The CAN module has zero delay IPT.

### 26.11.9 Length of Time Segments

- SYNC\_SEG is 1 time quantum long.
- PROP\_SEG is programmable to be between 1 and 8 time quanta long.



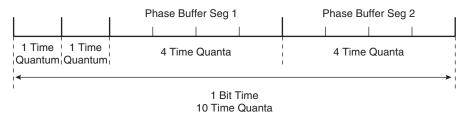


- PHASE\_SEG1 is programmable to be between 1 and 8 time quanta long.
- PHASE\_SEG2 is the maximum of PHASE\_SEG1 and the information processing time.
- The information processing time is 0 time quanta long.

The total number of time quanta in a bit time is programmable at least from 8 to 25.

Note: It is often intended that control units do not make use of different oscillators for the local CPU and its communication device. Therefore, the oscillator frequency of a CAN device tends to be that of the local CPU and is determined by the requirements of the control unit. In order to derive the desired bit rate, programmability of the bit timing is necessary. In case of CAN implementations that are designed for use without a local CPU, the bit timing cannot be programmable. On the other hand, these devices allow selection of an external oscillator in such a way that the device is adjusted to the appropriate bit rate so that the programmability is dispensable for such components.

The position of the sample point, however, should be selected in common for all nodes. Therefore, the bit timing of CAN devices without local CPU should be compatible with the definition of the bit time in Figure 26-15.



#### Figure 26-15. Example of Nominal Bit Time

#### 26.11.10 Hard Synchronization

After a hard synchronization, the internal bit time is restarted with SYNC\_SEG. Thus hard synchronization forces the edge that caused the hard synchronization to lie within the synchronization segment of the restarted bit time.

### 26.11.11 Resynchronization Jump Width

As a result of resynchronization, PHASE\_SEG1 may be lengthened or PHASE\_SEG2 may be shortened. The amount of lengthening or shortening of the phase buffer segments has an upper bound given by the resynchronization jump width. The resynchronization jump width is programmable between 1 and min(4, PHASE\_SEG1).

Clocking information may be derived from transitions from one bit value to the other. The property that only a fixed maximum number of successive bits have the same value provides the possibility of resynchronizing a bus unit to the bit stream during a frame. The maximum length between two transitions that can be used for resynchronization is 29 bit times.

Setting the resynchronization jump width to 1 time quanta (SJW[1:0] field in CAN\_MR register set to '0') is allowed only if the duration of segment PHASE\_SEG2 is lower or equal to 4 time quanta (PHSEG2[2:0] field in CAN\_MR lower or equal to 3). If the duration of segment PHASE\_SEG2 is greater or equal to 5 time quanta (PHSEG2[2:0] field in CAN\_MR greater or equal to 4), the resynchronization jump width must be configured at minimum to 2 time quanta (SJW[1:0] field in CAN\_MR register set to 1 at minimum).

### 26.11.12 Phase Error of an Edge

The phase error of an edge is given by the position of the edge relative to SYNC\_SEG, measured in time quanta. The sign of phase error is defined as follows:

- e = 0 if the edge lies within SYNC\_SEG.
- e > 0 if the edge lies before the SAMPLE POINT.
- e < 0 if the edge lies after the SAMPLE POINT of the previous bit.

#### 26.11.13 Resynchronization

The effect of a resynchronization is the same as that of a hard synchronization when the magnitude of the phase error of the edge causing the resynchronization is less than or equal to the programmed value of the resynchronization jump width. When the magnitude of the phase error is larger than the resynchronization jump width,

- and if the phase error is positive, then PHASE\_SEG1 is lengthened by an amount equal to the resynchronization jump width.
- and if the phase error is negative, then PHASE\_SEG2 is shortened by an amount equal to the resynchronization jump width.

#### 26.11.14 Synchronization Rules

Hard synchronization and resynchronization are the two forms of synchronization. They obey the following rules:

- 1. Only one synchronization within one bit time is allowed.
- 2. An edge is used for synchronization only if the value detected at the previous sample point (previous read bus value) differs from the bus value immediately after the edge.
- 3. Hard synchronization is performed whenever there is a recessive-to-dominant edge during bus idle.
- 4. All other recessive-to-dominant edges fulfilling rules 1 and 2 are used for resynchronization with the exception that a node transmitting a dominant bit does not perform a resynchronization as a result of a recessive-to-dominant edge with a positive phase error if only recessive-to-dominant edges are used for resynchronization.

### 26.12 Reception Mode

In reception, channels can be configured in two different modes:

- Normal mode (OVERWRITE bit reset to 0 in CAN\_CRX): the channel is disabled after a successful reception.
- Overwrite mode (OVERWRITE bit set to 1 in CAN\_CRX): the channel is still enabled after a successful reception.

### 26.13 Time Stamp

A 32-bit stamp dates all messages sent/received (depending on the producer/consumer bit).

The 32-bit register forming the second counter in the WT module is provided to the CAN module. After each transmission or reception of a CAN frame, the value of the current second counter is automatically written in the corresponding CAN channel CAN\_STPx register.

### 26.14 Power Management

The CAN is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).





# 26.15 Using the CAN Module

### 26.15.1 Message Transmission

#### 26.15.1.1 Processing Transmission Request Delay

When a CAN channel configured in transmission is enabled, the transmission of the frame does not start immediately but after a short delay. This delay is a result of channels scanning. When the bus is idle, the CAN state machine scans continually all channels looking for transmit channels. During a scanning cycle - starting from channel 0 and ending to channel 15 or 31 - if several channels are configured in transmission, the CAN state machine memorizes the channel with the lowest identifier, and at the end of the scanning cycle it starts the transmission of the highest priority frame.

Scanning delay of CAN channels depends on the channel state : a channel enabled and configured in transmission is scanned in 3 system clock periods, otherwise it takes 2 system clock periods. Then, scanning all the channels takes at maximum 49 system clock periods for a 16 channels CAN and 97 system clock periods for a 32 channels CAN.

As the CPU can enable a transmit channel at any moment during the scanning cycle, a result is that the delay from the moment the channel is enabled by the CPU to the moment the frame starts to be transmitted is indeterminate and can take at maximum two times the delay for scanning all channels. Moreover, as the scan of the channels is also asynchronous with the bit time, between 0 and 1 bit times should be added to this delay.

#### 26.15.1.2 Caution with Transmit Channel

Once a transmission has been request, users should not modify the frame to transmit (CAN\_IRX, CAN\_DRAX, CAN\_DRBX and CAN\_CRX registers) until the transmission is completed (flag TXOK set to '1' in the CAN\_SRx register). In order to modify a frame to transmit, users should cancel the transmission, change the frame and then request a new transmission. Find below how to cancel properly a transmission.

### 26.15.1.3 Canceling a Pending Transmission

It is allowed to cancel a pending transmission but users should respect scrupulously the following steps :

- Clear the CHANEN flag in the CAN\_CRx register. Note that other bits of this register should not be modified (a read-modify is mandatory).
- Wait for a minimum delay before setting a new reception or transmission in this channel. This delay depends of the identifier length configured in the pending transmission to cancel. The delay is :
  - for a standard frame (11 bit length identifier) : 97 system clock periods + 106 CAN bit time
  - for an extended frame (29 bit length identifier) : 97 system clock periods + 131 CAN bit time
- Afterwards, the frame related registers (CAN\_IRX, CAN\_DRAX, CAN\_DRBX and CAN\_CRX registers) can be modified.

It ensures that if the frame transmission process has started before the cancel has been requested, the frame transmitted has correct identifier, data and control.

#### 26.15.2 Frequency Limitation

In order to ensure proper CAN operation, the CAN module should be clocked with a minimum frequency. This minimum frequency can be computed with the following formula :

Min frequency : (1 + 3 \* number of channels enable in transmission + 2 \* reminder channels) / 3 CAN bit time.

For example, for an application using only one channel in transmission, a CAN 32 channels and a CAN baud rate of 1 MHz, the minimum frequency is 22 MHz ( $22MHz = (1 + 3*1 + 2*31)/3\mu s$ ).

### 26.16 Example of Use

This section gives an example of use of the CAN, explaining how to transmit a data frame of 5 bytes with an extended identifier on channel 0, and how to receive a data frame of 5 bytes with an extended identifier on channel 1.

The interrupt is configured in transmission and in reception. Core clock is 30 MHz.

### 26.16.1 Configuration

- Enable the clock on CAN peripheral by writing bit CAN in CAN\_ECR.
- Do a software reset of the CAN peripheral to be in a known state by writing bit SWRST in CAN\_CR. Software should wait 1 cycle\*nbchannel\*8 for channel register initialization. Users can also use the associated interrupt flag ENDINIT.
- Configuration of CAN\_MR: Choose a Time quantum to be  $t_{CAN} = 2/CORECLOCK$ , a propagation segment value to be  $t_{PRS} = t_{CAN} * 8$ , a synchronization jump width to be  $t_{SWJ} = t_{CAN} * 3$ , and the phase segments 1 and 2 to be  $t_{PHS1} = t_{PHS2} = t_{CAN} * 3$ .
- Enable the CAN by writing bit CANEN in CAN\_CR. An interrupt can be associated to know when the CAN is really enabled.
- Configuration of CAN\_IERX: In order to generate an interrupt at the end of the transmission, set the bit TXOK in CAN\_IER0 and in reception RXOK in CAN\_IER1; set the channel source 0 and 1 in CAN\_SIER. GIC must be configured.
- Fill the extended identifier to send in CAN\_IR0.
- Fill the extended identifier to receive in CAN\_IR1 and the mask on the identifier CAN\_MSK1 if users want to receive other identifiers.
- Fill in the four first data in CAN\_DRA0 and the fifth in CAN\_DRB0 to send.
- Set the length of byte (5) to receive field DLC, bit IDE for the extended identifier, and CHANEN to enable this channel in CAN\_CR1. An interrupt is generated if this CAN receives a frame with the identifier programmed.
- Set the length and then the transmit command by setting bit PCB, bit IDE, DLC filled with 5, and CHANEN to start the transmission in CAN\_CR0. An interrupt is generated after this transmission.

### 26.16.2 Interrupt Handling

- IRQ Entry and call C function.
- Read CAN\_SR and verify the source of the interrupt. If bit ISS is set, a channel generates an
  interrupt.
- Interrupt handling: If ISS is set in CAN\_SR, then we should look at the corresponding CAN\_SRx and treat the corresponding interrupt generated by the channel, then clear the interrupt flag by writing to the CAN\_CSRx.





# • IRQ Exit

# 26.17 Controller Area Network (CAN) Memory Map

Base Address CANO: 0xFFFD4000 (16 Channels)

Base Address CAN1: 0xFFFB8000 (16 Channels)

# Base Address CAN2: 0xFFFBC000 (32 Channels)

### Base Address CAN3: 0xFFFB0000 (16 Channels)

Table 26-6.CAN Memory Map

Offset	Register	Name	Access	Reset State
0x000 - 0x04C	Reserved	_	_	-
0x050	Enable Clock Register	CAN_ECR	Write -only	-
0x054	Disable Clock Register	CAN_DCR	Write-only	-
0x058	Power Management Status Register	CAN_PMSR	Read-only	0x00000000
0x05C	Reserved	_	_	-
0x060	Control Register	CAN_CR	Write-only	-
0x064	Mode Register	CAN_MR	Read/Write	0x00000000
0x068	Reserved	_	_	-
0x06C	Clear Status Register	CAN_CSR	Write-only	-
0x070	Status Register	CAN_SR	Read-only	0x0000002
0x074	Interrupt Enable Register	CAN_IER	Write-only	-
0x078	Interrupt Disable Register	CAN_IDR	Write-only	-
0x07C	Interrupt Mask Register	CAN_IMR	Read-only	0x00000000
0x080	Clear Interrupt Source Status Register	CAN_CISR	Write-only	-
0x084	Interrupt Source Status Register	CAN_ISSR	Read-only	0x00000000
0x088	Source Interrupt Enable Register	CAN_SIER	Write-only	-
0x08C	Source Interrupt Disable Register	CAN_SIDR	Write-only	-
0x090	Source Interrupt Mask Register	CAN_SIMR	Read-only	0x00000000
0x094 - 0x0FC	Reserved	_	_	-
0x100	Channel 0 Data Register A	CAN_DRA0	Read/Write	0x00000000
0x104	Channel 0 Data Register B	CAN_DRB0	Read/Write	0x00000000
0x108	Channel 0 Mask Register	CAN_MSK0	Read/Write	0x00000000
0x10C	Channel 0 Identifier Register	CAN_IR0	Read/Write	0x00000000
0x110	Channel 0 Control Register	CAN_CR0	Read/Write	0x00000000
0x114	Channel 0 Stamp Register	CAN_STP0	Read-only	0x00000000
0x118	Channel 0 Clear Status Register	CAN_CSR0	Write-only	-
0x11C	Channel 0 Status Register	CAN_SR0	Read-only	0x00000000
0x120	Channel 0 Interrupt Enable Register	CAN_IER0	Write-only	-
0x124	Channel 0 Interrupt Disable Register	CAN_IDR0	Write-only	-

# AT91SAM7A2

Offset	Register	Name	Access	Reset State
0x128	Channel 0 Interrupt Mask Register	CAN_IMR0	Read-only	0x00000000
0x12C - 0x13C	Reserved	_	_	_
0x140	Channel 1 Data Register A	CAN_DRA1	Read/Write	0x0000000
0x144	Channel 1 Data Register B	CAN_DRB1	Read/Write	0x0000000
0x148	Channel 1 Mask Register	CAN_MSK1	Read/Write	0x0000000
_	_	_	_	_
0x8E0	Channel 31 Interrupt Enable Register	CAN_IER31	Read/Write	0x0000000
0x8E4	Channel 31 Interrupt Disable Register	CAN_IDR31	Read/Write	0x0000000
0x8E8	Channel 31 Interrupt Mask Register	CAN_IMR31	Read/Write	0x0000000

# Table 26-6. CAN Memory Map (Continued)





# 26.18 CAN Enable Clock Register

Name:	CAN_ECR
Access:	Write-only
Base Address:	0x050

# 26.19 CAN Disable Clock Register

Name: Access: Base Address	CAN_D Write-or 0x054						
31	30	29	28	27	26	25	24
_	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	_	-	-	-	-	—	—
15	14	13	12	11	10	9	8
_	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	-	-	-	-	CAN	-

### • CAN: CAN Clock Status

0: CAN clock disabled.

1: CAN clock enabled.

Note: The CAN\_PMSR register is not reset by software reset.

# 26.20 CAN Power Management Status Register

Name: Access: Base Address:	CAN_P Read-or 0x058		-				
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	_	-	-	-
15	14	13	12	11	10	9	8
-	—	-	—	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	_	_	_	CAN	_

# • CAN: CAN Clock Status

0: CAN clock disabled.

1: CAN clock enabled.

Note: The CAN\_PMSR register is not reset by software reset.





# 26.21 CAN Control Register

Name: Access: Base Address:	CAN_C Write-or 0x060	R					
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	—	—	-	_	_	-
15	14	13	12	11	10	9	8
-	_	_	-	-	_	_	-
7	6	5	4	3	2	1	0
-	OVDIS	OVEN	ABDIS	ABEN	CANDIS	CANEN	SWRST

### • SWRST: CAN Software Reset

0: No effect.

1: Resets the CAN.

A software reset triggered hardware reset of the CAN is performed. It resets all the registers (except the CAN\_PMSR).

### • CANEN: CAN Enable

0: No effect.

1: Enables the CAN.

This enables the CAN to transfer and receive data.

A CAN bus synchronization delay is observed between the moment the CAN is enabled and the moment the CAN becomes an active node of the bus, ready to receive or transmit CAN frames. This synchronization delay is specified as part of the CAN standard and is equal to eleven nominal bit times. An additional internal delay between 0 and 1 bit times could be observed due to the re-synchronization of the enable command on the CAN bit-time clock.

### • CANDIS: CAN Disable

0: No effect.

1: Disables the CAN.

No data is received or transmitted.

In case a transfer is in progress, the transfer is finished before the CAN is disabled.

In case the CPU disables CAN while a transmit channel is enabled and ready to transmit, a spontaneous frame could be generated from that channel when re-enabling the CAN. To avoid spontaneous frame, CPU should disable transmit channels, and then wait for 'channel scan' before disabling the CAN module. The channel scan delay equals  $3 \times 32 = 192$  system clock periods for CAN2, and  $3 \times 16 = 96$  system clock periods for CAN3. Waiting for the channel scan ensures that no transmission is memorized by the CAN state machine. Another solution consists in resetting the CAN module (bit SWRST in CAN\_CR register) before re-enabling CAN.

In case both CANEN and CANDIS are equal to one when the control register is written, the CAN will be disabled.

### • ABEN: Abort Request Activate

0: No effect.

1: Activates the CAN abort request.

When set to 1, it masks CHANEN bits of the control register channels. In this mode, the CAN module still acknowledges messages on the bus, and its error counters (TEC and REC) continue to count.

### • ABDIS: Abort Request Deactivate

0: No effect.

1: Deactivates the CAN abort request.

In case both ABEN and ABDIS are equal to one when the control register is written, the CAN abort request is deactivated.

### OVEN: Overload Request Activate

0: No effect.

1: Activates the CAN overload request.

### • OVDIS: Overload Request Deactivate

0: No effect.

1: Deactivates the CAN overload request.

In case both OVEN and OVDIS are equal to one when the control register is written, the CAN overload request is deactivated.





# 26.22 CAN Mode Register

Name: Access: Base Address:	CAN_M Read/W 0x064							
31	30	29	28	27	26	25	24	
-	_	-	-	-	-	-	-	
23	22	21	20	19	18	17	16	
-		PHSEG2		-		PHSEG1		
15	14	13	12	11	10	9	8	
-	SMP	S	JW	-		PROP		
7	6	5	4	3	2	1	0	
-	_	BD						

### • BD[5:0]: Time Quantum Period

These bits are used to determine the time quantum t<sub>CAN</sub>:

$$t_{CAN} = \frac{(BD[5:0] + 1)}{CORECLK}$$

### • PROP[2:0]: Propagation Segment Value

This data reflects the physical delay within the network, including the signal propagation time and the chip internal delay.

$$t_{PBS} = t_{CAN} \times (PROP[2:0] + 1)$$

### • SJW[1:0]: Synchronization Jump Width

The CAN controller re-synchronizes on each edge of the transmission. The SJW value defines the maximum number of CAN clock cycles a bit period may be shortened or lengthened.

$$t_{SWJ} = t_{CAN} \times (SWJ[1:0] + 1)$$

Note: If the duration of segment PHASE\_SEG2 is greater or equal to 5 time quanta (PHSEG2[2:0] field in CAN\_MR greater or equal to 4), the resynchronization jump width must be configured at minimum to 2 time quanta (SJW[1:0] field in CAN\_MR register set to 1 at minimum).

### • SMP: Sampling Mode

0: The incoming stream is sampled once at sample point.

1: The incoming stream is sampled 3 times with a period of half a CAN clock period, centered at sample point.

### • PHSEG1[2:0]: Phase Segment 1 Value

This data is used to compensate edge phase errors. This segment can be shortened or lengthened by SJW.

$$t_{PHS1} = t_{CAN} \times (PHSEG1[2:0] + 1)$$

### • PHSEG2[2:0]: Phase Segment 2 Value

This data is used to compensate edge phase errors. This segment can be shortened or lengthened by SJW.

 $t_{PHS2} = t_{CAN} \times (PHSEG2[2:0] + 1)$ 

Note: Setting the duration of segment PHASE\_SEG2 to 1 time quanta is forbidden (PHSEG2[2:0] field in CAN\_MR must be set to 1 at minimum).

302 AT91SAM7A2

# 26.23 CAN Clear Status Register

Name: Access: Base Address:	CAN_C Write-or 0x06C						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	_	-	—	-
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	-	_	ENDINIT	—	-

# • ENDINIT: Clear End of CAN Initialization

0: No effect.

1: Clears end of CAN initialization interrupt.





# 26.24 CAN Status Register

Name: Access: Base Address:	CAN_SI Read-or : 0x070						
31	30	29	28	27	26	25	24
			T	EC			
23	22	21	20	19	18	17	16
			R	EC			
15	14	13	12	11	10	9	8
-	_	-	-	-	_	-	-
7	6	5	4	3	2	1	0
ISS	OVRQ	ABRQ	BUSOFF	ERPAS	ENDINIT	CANINIT	CANENA

### CANENA: CAN Enabled

0: CAN is disabled.

1: CAN is enabled.

No interrupt is generated on CAN enable or disable.

### • CANINIT: CAN Initialized

0: CAN is initialized.

1: CAN is in initialization phase.

During initialization phase, channel registers cannot be written. This bit does not generate an interrupt. The reset value is equal to 1, but after a short delay (8 times the number of Channel x CORECLOCK periods) corresponding to the initialization phase, this bit goes to 0.

### • ENDINIT: End of CAN Initialization

0: No end of CAN initialization.

1: End of CAN initialization phase.

### • ERPAS: Error Passive

0: No transition in error passive mode.

1: CAN enters in error passive mode.

### • BUSOFF: Bus Off

0: No transition in bus off mode.

1: CAN enters in bus off mode.

### ABRQ: CAN Abort Request

0: No CAN abort requested.

1: All the enabled channels are disabled. If this bit is set during communication, the transmission will end.

No interrupt is generated on CAN abort request activation or deactivation.

### OVRQ: Overload Frame Request

0: No overload frame requested.

1: A channel, programmed as a reception channel, follows its data of remote frame by an overload frame.

The purpose of the overload frame is to introduce a delay between received frames.

No interrupt is generated on CAN overload frame request activation or deactivation.

#### ISS: Interrupt Source Status

0: No interrupt in any channel.

1: At least one interrupt occurred in a channel (read CAN\_ISSR for more information).

#### • REC[7:0]: Reception Error Counter

Value of the reception error counter.

### • TEC[7:0]: Transmit Error Counter

Value of the transmit error counter.





# 26.25 CAN Interrupt Enable Register

Name:CAN\_IERAccess:Write-onlyBase Address:0x074

# 26.26 CAN Interrupt Disable Register

Name: Access: Base Address	CAN_ID Write-or : 0x078						
31	30	29	28	27	26	25	24
—	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
_	_	_	_	_	_	_	_
7	6	5	4	3	2	1	0
-	-	_	BUSOFF	ERPAS	ENDINIT	-	-

### • ENDINIT: End of CAN Initialization Mask

0: End of CAN initialization interrupt is disabled.

1: End of CAN initialization interrupt is enabled.

### • ERPAS: Error Passive Mask

0: Error passive interrupt is disabled.

1: Error passive interrupt is enabled.

# • BUSOFF: Bus Off Mask

0: Bus off interrupt is disabled.

1: Bus off interrupt is enabled.

# 26.27 CAN Interrupt Mask Register

Name: Access: Base Address:	CAN_IM Read-or 0x07C						
31	30	29	28	27	26	25	24
-	_	_	-	_	-	_	-
23	22	21	20	19	18	17	16
-	_	-	_	_	_	_	_
15	14	13	12	11	10	9	8
-	_	-	-	_	_	_	-
7	6	5	4	3	2	1	0
-	_	_	BUSOFF	ERPAS	ENDINIT	_	—

### • ENDINIT: End of CAN Initialization Mask

0: End of CAN initialization interrupt is disabled.

1: End of CAN initialization interrupt is enabled.

### • ERPAS: Error Passive Mask

0: Error passive interrupt is disabled.

1: Error passive interrupt is enabled.

### • BUSOFF: Bus Off Mask

0: Bus off interrupt is disabled.

1: Bus off interrupt is enabled.





# 26.28 CAN Clear Interrupt Source Status Register

Name: Access: Base Address:	CAN_CI Write-or 0x080						
31	30	29	28	27	26	25	24
CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24
23	22	21	20	19	18	17	16
CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16
15	14	13	12	11	10	9	8
CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

### • CHX: Channel X Interrupt Clear

0: No effect.

1: Clears interrupt for Channel X.

# 26.29 CAN Interrupt Source Status Register

Name: Access: Base Address:	CAN_IS Write-or 0x084		C				
31	30	29	28	27	26	25	24
CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24
23 CH23	22 CH22	21 CH21	20 CH20	19 CH19	18 CH18	17 CH17	16 CH16
01123	01122	GHZT	01120	CITI9	CITIO	GIII7	CITIO
15	14	13	12	11	10	9	8
CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

# • CHX: Channel X Interrupt

0: No interrupt occurred on Channel X.

1: An interrupt occurred on Channel X.

# 26.30 CAN Source Interrupt Enable Register

Name:	CAN_SIER
Access:	Write-only
Base Address:	0x088

# 26.31 CAN Source Interrupt Disable Register

Name: Access: Base Address:	CAN_SII Write-on 0x08C						
31	30	29	28	27	26	25	24
CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24
23	22	21	20	19	18	17	16
CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16
15	14	13	12	11	10	9	8
CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

### • CHX: Channel X Interrupt Mask

0: Channel X interrupt is disabled.

1: Channel X interrupt is enabled.





# 26.32 CAN Source Interrupt Mask Register

Name: Access: Base Address:	CAN_SI Read-or 0x090		0				
31	30	29	28	27	26	25	24
CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24
23 CH23	22 CH22	21 CH21	20 CH20	19 CH19	18 CH18	17 CH17	16 CH16
15	14	13	12	11	10	9	8
CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8
7 CH7	6 CH6	5 CH5	4 CH4	3 CH3	2 CH2	1 CH1	0 CH0

# • CHX: Channel X Interrupt Mask

0: Channel X interrupt is disabled.

1: Channel X interrupt is enabled.

# 26.33 CAN Channel Data Register A

Name: Access: Base Address:		RA0CAN_DR/					
31	30	29	28	27	26	25	24
			DA	TA3			
23	22	21	20	19	18	17	16
			DA	TA2			
15	14	13	12	11	10	9	8
			DA	TA1			
7	6	5	4	3	2	1	0
			DA	TA0			

• DATAy [y = 3..0]: Data y of Channel X

Data number y of Channel x.

# 26.34 CAN Channel Data Register B

Name: Access: Base Address:		RB0CAN_DR					
31	30	29	28	27	26	25	24
			DA	TA7			
23	22	21	20	19	18	17	16
			DA	TA6			
15	14	13	12	11	10	9	8
			DA	TA5			
7	6	5	4	3	2	1	0
			DA	TA4			

### • DATAy [y = 7..4]: Data y of Channel X

Data number y of Channel x.

# 26.35 CAN Channel Mask Register

20.00 OAN C		Skillegister					
Name: Access:	Read/W	SK0CAN_MS /rite	K31				
Base Address:	0xXX8						
31	30	29	28	27	26	25	24
MRTR	M	RB			MASK		
23	22	21	20	19	18	17	16
			MA	ASK			
15	14	13	12	11	10	9	8
			MASI	<b>K</b> [15:8]			
7	6	5	4	3	2	1	0
			MAS	K[7:0]			

# • MASK[28:0]: Identifier Mask of Channel X

29-bit mask for identifier.

If CAN operates with 11-bit identifier (standard frame), the upper bits MASK[28:11] are not used.

If CAN operates with 29-bit identifier (extended frame), the lower bits MASK[10:0] are used against the first received identifier bits and the upper bits MASK[28:11] are used with the last received identifier bits.

### • MRB[1:0]: Reserved Mask Bits

Masks the reserved bits. MRB[1] is only used in extended format.

### • MRTR: Remote Transmission Request Mask

Masks the RTR bit.





Name: Access: Base Address:	Read/W						
31	30	29	28	27	26	25	24
RTR	F	RB			ID		
23	22	21	20	19	18	17	16
			I	D			
15	14	13	12	11	10	9	8
				D			
7	6	5	4	3	2	1	0
			I	D			

# 26.36 CAN Channel Identifier Register

# • ID[28:0]: Identifier of Channel X

29-bit value for identifier.

For channels configured in standard frame mode (IDE bit reset to '0' in CAN\_CRx), upper bits ID[28:11] are not used. In case of a transmission, only the base identifier bits ID[10:0] are sent from ID10 to ID0. In case of a reception, for acceptance filtering, only the base identifier ID[10:0] (together with MASK[10:0] bits from CAN\_MSKx register) are checked against the received identifier.

For channels configured in extended frame mode (IDE bit set to '1' in CAN\_CRx), upper bits ID[28:11] are used for the extended identifier and lower bits ID[10:0] are used for the base identifier. In case of a transmission, the base identifier ID[10:0] are sent first from ID10 to ID0, and extended identifier ID[28:11] are sent later from ID28 to ID11. And in case of a reception, for acceptance filtering, base identifier ID[10:0] (together with MASK[10:0] bits from CAN\_MSKx register) are checked against first received identifier bits, and the extended identifier ID[28:11] (together with MASK[28:11] bits from CAN\_MSKx register) are checked against last received identifier bits.

Frame Format	Identifier Length	Base ID	Extended ID
Standard frame	11 bits	ID[10:0]	N/A
Extended frame	29 bits	ID[10:0]	ID[28:11]

### • RB[1:0]: Reserved Bits

These bits are sent to the control field. RB[1] generates the IDE bit in the standard frame format and the r1 bit in the extended frame format, respectively. RB[0] generates the r0 bit in the standard and extended frame formats. In the CAN 2.0A specification, RB[1] and RB[0] must be set dominant (i.e., to logical 0).

### • RTR: Remote Transmission Request

In data frames, the RTR bit has to be dominant. Within a remote frame, the RTR bit has to be recessive.

26.37 CAN Ch	annel Control Register	
Name:	CAN_CR0CAN_CR31	

Name: Access: Base Address:	CAN_C Read/W 0xXX0	roCAN_CR3 /rite	1				
31	30	29	28	27	26	25	24
-	_	_	-	_	_	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	-	-	_	_	—	OVERWRITE
7	6	5	4	3	2	1	0
CHANEN	PCB	RPLYV	IDE		DLC	[3:0]	

# • DLC[3:0]: Data Length Code

This is the number of bytes in the data field of a message (from 0 to 8). This value is updated whenever a frame is received (data or remote). If the incoming DLC differs from the expected one, a warning is issued in the status register of the channel (CAN\_SRX).

### • IDE: Extended Identifier Flag

0: Identifier is 11 bits long (CAN rev 2.0A).

1: Identifier is 29 bits long (CAN rev 2.0B).

# • RPLYV: Automatic Reply

0: No effect.

1: Channel x makes an automatic reply after receiving a remote frame.

### • PCB: Channel Producer

0: Channel x is consumer.

1: Channel x is producer.

Bit PCB resets to 0 after a transmission.

### • CHANEN: Channel Enable

0: Channel x disabled.

1: Channel x enabled.

Note: For detailed information, refer to Section 26.15 "Using the CAN Module" on page 294.

# • OVERWRITE: Channel Overwrite Mode

- 0: Channel x in normal mode.
- 1: Channel x in overwrite mode.

In overwrite mode, the channel is not disabled after each reception. New frames overwrite the previous frame.





# 26.38 CAN Channel Stamp Register

Name: Access: Base Address:	CAN_S Read-or 0xXX4	TP0CAN_STF nly	231				
31	30	29	28	27	26	25	24
			STAMF	2[31:24]			
23	22	21	20	19	18	17	16
			STAMF	2[23:16]			
15	14	13	12	11	10	9	8
			STAMI	P[15:8]			
7	6	5	4	3	2	1	0
			STAM	P[7:0]			

### • STAMP[31:0]: Stamp Value

These 32 bits stamp the date at which the message linked with Channel x has been emitted/received (depending on the producer/consumer bit). The value is copied from the WT second counter register.

# 26.39 CAN Channel Clear Status Register

Name: Access: Base Address:		CAN_CSR0CAN_CSR31 Write-only 0xXX8						
31	30	29	28	27	26	25	24	
-	_	-	-	-	-	-	-	
23	22	21	20	19	18	17	16	
-	_	_	-	_	_	_	_	
15	14	13	12	11	10	9	8	
-	_	-	_	OVRUN	FILLED	DLCW	-	
7	6	5	4	3	2	1	0	
RFRAME	TXOK	RXOK	BUS	STUFF	CRC	FRAME	ACK	

### • ACK: Acknowledge Error Clear

0: No effect.

1: Clears acknowledge error interrupt.

### • FRAME: Frame Error Clear

0: No effect.

1: Clears frame error interrupt.

### • CRC: CRC Error Clear

0: No effect.

1: Clears CRC error interrupt.

### • STUFF: Stuffing Error Clear

0: No effect.

1: Clears stuffing error interrupt.

### • BUS: Bus Error Clear

0: No effect.

1: Clear bus error interrupt.

### RXOK: Reception Completed Clear

0: No effect.

1: Clears reception completed interrupt.

### • TXOK: Transmission Completed Clear

0: No effect.

1: Clears transmission completed interrupt.

### • RFRAME: Remote Frame Clear

0: No effect.





1: Clears remote frame interrupt.

# • DLCW: DLC Warning Clear

0: No effect.

1: Clears DLC warning.

# • FILLED: Filled Flag Clear

0: No effect.

1: Clears the FILLED flag.

# • OVRUN: Overrun Flag Clear

0: No effect.

1: Clears the OVERRUN flag.

# 26.40 CAN Channel Status Register

Name: Access: Base Address:	Read-or	CAN_SR0CAN_SR31 Read-only 0xXXC						
31	30	29	28	27	26	25	24	
-	_	-	-	-	-	-	-	
23	22	21	20	19	18	17	16	
	_	_	_	_	_	_	_	
15	14	13	12	11	10	9	8	
-	_	-	-	OVRUN	FILLED	DLCW	-	
7	6	5	4	3	2	1	0	
RFRAME	TXOK	RXOK	BUS	STUFF	CRC	FRAME	ACK	

### • ACK: Acknowledge Error

0: No acknowledge error during last transmission.

1: An acknowledge error occurred during the last transmission. There was not a dominant bit during the ACK slot.

### • FRAME: Frame Error

0: No frame error during last communication.

1: A frame error occurred during last communication. A fixed form bit contained one or more illegal bits.

### CRC: CRC Error

0: No CRC error during last reception.

1: A CRC error has been detected during the last reception. Received CRC sequence is not equal to the calculated one.

### • STUFF: Stuffing Error

0: No stuffing error during the last communication.

1: A stuffing error occurred during the last communication. At least 6 consecutive equal bits have been detected.

### • BUS: Bus Error

0: No bus error during last transmission.

1: A bus error occurred during last transmission. The transmitter sent a dominant bit but a recessive bit was detected on the network.

### • RXOK: Reception Completed

0: No new reception completed.

1: A reception was completed without any error.

### • TXOK: Transmission Completed

- 0: No new transmission completed.
- 1: A transmission was completed without any error.
- RFRAME: Remote Frame





0: No remote frame received or sent since last clear of RFRAME bit.

1: A remote frame has been received or sent since last clear of RFRAME bit.

# • DLCW: DLC Warning

0: No DLC warning.

1: DLC warning. Last message accepted with a different DLC than programmed in the CAN channel control register (CAN\_CRx).

This bit does not generate an interrupt.

### • FILLED: Reception Buffer Filled

0: No frame has been received since last clear of FILLED bit.

1: The buffers CAN\_DRA and CAN\_DRB not read since last frame has been received.

This flag can be enabled only when OVERWRITE mode is selected.

### • OVRUN: Overrun

0: No frame has been received while FILLED flag is set to logical 1.

1: A frame has been received while FILLED flag is set to logical 1.

This flag can be raised only when OVERWRITE mode is selected.

This bit does not generate an interrupt.

# 26.41 CAN Channel Interrupt Enable Register

Name:	CAN_IER0CAN_IER31
Access:	Write-only
Base Address:	0xXX0

# 26.42 CAN Channel Interrupt Disable Register

Name: Access: Base Address:	Write-or	CAN_IDR0CAN_IDR31 Write-only 0xXX4						
31	30	29	28	27	26	25	24	
-	_	—	-	-	-	-	-	
23	22	21	20	19	18	17	16	
-	_	_	-	_	_	_	_	
15	14	13	12	11	10	9	8	
-	-	_	-	_	_	_	_	
7	6	5	4	3	2	1	0	
RFRAME	ТХОК	RXOK	BUS	STUFF	CRC	FRAME	ACK	

### • ACK: Acknowledge Error Mask

0: Acknowledge error interrupt is disabled.

1: Acknowledge error interrupt is enabled.

### • FRAME: Frame Error Mask

- 0: Frame error interrupt is disabled.
- 1: Frame error interrupt is enabled.

### • CRC: CRC Error Mask

- 0: CRC error interrupt is disabled.
- 1: CRC error interrupt is enabled.

### • STUFF: Stuffing Error Mask

- 0: Stuffing error interrupt is disabled.
- 1: Stuffing error interrupt is enabled.

### • BUS: Bus Error Mask

- 0: Bus error interrupt is disabled.
- 1: Bus error interrupt is enabled.

### RXOK: Reception Completed Mask

- 0: Completed reception interrupt is disabled.
- 1: Completed reception interrupt is enabled.





# 27. General-purpose Timer (GPT)

# 27.1 Overview

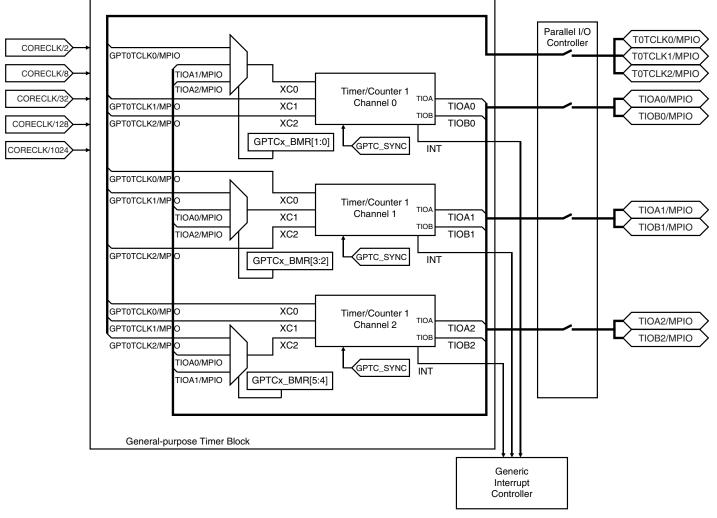
The AT91SAM7A2 has four independent general-purpose timer blocks. Three timers are grouped in the GPT0 module and can be cascaded; the fourth timer is the GPT1 module.

Each timer has a 16-bit Timer/counter channel, a Power Management Controller and a Parallel I/O Controller. Each channel can be independently programmed using the two operating modes (capture mode or waveform mode) to perform a range of functions including frequency measurement, event counting, interval measurement, pulse generation, delay timing, pulse width modulation and interrupt generation.

After a hardware reset, the timer pins are set as general-purpose I/O (configured as input), the interrupts are disabled in the interrupt controller and the Timer Controller Clock and the PIO Controller Clock are disabled.

# 27.2 Block Diagram





Each channel, shown in its configuration in Figure 27-1, has the following components:

- one 16-bit counter
- one 16-bit compare register (RC)
- two 16-bit capture/compare registers, RA and RB
- one multiplexer allowing the selection of eight clocks: five internal and three external (common to all channels). It is possible to combine two clocks to generate a burst clock. Each external clock can be used as external trigger source.
- an internal interrupt signal that can be programmed to generate processor interrupts via the Generic Interrupt Controller (GIC module). They are generated when one of the following events occurs:
  - Counter overflow
  - Load Register (A or B)
  - Equal Compare Register (A or B or C)
  - External edge detection
  - Overrun
- · one software trigger
- one software reset that resets the channel and its associated registers (except Power Management registers)
- three parallel I/O pins that can be dedicated for multiple counter functions (Operation mode dependent) or in PIO mode.
  - TIOA, in capture mode, is an event input to load register A or register B or an input trigger. In waveform mode, it is an output to generate a waveform.
  - TIOB, in capture mode, is an input trigger. In waveform mode, it is either an output to generate a waveform (dual waveform mode) or an input trigger (single waveform mode).
  - TCLK, in capture mode and in waveform mode, is an external clock input.
- the synchronization bit that allows the synchronization of the three channels by generating a software trigger simultaneously on the three channels.
- the reset bit of the Block Control Register that resets the three channels simultaneously.

It is possible to chain two or three channels to increase the counter capacity (see "Timer Controller Block Programming" on page 376).

# 27.3 Pin Description

Table 27-1 shows the internal pin configurations in the different operating modes.

Pin (x = channel)	x = channel) Capture Mode		Dual Waveform Mode	
TIOAx	Input	Output	Output	
HOAX	Capture or Trigger	Waveform	Waveform	
TIODY	Input	Input	Output	
TIOBx	Trigger	Trigger	Waveform	
TCLKx	Input	Input	Input	
TOLKX	External Clock	External Clock	External Clock	

Table 27-1. Pin Definition



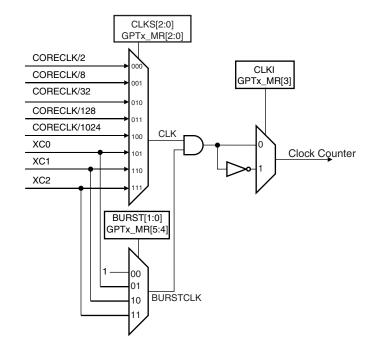


# 27.4 Clock Sources

The timer controller can use one of eight different clocks. The CLKS[2:0] bits of the mode registers GPTx\_MR determine whether the counter is clocked by one of the five internal clock sources generated in the prescalar block and derived from CORECLK or one of the three external clock sources (TCLKx).

Figure 27-2 shows the clock selection block.

### Figure 27-2. Clock Selection Block



The counter clock sources can be any of the following:

- External event input XCx
- One of 5 internal clocks (CORECLK/2, CORECLK/8, CORECLK/32, CORECLK/128, CORECLK/1024)
- A burst clock (see "Burst Clock" on page 323).

The maximal count duration when an internal clock is used is determined by the internal clock MCK and the prescale number.

Maximal Count Duration (seconds) =  $2^{16}$ /CLK where CLK is in Hz.

Counter Resolution = 1/CLK

### 27.4.1 External Clock

When an external clock source is used, the 16-bit counter can be programmed as a 16-bit event counter. An external transition (rising or falling following the state of the bit CLKI of the Mode register) increments the counter.

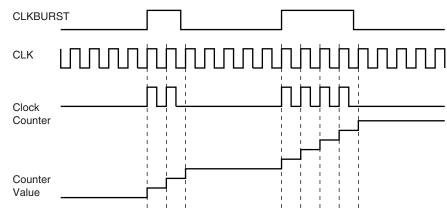
If an external clock is used, make sure that each of its pulse has a duration strictly higher than the CORECLK period.

### 27.4.2 Burst Clock

If the field BURST of the mode register selects an external clock, this clock is combined with CLK through a logical AND.

Thus the considered timer channel is clocked by CLK only when CLKBURST is high as shown in Figure 27-3.





### 27.5 16-bit Counter

The 16-bit counter is a free-running counter clocked by eight sources: 5 internal and 3 external.

The program can access the counter value in real-time in read-only access with the counter value register GPT\_CV.

When counter reset occurs, the counter is loaded with 0x0000 and begins its count if its clock is enabled (the clock is disabled or enabled by CLKDIS and CLKEN of the control register GPT\_CR).

When the maximal value is reached (0xFFFF), the counter rolls over to a count of 0x0000, sets an overflow flag (the bit COVFS of the status register GPT\_SR), may generate an interrupt (enabled by the bit COVFS of the interrupt enable register GPT\_IER and disabled by the bit COVFS of the interrupt disable register GPT\_IDR), and continues to count up.

### 27.5.1 Counter Reset

During counting, the counter can be reset to 0x0000 following:

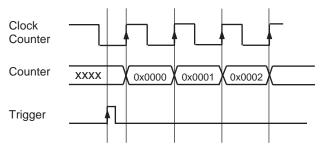
- a software Trigger
- an external Trigger
- an equality on Compare C
- the synchronous bit TCSYNC of the block control register GPT\_BCR.

Each time it is reset, the counter passes to 0x0000 at the next valid counter clock edge. See Figure 27-4.





# Figure 27-4. Counter Reset Diagram



# 27.6 16-bit Registers

Each channel contains three 16-bit registers.

The mode determines whether the capture/compare registers are used as capture registers or compare registers.

In capture mode, registers A and B are capture registers and can be loaded by TIOAx edges.

In waveform mode, registers A and B are compare registers.

Register C is always a compare register. It can generate a counter reset when the programmed value is reached.

RA, RB and RC compare registers can generate a waveform when their counters reach the programmed values.

In an application, the required compare register values must be calculated using the following equation:

CompareValue =  $(t \times CLK) - 1$ 

where:

t = desired timer compare period (in seconds)

CLK = counter clock (in Hertz)

### 27.6.1 Example

To determine the value needed in a compare register to obtain an equality with the counter after 0.1 second with CORECLK = 30 MHz:

 Determine the minimal prescale value by dividing CORECLK by the maximal counter value 0xFFFF (65,535) to know the divisor factor: 30,000,000

$$\frac{50, 000, 000}{65, 535} = 457.77$$

The value of the divider greater than or equal to 457.77 is DIV<sub>min</sub> = 1024.

CLK must therefore be at least CORECLK/1024 (29.3 kHz) in order to obtain an equal condition after 0.1 seconds.

2. Calculate the register value with this clock using the following equation:  $CORECLK = 0.1 \times \frac{CORECLK}{1024} - 1 = 0.1 \times \frac{30, 000, 000}{1024} - 1 = 2928.65$  Rounding value to 2929 (0x0B71) generates a 0.01% error.

## 27.7 External Edge Detection

The timer contains many external edge detection options.

The operating mode determines their number:

- Capture mode
  - TIOAx as load register and external trigger
  - TIOBx as external trigger
- Waveform mode: Dual waveform mode

- XC0, XC1 or XC2 as external trigger

- Waveform mode: Single waveform mode
  - TIOBx as external trigger

For each edge detection, it is possible to choose a rising edge, a falling edge or both.

Whereas when a reset is caused, it occurs at the next valid counter clock edge.

If an external trigger is used, each of its pulses must have a duration strictly greater than the CORECLK period.

## 27.8 Interrupts

Each timer contains a total of eight timer interrupts and three PIO interrupts. They can be enabled or disabled from the GPT\_IER and GPT\_IDR. The programming mode determines which interrupts are available in Table 27-2.

Interrupt Name	Capture Mode	Waveform Mode
Counter Overflow Interrupt COVFS	X	Х
Load Overrun Interrupt LOVRS	X	
Compare Register A Interrupt CPAS		Х
Compare Register B Interrupt CPBS		Х
Compare Register C Interrupt CPCS	X	Х
Load Capture Register A Interrupt LDRAS	x	
Load Capture Register B Interrupt LDRBS	x	
External Trigger Interrupt ETRGS	x	х
TCLK/MPIO Interrupt TCLKS	x	Х
TIOA/MPIO Interrupt TIOAS	X	Х
TIOB/MPIO Interrupt TIOBS	Х	Х

Table 27-2.	Available Interrupts
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#### 27.9 PIO Controller

Each timer channel has three programmable I/O lines. These I/O lines are multiplexed with signals (TIOA, TIOB, TCLK) of the timer channel to optimize the use of available package pins. These lines are controlled by the timer channel PIO controller.





# 27.10 Power Management

Each timer channel (GPT0, GPT1 and GPT2) is provided with a power management block allowing optimization of power consumption (see "Power Management Block" on page 25).

# 27.11 Example Of Use

This section gives an example of use of the General-purpose Timer, explaining how to generate a tick (usually used in RTOS) of 10 ms then in the interrupt. The timer is restarted. Core clock is 30MHz

#### 27.11.1 Configuration

- Enable the clock on GPT peripheral by writing bit TC in GPT\_ECR.
- Do a software reset of the GPT peripheral to be in a known state by writing bit SWRST in GPT\_CR.
- Configuration of GPT\_MR: Choose a clock divider of 1024, Wave mode selected, CPCTRG is selected when counter is equal to RC register, then it causes a trigger on the counter (restart to 0), CPCSTOP allows the counter to stop when equal to RC.
- Configuration of GPT\_RC: This sets the period to 10 ms. Period = CORECLK/(GPT CLOCK divider\*tick frequency) giving 30 MHz/(1024\*100).
- Configuration of GPT\_IER: To generate an interrupt when the counter is equal to the RC register value, write bit CPCS in GPT\_IER. GIC must be configured.
- Start the timer by writing bit CLKEN and SWTRG in GPT\_CR. After 10 ms, the interrupt is generated.

#### 27.11.2 Interrupt Handling

- IRQ Entry and call C function.
- Read GPT\_SR and verify the source of the interrupt. This register is read and clear, care should be taken to maintain status information to be able to proceed in all cases.
- Interrupt handling: Restart the timer by writing bit CLKEN and SWTRG in GPT\_CR. After 10 ms, another interrupt is generated.
- IRQ Exit.

# 27.12 General-purpose Timer (GPT) Memory Map

Base Address GPT0 Channel 0: 0xFFFC8000

Base Address GPT0 Channel 1: 0xFFFC8100

Base Address GPT0 Channel 2: 0xFFFC8200

Base Address GPT1: 0xFFFCC000

# Table 27-3. GPT Memory Map and Control Registers

Offset	Register	Name	Access	Reset State
0x00	PIO Enable Register	GPT_PER	Write-only	-
0x04	PIO Disable Register	GPT_PDR	Write-only	-
0x08	PIO Status Register	GPT_PSR	Read-only	0x00070000
0x0C	Reserved	-	_	-
0x10	PIO Output Enable Register	GPT_OER	Write-only	-
0x14	PIO Output Disable Register	GPT_ODR	Write-only	-
0x18	PIO Output Status Register	GPT_OSR	Read-only	0x00000000
0x1C - 0x2C	Reserved	-	_	_
0x30	PIO Set Output Data Register	GPT_SODR	Write-only	-
0x34	PIO Clear Output Data Register	GPT_CODR	Write-only	_
0x38	PIO Output Data Status Register	GPT_ODSR	Read-only	0x0000000
0x3C	PIO Pin Data Status Register	GPT_PDSR	Read-only	0x000X0000
0x40	PIO Multi-Driver Enable Register	GPT_MDER	Write-only	_
0x44	PIO Multi-Driver Disable Register	GPT_MDDR	Write-only	_
0x48	PIO Multi-Driver Status Register	GPT_MDSR	Read-only	0x0000000
0x4C	Reserved	-	_	_
0x50	Enable Clock Register	GPT_ECR	Write-only	_
0x54	Disable Clock Register	GPT_DCR	Write-only	-
0x58	Power Management Status Register	GPT_PMSR	Read-only	0x00000000
0x5C	Reserved	-	_	-
0x60	Control Register	GPT_CR	Write-only	-
0x64	Mode Register	GPT_MR	Read/Write	0x00000000
0x68	Reserved	-	-	-
0x6C	Reserved	-	_	-
0x70	Status Register	GPT_SR	Read-only	0x00000X00
0x74	Interrupt Enable Register	GPT_IER	Write-only	-
0x78	Interrupt Disable Register	GPT_IDR	Write-only	-
0x7C	Interrupt Mask Register	GPT_IMR	Read-only	0x0000000
0x80	Counter Value	GPT_CV	Read-only	0x0000000
0x84	Capture - Compare Register A	GPT_RA	Read/Write	0x00000000





Table 27-3.	GPT Memory Map and Control Registers	(Continued)
-------------	--------------------------------------	-------------

Offset	Register	Name	Access	Reset State			
0x88	Capture - Compare Register B	GPT_RB	Read/Write	0x0000000			
0x8C	Compare Register C	GPT_RC	Read/Write	0x0000000			
Control and Test Registers							
0x0300	Block Control Register	GPT_BCR	Write-only	_			
0x0304	Block Mode Register	GPT_BMR	Read/Write	0x0000000			

# 27.13 GPT PIO Enable Register

Name:	GPT_PER
Access:	Write-only
Base Address:	0x00

# 27.14 GPT PIO Disable Register

Name: Access: Base Address:	GPT_PI Write-or 0x04						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	—	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
-	_	_	-	_	-	_	_
7	6	5	4	3	2	1	0
-	_	-	-	-	-	-	_

## • TIOB: TIOB Pin

0: PIO is inactive on the TIOBx pin (Timer Controller is active).

1: PIO is active on the TIOBx pin (Timer Controller is inactive).

## • TIOA: TIOA Pin

0: PIO is inactive on the TIOAx pin (Timer Controller is active).

1: PIO is active on the TIOAx pin (Timer Controller is inactive).

# • TCLK: TCLK Pin

0: PIO is inactive on the TCLKx pin (Timer Controller is active).

1: PIO is active on the TCLKx pin (Timer Controller is inactive).





# 27.15 GPT PIO Status Register

Name: Access: Base Address:	GPT_P3 Read-or 0x08						
31	30	29	28	27	26	25	24
-	_	-	-	_	-	-	-
23	22	21	20	19	18	17	16
_	_	_	-	-	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
-	—	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	_	_	-	-	-	-	_

## • TIOB: TIOB Pin

0: PIO is inactive on the TIOBx pin (Timer Controller is active).

1: PIO is active on the TIOBx pin (Timer Controller is inactive).

## • TIOA: TIOA Pin

0: PIO is inactive on the TIOAx pin (Timer Controller is active).

1: PIO is active on the TIOAx pin (Timer Controller is inactive).

#### • TCLK: TCLK Pin

0: PIO is inactive on the TCLKx pin (Timer Controller is active).

1: PIO is active on the TCLKx pin (Timer Controller is inactive).

# 27.16 GPT PIO Output Enable Register

Name:	GPT_OER
Access:	Write-only
Base Address:	0x10

# 27.17 GPT PIO Output Disable Register

Name: Access: Base Address	GPT_O Write-or : 0x14						
31	30	29	28	27	26	25	24
—	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
_	_	_	-	_	_	-	-
7	6	5	4	3	2	1	0
_	-	_	-	-	-	-	-

## • TIOB: TIOB Pin

0: The TIOBx PIO pin is input.

1: The TIOBx PIO pin is output

## • TIOA: TIOA Pin

0: The TIOAx PIO pin is input.

1: The TIOAx PIO pin is output

## • TCLK: TCLK Pin

0: The TCLKx PIO pin is input.

1: The TCLKx PIO pin is output.





# 27.18 GPT PIO Output Status Register

Name: Access: Base Address:	GPT_O Read-or 0x18						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	-	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
-	_	—	-	-	-	-	-
7	6	5	4	3	2	1	0
_	_	_	-	-	-	_	-

## • TIOB: TIOB Pin

0: The TIOBx PIO pin is input.

1: The TIOBx PIO pin is output

## • TIOA: TIOA Pin

0: The TIOAx PIO pin is input.

1: The TIOAx PIO pin is output

## • TCLK: TCLK Pin

0: The TCLKx PIO pin is input.

1: The TCLKx PIO pin is output.

# 27.19 GPT PIO Set Output Data Register

Name:GPT\_SODRAccess:Write-onlyBase Address:0x30

# 27.20 GPT PIO Clear Output Data Register

Name: Access: Base Address	GPT_C0 Write-or : 0x34						
31	30	29	28	27	26	25	24
—	-	—	-	-	_	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
_	_	_	_	_	-	-	-
7	6	5	4	3	2	1	0
-	-	_	-	-	-	-	-

## • TIOB: TIOB Pin

0: The output data for the TIOBx is programmed to 0.

1: The output data for the TIOBx is programmed to 1.

## • TIOA: TIOA Pin

0: The output data for the TIOAx is programmed to 0.

1: The output data for the TIOAx is programmed to 1.

# • TCLK: TCLK Pin

0: The output data for the TCLKx is programmed to 0.

1: The output data for the TCLKx is programmed to 1.





# 27.21 GPT PIO Output Data Status Register

Name: Access: Base Address:	GPT_O Read-oi 0x38						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	_	-	TCLK	TIOA	TIOB
15	14	13	12	11	10	9	8
-	_	-	—	-	-	-	-
7	6	5	4	3	2	1	0
-	_	-	—	-	-	-	-

#### • TIOB: TIOB Pin

0: The output data for the TIOBx is programmed to 0.

1: The output data for the TIOBx is programmed to 1.

## • TIOA: TIOA Pin

0: The output data for the TIOAx is programmed to 0.

1: The output data for the TIOAx is programmed to 1.

#### • TCLK: TCLK Pin

0: The output data for the TCLKx is programmed to 0.

1: The output data for the TCLKx is programmed to 1.

# 27.22 GPT PIO Pin Data Status Register

Name: Access: Base Address:	Read-o	GPT_PDSR Read-only 0x3C							
31	30	29	28	27	26	25	24		
-	-	-	-	-	-	-	-		
23	22	21	20	19	18	17	16		
-	_	-	-	_	TCLK	TIOA	TIOB		
15	14	13	12	. 11	10	9	8		
-	_	-	-	-	-	-	-		
7	6	5	4	3	2	1	0		
_	—	_	-	-	-	-	_		

## • TIOB: TIOB Pin

0: The pin TIOBx is at logic 0.

1: The pin TIOBx is at logic 1.

# • TIOA: TIOA Pin

0: The pin TIOAx is at logic 0.

1: The pin TIOAx is at logic 1.

## • TCLK: TCLK Pin

0: The pin TCLKx is at logic 0.

1: The pin TCLKx is at logic 1.





# 27.23 GPT PIO Multi-driver Enable Register

Name:GPT\_MDERAccess:Write-onlyBase Address:0x40

# 27.24 GPT PIO Multi-driver Disable Register

Name: Access: Base Address:	Write-or	GPT_MDDR Write-only 0x44						
31	30	29	28	27	26	25	24	
-	_	-	-	-	-	-	-	
23	22	21	20	19	18	17	16	
-	_	-	-	-	TCLK	TIOA	TIOB	
15	14	13	12	11	10	9	8	
_	_	_	_	_	_	-	-	
7	6	5	4	3	2	1	0	
-	—	_	-	-	-	-	-	

## • TIOB: TIOB Pin

0: TIOBx pin is not configured as an open drain.

1: TIOBx pin is configured as an open drain.

## • TIOA: TIOA Pin

0: TIOAx pin is not configured as an open drain.

1: TIOAx pin is configured as an open drain.

# • TCLK: TCLK Pin

0: TCLKx pin is not configured as an open drain.

1: TCLKx pin is configured as an open drain.

# 27.25 GPT PIO Multi-driver Status Register

Name: Access: Base Address:	Read-or	GPT_MDSR Read-only 0x48							
31	30	29	28	27	26	25	24		
-	_	-	-	-	-	-	-		
23	22	21	20	19	18	17	16		
-	_	-	-	-	TCLK	TIOA	TIOB		
15	14	13	12	11	10	9	8		
-	—	—	-	-	-	-	-		
7	6	5	4	3	2	1	0		
_	_	—	-	—	—	—	-		

## • TIOB: TIOB Pin

0: TIOBx pin is not configured as an open drain.

1: TIOBx pin is configured as an open drain.

## • TIOA: TIOA Pin

0: TIOAx pin is not configured as an open drain.

1: TIOAx pin is configured as an open drain.

## • TCLK: TCLK Pin

0: TCLKx pin is not configured as an open drain.

1: TCLKx pin is configured as an open drain.





# 27.26 GPT Enable Clock Register

Name:	GPT_ECR
Access:	Write-only
Base Address:	0x50

# 27.27 GPT Disable Clock Register

Name: Access: Base Address:	GPT_D0 Write-or 0x54						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	_	-	_	-	—	—
15	14	13	12	11	10	9	8
-	-	—	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	_	-	_	_	TC	PIO

## • PIO: PIO Clock

1: PIO controller clock is enabled.

0: PIO controller clock is disabled.

## • TC: General-purpose Timer Clock

1: General-purpose Timer channel and clock divider clock enabled.

0: General-purpose Timer channel and clock divider clock disabled.

# 27.28 GPT Power Management Status Register

Name: Access: Base Address:	GPT_P Read-o	MSR	- J				
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
_	_	-	-	-	—	_	_
15	14	13	12	. 11	10	9	8
-	—	-	-	-	-	-	—
7	6	5	4	3	2	1	0
-	_	_	—	_	—	TC	PIO

## • PIO: PIO Clock

1: PIO controller clock is enabled.

0: PIO controller clock is disabled.

## • TC: General-purpose Timer Clock

1: General-purpose Timer channel and clock divider clock enabled.

0: General-purpose Timer channel and clock divider clock disabled.





# 27.29 General-purpose Timer in Capture Mode

#### 27.29.1 Description

The capture (wave measurement) mode is entered by setting WAVE (bit [15] in the Mode Register) to 0.

It is the default operating mode after a hardware reset. It forces TIOAx and TIOBx pins as input pins.

The capture mode provides the possibility to determine the duration between two events. An event may either be an external input signal on TIOAx or TIOBx or an internal event (software trigger or equality between the counter and a predefined compare value). An external event (rising or falling edge) on TIOAx can result in capture register A being loaded, capture register B being loaded or a trigger effect (reset and start the counter).

A predefined compare value (16-bit) can be set in the compare register C.

When the capture register B is loaded, it can disable the counter clock and/or stop the counter.

The user may choose an internal clock source (CORECLK/2, CORECLK/8, CORECLK/32, CORECLK/128 or CORECLK/1024) or an external clock (TCLK0, TCLK1 or TCLK2).

A burst mode is available. It generates a burst clock. For more details, refer to "Clock Sources" on page 322.

Six interrupts can be produced:

- External trigger detected
- RA loaded
- RB loaded
- Counter overflow (when the counter passes from 0xFFFF to 0x0000)
- Overrun (when RA or RB is reloaded before the old value is read)
- Compare RC (the counter reaches the value stored in register C)

Finally, the synchronize register can be used to cause a software trigger for reset and start the counter at the next valid counter clock edge on all channels at the same time.

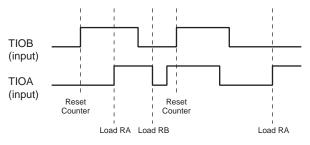
Figure 27-5 to Figure 27-8 show different applications using the capture mode.

For more details, refer to application notes.

#### 27.29.1.1 Measure TIOA Pulse and Phase between TIOB and TIOA

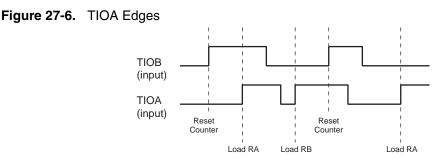
A TIOBx rising edge resets and starts the counter. A rising TIOAx edge loads RA and a falling TIOAx edge loads RB. Once RB is loaded, a trigger restarts a capture cycle. RA contains the phase between TIOBx and TIOAx. (RB - RA) is the duration of the TIOAx pulse.

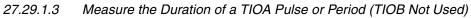
Figure 27-5. TIOA Pulse



27.29.1.2 Measure Duration between Two Successive Rising TIOA Edges

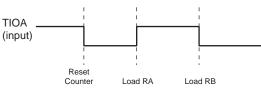
A TIOBx rising edge resets and starts the counter. The first rising TIOAx edge after the reset loads RA and a second loads RB. RA contains the phase between TIOBx and TIOAx. (RB-RA) is the period of the TIOAx pulse.





A TIOAx falling edge resets, starts the counter and loads RB if RA is already loaded. A TIOAx rising edge loads RA. RA contains the duration of a TIOAx pulse (low level). RB contains the duration of the TIOAx period.

Figure 27-7. TIOA Pulse or Period

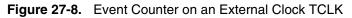


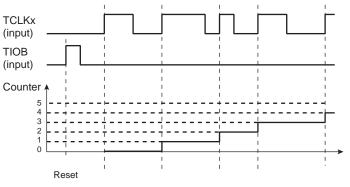
27.29.1.4 Event Counter on an External Clock TCLK

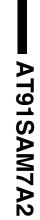
The counter is incremented with each TCLKx rising edge. This application can be generated in waveform mode. The counter value contains the number of detected TCLKx rising edges.





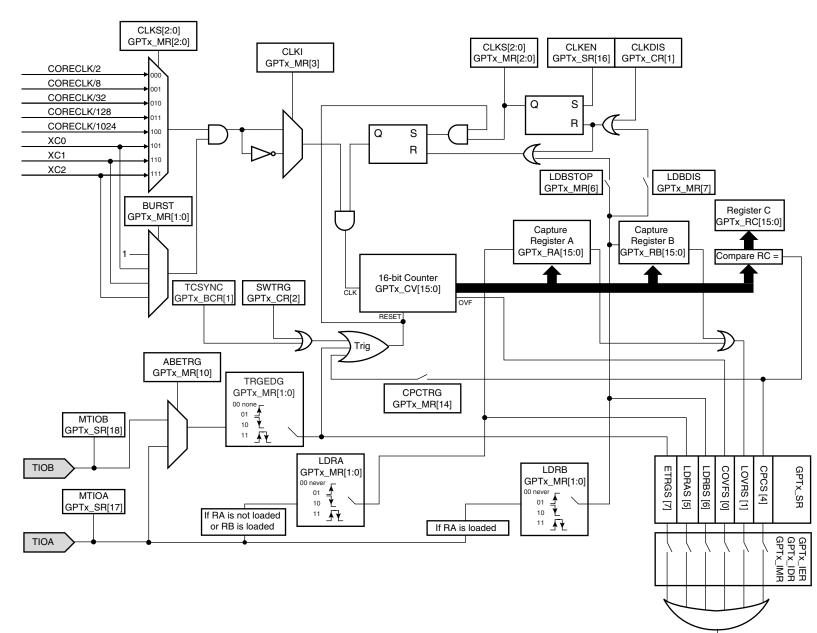






INT







# 27.30 GPT Control Register in Capture Mode

Name: Access: Base Address:	GPT_CI Write-or 0x60						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	—	-	-	—	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	—	-	-	-
7	6	5	4	3	2	1	0
_	—	_	-	SWTRG	CLKDIS	CLKEN	SWRST

#### • SWRST: Software Reset

0: No effect.

1: Generates a software reset.

A software triggered hardware reset of the channel is performed. It resets all the registers, including PIO and PMC registers (except GPTX\_PMSR).

#### • CLKEN: Counter Clock Enable

0: No effect.

1: Enables counter clock if CLKDIS = 0.

#### • CLKDIS: Counter Clock Disable

0: No effect.

1: Disables counter clock.

#### • SWTRG: Software Trigger

0: No effect.

1: Generates a software trigger.

This bit generates a software trigger for resetting and starting the counter at the next valid counter clock edge when the counter clock is enabled.

# 27.31 GPT Mode Register in Capture Mode

Name: Access: Base Address:	GPT_MF Read/Wi 0x64						
31	30	29	28	27	26	25	24
-	-	_	-	-	-	_	-
23	22	21	20	19	18	17	16
-	_	_	-	LDR	B[1:0]	LDR	A[1:0]
15	14	13	12	11	10	9	8
WAVE = 0	CPCTRG	_	-	—	ABETRG	ETRGE	EDG[1:0]
7	6	5	4	3	2	1	0
LDBIS	LDBSTOP	BUR	ST[1:0]	CLKI		CLKS[2:0]	

## CLKS[2:0]: Clock Select

	CLKS[2:0]		Counter Clock Source
0	0	0	CORECLK/2
0	0	1	CORECLK/8
0	1	0	CORECLK/32
0	1	1	CORECLK/128
1	0	0	CORECLK/1024
1	0	1	XC0
1	1	0	XC1
1	1	1	XC2

MCKx consists of five external clocks.

XCx consists of three external clocks.

For more details, see "Clock Sources" on page 322.

## • CLKI: Clock Inverter

0: Normal clock (The counter is incremented on a rising edge)

1: Inverted clock (The counter is incremented on a falling edge)

## • BURST[1:0]: Burst

This signal is combined with the selected clock through a logical AND.

BURS	T[1:0]	Burst Signal Selected
0	0	None
0	1	XC0
1	0	XC1
1	1	XC2





For more details, see "Clock Sources" on page 322.

#### • LDBSTOP Load RB Stops Counter

- 0: The counter is not stopped when RB is loaded.
- 1: The counter is stopped when RB is loaded.

If the counter is stopped, it can restart (to 0x0000) just with a trigger condition.

If a TIOAx edge both induces a trigger condition and loads capture register B which in turn stops the counter, the trigger has no effect.

#### • LDBDIS: Load RB Disables Clock

- 0: The counter clock is not disabled when RB is loaded.
- 1: The counter clock is disabled and the counter stopped when RB is loaded.

If the counter clock is disabled, it can be enabled only by asserting CLKEN, bit [0] of the control register.

## • ETRGEDG[1:0]: External Trigger Edge

The external trigger source is either TIOAx or TIOBx following ABETRG, bit [10] of the mode register.

ETRGEDG[1:0]		Edge
0	0	None
0	1	Rising edge
1	0	Falling edge
1	1	Each edge

When an external trigger is generated, three events occur:

- It resets and starts the counter.
- The ETRGS flag is set in the status register.
- If enabled, ETRGS interrupt is generated.

#### • ABETRG: TIOA or TIOB as External Trigger

- 0: Select TIOBx as external trigger
- 1: Select TIOAx as external trigger

Note: The counter can start only if the clock is enabled.

#### • CPCTRG: Compare RC Trigger

- 0: An equal condition on RC does not cause a trigger.
- 1: An equal condition on RC causes a trigger.

Note: The counter can start only if the clock is enabled.

#### • WAVE: Waveform

- 0: Capture mode.
- 1: Waveform mode.

Note: The capture mode is the default mode after hardware reset.

# 346 **AT91SAM7A2**

#### • LDRA[1:0]: Load RA

These two bits activate one of four possible TIOAx edge conditions to load RA.

Note: The application must ensure that the event that loads RA occurs after the next counter clock edge following the configuration of LDRA (the counter is reset on the next counter clock edge following the configuration of LDRA).

#### • LDRB[1:0]: Load RB

These two bits activate one of four possible TIOAx edge conditions to load RB.

LD	Rx	Edge
0	0	None
0	1	Rising edge
1	0	Falling edge
1	1	Each edge





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Name:	GPT_S	SR					
Access:	Read-o	only					
Base Address	: 0x70						
31	30	29	28	27	26	25	24
_	—	—	_	—	-	_	-
				10	10	. –	10
23	22	21	20	19	18	17	16
_	_	-	-	—	TCLKS	TIOAS	TIOBS
15	14	13	12	11	10	9	8
_	-	-	-	-	MTIOB	MTIOA	CLKSTA
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	-	-	LOVRS	COVFS

# 27.32 GPT Status Register in Capture Mode

Note: This register is a "read-active" register, which means that reading it can affect the state of some bits. When reading GPT\_SR register, following bits are cleared if set: COVFS, LOVRS, CPCS, LDRAS, LDRBS, ETRGS, TIOBS, TIOAS and TCLKS. When debugging, to avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

#### COVFS: Counter Overflow Status

This bit is set when a counter overflow is detected. An overflow occurs when the counter reaches its maximal value 0xFFFF (216 - 1) and passes to 0x0000.

0: No overflow detected.

1: Overflow detected since last read of GPTX\_SR.

## • LOVRS: Load Overrun Status

This bit is set when an overrun is detected. An overrun occurs when the capture registers A or B are reloaded before being read.

0: No overrun detected.

1: An overrun detected since last read of GPTX\_SR.

#### • CPCS: Compare Register C Status

This bit is set when the counter reaches the register C value.

0: Compare C condition has not occurred since last read of GPTX\_SR.

1: Compare C condition has occurred since last read of GPTX\_SR.

#### LDRAS: Load Register A Status

0: Register A not loaded.

1: Register A loaded since last read of GPTX\_SR.

#### • LDRBS: Load Register B Status

0: Register B not loaded.

1: Register B loaded since last read of GPTX\_SR.

# 348 AT91SAM7A2

AT91SAM7A2

#### • ETRGS: External Trigger Status

This bit is set when an external trigger is detected. An external trigger occurs with a valid edge (the edge polarity is set by ETRGEDG[1:0] of the mode register) on the valid trigger pin (set by ABETRG of the mode register).

0: External trigger not detected.

1: External trigger detected since last read of GPTX\_SR.

#### • CLKSTA: Clock Status

0: Clock disabled.

1: Clock enabled.

#### • MTIOA: TIOA Mirror

This bit reflects the TIOAx pin value.

As TIOAx is an input after a hardware reset, its reset value is undefined.

#### • MTIOB: TIOB Mirror

This bit reflects the TIOBx pin value.

As TIOBx is an input after a hardware reset, its reset value is undefined.

#### • TIOBS: TIOB Status

0: At least one input change has been detected on the pin TIOBx since the register was last read.

1: No input change has been detected on the TIOBx pin since the register was last read.

#### • TIOAS: TIOA Status

0: At least one input change has been detected on the TIOAx pin since the register was last read.

1: No input change has been detected on the TIOAx pin since the register was last read.

#### • TCLKS: TCLK Status

0: At least one input change has been detected on the TCLKx pin since the register was last read.

1: No input change has been detected on the TCLKx pin since the register was last read.





# 27.33 GPT Interrupt Enable Register in Capture Mode

Name:GPT\_IERAccess:Write-onlyBase Address:0x74

# 27.34 GPT Interrupt Disable Register in Capture Mode

Name: Access: Base Address:	GPT_ID Write-or 0x78						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	TCLKS	TIOAS	TIOBS
15	14	13	12	11	10	9	8
-	_	_	-	-	_	_	_
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	-	_	LOVRS	COVFS

## • COVFS: Counter Overflow Status

0: COVFS interrupt is disabled.

1: COVFS interrupt is enabled.

#### • LOVRS: Load Overrun Status

0: LOVRS interrupt is disabled.

1: LOVRS interrupt is enabled.

## • CPCS: Compare Register C Status

0: CPCS interrupt is disabled.

1: CPCS interrupt is enabled.

#### LDRAS: Load Register A Status

0: LDRAS interrupt is disabled.

1: LDRAS interrupt is enabled.

## • LDRBS: Load Register B Status

0: LDRBS interrupt is disabled.

1: LDRBS interrupt is enabled.

#### • ETRGS: External Trigger Status

0: ETRGS interrupt is disabled.

1: ETRGS interrupt is enabled.

• TIOBS: TIOB Status

350 **AT91SAM7A2** 

# AT91SAM7A2

0: TIOBS interrupt is disabled.

1: TIOBS interrupt is enabled.

#### • TIOAS: TIOA Status

0: TIOAS interrupt is disabled.

1: TIOAS interrupt is enabled.

## • TCLKS: TCLK Status

0: TCLKS interrupt is disabled.

1: TCLKS interrupt is enabled.





# 27.35 GPT Interrupt Mask Register in Capture Mode

Name: Access: Base Address:	GPT_IN Read-or 0x7C						
31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	—	-	-	TCLKS	TIOAS	TIOBS
15	14	13	12	. 11	10	9	8
-	_	-	-	-	_	_	_
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	-	_	LOVRS	COVFS

#### COVFS: Counter Overflow Status

0: COVFS interrupt is disabled.

1: COVFS interrupt is enabled.

## • LOVRS: Load Overrun Status

0: LOVRS interrupt is disabled.

1: LOVRS interrupt is enabled.

#### CPCS: Compare Register C Status

0: CPCS interrupt is disabled.

1: CPCS interrupt is enabled.

#### • LDRAS: Load Register A Status

0: LDRAS interrupt is disabled.

1: LDRAS interrupt is enabled.

#### • LDRBS: Load Register B Status

0: LDRBS interrupt is disabled.

1: LDRBS interrupt is enabled.

## • ETRGS: External Trigger Status

0: ETRGS interrupt is disabled.

1: ETRGS interrupt is enabled.

#### • TIOBS: TIOB Status

0: TIOBS interrupt is disabled.

1: TIOBS interrupt is enabled.

#### • TIOAS: TIOA Status

0: TIOAS interrupt is disabled.

# 352 AT91SAM7A2

# AT91SAM7A2

# 1: TIOAS interrupt is enabled.

# • TCLKS: TCLK Status

- 0: TCLKS interrupt is disabled.
- 1: TCLKS interrupt is enabled.





# 27.36 GPT Counter Value in Capture Mode

Name: Access: Base Address:	GPT_C\ Read-on 0x80								
31	30	29	28	27	26	25	24		
-	-	_	-	-	-	-	-		
23	22	21	20	19	18	17	16		
-	-	_	-	-	-	-	-		
15	14	13	12	11	10	9	8		
	CV[15:8]								
7	6	5	4	3	2	1	0		
			CV	[7:0]					

# • CV[15:0]: Counter Value

These 16 bits contain the counter value in real time.

The maximal counter value is 0xFFFF = 65535.

When a trigger occurs, the counter will be reset to 0x0000 at the next valid counter clock edge.

# 27.37 GPT Register A in Capture Mode

Name: Access: Base Address:	GPT_RA Read-on 0x84								
31	30	29	28	27	26	25	24		
-	-	-	-	-	-	_	-		
23	22	21	20	19	18	17	16		
-	_	_	_	-	-	_	-		
15	14	13	12	11	10	9	8		
RA[15:8]									
7	6	5	4	3	2	1	0		
			RA	[7:0]					

# • RA[15:0]: Register A Value

This register is loaded with the current counter value when a valid edge occurs on TIOAx pin.

This valid edge is defined by LDRA[1:0] of the mode register.

When this register is loaded, two events occur:

- The LDRAS flag is set in the status register.
- If enabled, LDRAS interrupt is generated.

This register can not be loaded if the counter is stopped or the counter clock disabled.

Name: Access: Base Address:	GPT_R/ Read-or 0x88							
31	30	29	28	27	26	25	24	
-	—	-	-	-	—	_	-	
23	22	21	20	19	18	17	16	
-	_	-	-	-	-	-	-	
15	14	13	12	11	10	9	8	
RB[15:8]								
7	6	5	4	3	2	1	0	
			RB	[7:0]				

# 27.38 GPT Register B in Capture Mode

#### • RB[15:0]: Register B Value

This register is loaded with the current counter value when a valid edge occurs on TIOBx pin.

This valid edge is defined by LDRB[1:0] of the mode register.

When this register is loaded, four events occur:

- The LDRBS flag is set in the status register.
- If enabled, LDRBS interrupt is generated.
- The counter clock can be disabled according to LDBDIS (bit [7] of the mode register).
- The counter can be stopped according to LDBSTOP (bit [6] of the mode register).

This register cannot be loaded if the counter is stopped or the counter clock disabled.





	- <u>g</u>						
Name: Access: Base Address:	GPT_R( Read/W 0x8C						
31	30	29	28	27	26	25	24
—	_	_	_	_	_	_	_
23	22	21	20	19	18	17	16
-	-	—	—	_	_	—	-
15	14	13	12	11	10	9	8
			RC	[15:8]			
7	6	5	4	3	2	1	0
			RC	[7:0]			

# 27.39 GPT Register C in Capture Mode

# • RC[15:0]: Register C Value

When the counter reaches this value, three events occur:

- The CPCS flag is set in the status register.
- If enabled, CPCS interrupt is generated.
- If CPCTRG (bit [14] of the mode register) is high, the counter is reset and restarts at 0x0000.

# 27.40 General-purpose Timer in Waveform Mode

#### 27.40.1 Description

The waveform mode is entered by setting the bit WAVE in the GPTX\_MR to 1. It forces TIOAx as an output pin. TIOBx can be used either as an output (dual waveform mode) or as an input (single waveform mode).

The waveform mode provides the possibility to generate either symmetrical or variable dutycycle waveforms.

The TIOA pin is controlled (set, cleared or toggled) by four events:

- a software trigger
- an external event edge (rising, falling edge or both)
- an equality between the counter and compare register A value
- an equality between the counter and compare register C value

As an output pin, the TIOB pin is controlled (set, cleared or toggled) by four events:

- a software trigger
- an external event edge (rising, falling edge or both)
- an equality between the counter and compare register B value
- an equality between the counter and compare register C value

When TIOB is used as an external trigger source, the compare register B is not used.

When an equal condition on compare register C is detected, one of three events can occur:

- The counter can reset and start at the next valid counter clock edge.
- The counter can be stopped.
- The counter can be stopped and the counter clock disabled.

If this condition restarts the counter, the user can generate a continuous wave with a period proportional to compare register C +1 value. If it does not restart the counter, the user can generate a continuous waveform with a period proportional to 0xFFFF (maximal counter value:  $2^{16}$  - 1).

The user may choose an internal clock source (CORECLK/2, CORECLK/8, CORECLK/32, CORECLK/128 or CORECLK/1024) or external clock (TCLK0, TCLK1 or TCLK2). A burst mode is available. It generates a burst clock. For more details, refer to "Clock Sources" on page 322.

Five interrupts can be produced:

- External trigger detected
- Counter overflow (when the counter passes from 0xFFFF to 0x0000)
- Compare RA (the counter reaches the value stored in register A)
- Compare RB (the counter reaches the value stored in register B)
- Compare RC (the counter reaches the value stored in register C)

Finally, the synchronize register can be used to cause a software trigger for reset and start the counter at the next valid counter clock edge on all channels at the same time.

Figure 27-10 on page 358 to Figure 27-14 on page 360 show different applications possible with the waveform mode.



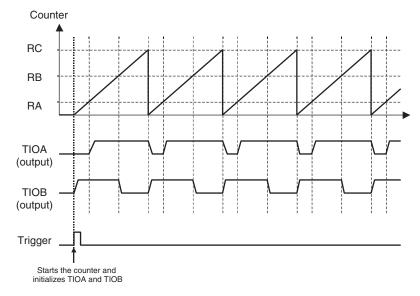


27.40.1.1 Dual Pulse Width Modulation (PWM) Generation

TIOAx is toggled by RA and RC, TIOBx by RB and RC.

RC contains frequency of both signals. RA determines the TIOAx duty cycle and RB the TIOBx duty cycle.





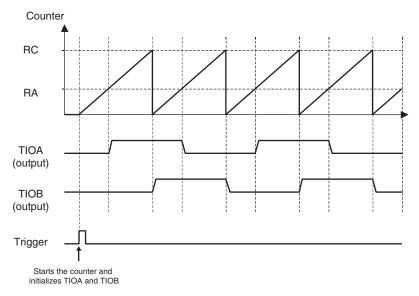
27.40.1.2 Generation of Two Identical Out-of-phase Square-wave Signals

TIOAx is toggled each time the counter reaches RA value, TIOBx each time the counter reaches RC value.

A trigger (external or software) starts the counter and initializes TIOAx and TIOBx.

RC contains the frequency of both signals. RA contains the delay between the signals.

## Figure 27-11. Two Square Signals



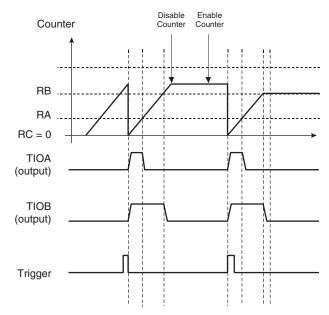
#### 27.40.1.3 Pulse Generation

This application generates only one pulse on TIOAx and on TIOBx between two triggers. The pulses start with a trigger. The duration is given by the value of RA for TIOAx pulse and the value of RB for TIOBx pulse. After each new trigger, another pulse is generated. When a trigger occurs, the counter is reset at the next valid counter clock edge when the counter clock is enabled.

If we want to start the pulse exactly when the counter is reset, RC must equal 0. Thus, it is the comparison with RC = 0 that starts the pulse, and not the trigger.

In this case, the user must disable the counter clock when a compare RB is detected to stop the counter.

To accept a new trigger, the user must then re-enable the counter clock.



#### Figure 27-12. Pulse Generation

#### 27.40.1.4 Trigger on TIOB Input Pin

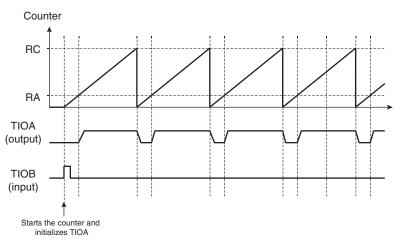
In each of the previous examples, TIOBx is used as an output. It is possible to use it as a trigger input and generate only TIOAx output signal.

The following application is the same as "Dual Pulse Width Modulation (PWM) Generation" on page 358, where TIOBx is not used as an output but as an input.







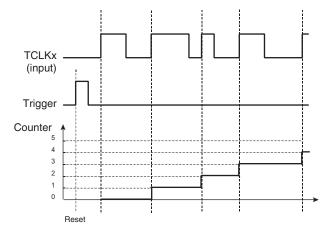


27.40.1.5 Event Counter on an External Clock (TCLK)

The counter is incremented with each TCLKx rising edge.

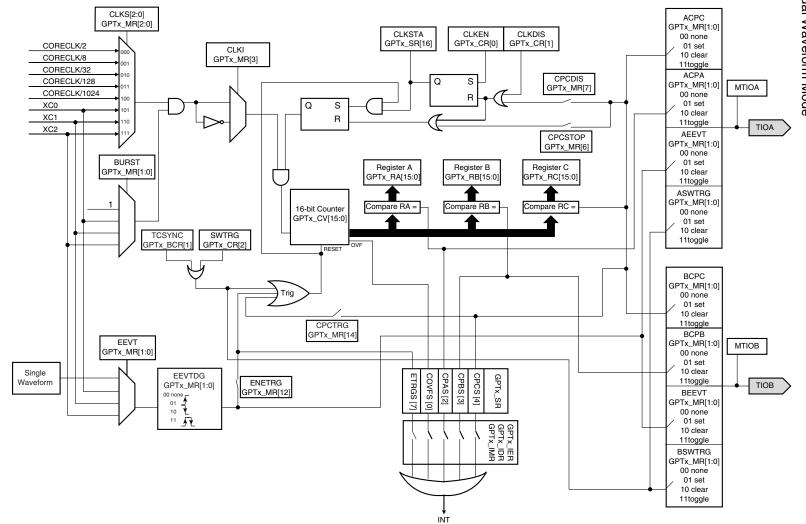
This application can be generated in capture mode.

Figure 27-14. Event Counter



27.40.2 Dual Waveform Mode

Figure 27-15. Dual Waveform Mode

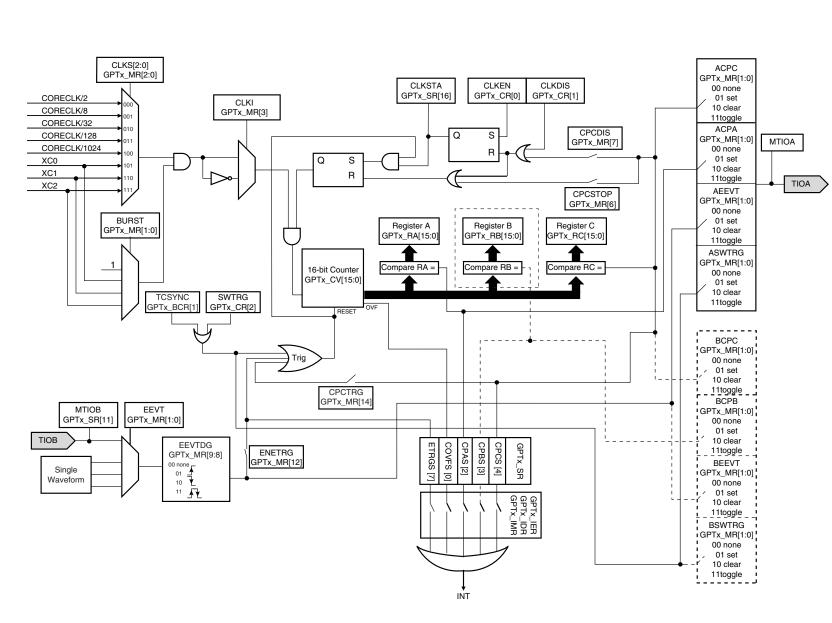






# 27.40.3 Single Waveform Mode

Figure 27-16. Single Waveform Mode (The dotted blocks are not used in this case.)





# 27.41 GPT Control Register in Waveform Mode

Name: Access: Base Address:	GPT_C Write-or 0x60						
31	30	29	28	27	26	25	24
-	_	-	-	-	_	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	—	-
15	14	13	12	11	10	9	8
-	_	-	-	-	-	_	-
7	6	5	4	3	2	1	0
-	_	-	-	SWTRG	CLKDIS	CLKEN	SWRST

## • SWRST: Software Reset

0: No effect.

1: Generates a software reset.

A software triggered hardware reset of the channel is performed. It reset all the registers, including PIO and PMC registers (except GPTX\_PMSR).

# • CLKEN: Counter Clock Enable

0: No effect.

1: Enables counter clock if CLKDIS = 0.

# • CLKDIS: Counter Clock Disable

0: No effect.

1: Disables counter clock.

#### • SWTRG: Software Trigger

0: No effect.

1: Generates a software trigger.

This bit generates a software trigger for resetting and starting the counter at the next valid counter clock edge when the counter clock is enabled.





# 27.42 GPT Mode Register in Waveform Mode

Name: Access: Base Address:	GPT_M Read/W 0x64						
31	30	29	28	27	26	25	24
BSWTRG[1:0]		BEE	VT[1:0]	BCP	C[1:0]	BCP	B[1:0]
23	22	21	20	19	18	17	16
ASWTF	RG[1:0]	AEEVT[1:0]		ACPC[1:0]		ACPA[1:0]	
15	14	13	12	11	10	9	8
WAVE = 1	CPCTRG	_	ENETRG	EEV.	T[1:0]	EEVTE	DG[1:0]
7	6	5	4	3	2	1	0
CPCDIS	CPCSTOP	BUF	ST[1:0]	CLKI		CLKS[2:0]	

# • CLKS[2:0]: Clock Select

	CLKS[2:0]		Counter Clock Source
0	0	0	CORECLK/2
0	0	1	CORECLK/8
0	1	0	CORECLK/32
0	1	1	CORECLK/128
1	0	0	CORECLK/1024
1	0	1	XC0
1	1	0	XC1
1	1	1	XC2

For more details, see "Clock Sources" on page 322.

## • CLKI: Clock Inverter

0: Normal clock (The counter is incremented on a rising edge)

1: Inverted clock (The counter is incremented on a falling edge)

## • BURST[1:0]: Burst

This signal is combined with the selected clock through a logical AND.

BURST[1:0]		Burst signal selected
0	0	None
0	1	XC0
1	0	XC1
1	1	XC2

For more details, see "Clock Sources" on page 322.

## • CPCSTOP: Compare RC Stops the Counter

0: The counter is not stopped when an equal condition on RC is detected.

1: The counter is stopped when an equal condition on RC is detected.

If the counter is stopped, it can restart (to 0x0000) just with a trigger condition.

If an equal condition on RC induces both trigger condition and stop counter, the trigger will have no effect.

#### • CPCDIS: Compare RC Disables Clock

0: The counter clock is not disabled when an equal condition on RC is detected.

1: The counter clock is disabled and the counter stopped when an equal condition on RC is detected.

If the counter clock is disabled, it can be enabled only by asserting CLKEN, bit [1] of the control register.

## • EEVTEDG[1:0]: External Event Edge

These two bits activate one of four possible external event modes. The external event source is selected by EEVT[1:0] of the mode register.

EEVTEDG[1:0]		Edge
0	0	None
0	1	Rising edge
1	0	Falling edge
1	1	Each edge

When an external event is generated, five events occur:

- The ETRGS flag is set in the status register.
- If enabled, ETRGS interrupt is generated.
- It can reset and start the counter at the next valid counter clock edge if ENETRG (bit [12] of the mode register) is high.
- TIOAx pin can be set, clear, toggle or unchanged following AEEVT[1:0] of the mode register.
- TIOBx pin can be set, clear, toggle or unchanged following BEEVT[1:0] of the mode register.

#### • EEVT[1:0]: External Event

These bits select an external event source among four pins:

EEVT[1:0]		External Trigger
0	0	TIOBx
0	1	XC0
1	0	XC1
1	1	XC2

If TIOBx is selected, the mode is in single waveform mode (see Figure 27-16 on page 362). TIOAx is used as an output and TIOBx as an input. The following bits are disabled:

- BSWTRG[1:0] of the mode register
- BEEVT[1:0] of the mode register
- BCPC[1:0] of the mode register





- BCPB[1:0] of the mode register
- Compare register B

If an external clock is selected, the mode is in dual waveform mode. TIOAx and TIOBx are used as outputs.

## • ENETRG: Enable External Trigger

This bit determines whether an external event can be used as a trigger to reset and start the counter at the next valid counter clock edge. The external event source is selected by EEVT[1:0] of the mode register.

0: External event does not reset and start the counter. Selected external trigger can only be used to control TIOAx and TIOBx.

1: External trigger resets and starts the counter.

Note: The counter can start only if the clock is enabled.

#### • CPCTRG: Compare RC Trigger

This bit determines whether an equal condition on the Compare C register can cause a trigger (reset and start the counter at the next valid counter clock edge).

0: An equal condition on RC does not cause a trigger.

1: An equal condition on RC causes a trigger.

Note: The counter can start only if the clock is enabled.

#### • WAVE: Waveform

0: Capture mode.

1: Waveform mode.

#### • ACPA: TIOA Compare A

These two bits determine the effect on the TIOAx output pin caused by an equal comparison between the counter and the Compare Register A value. See the bit description table in "ASWTRG: TIOA Software Trigger" where xxx = CPA.

Note: If several events that control TIOAx output (set, clear or toggle) arrive at the same time, only one has an action according to the following priority order:

- 1. ASWTRG (highest priority)
- 2. AEEVT
- 3. ACPC
- 4. ACPA

#### • ACPC: TIOA Compare C

These two bits determine the effect on the TIOAx output pin caused by an equal comparison between the counter and the Compare register C value. See the bit description table in "ASWTRG: TIOA Software Trigger" where xxx = CPC.

## • AEEVT: TIOA External Event

These two bits determine the effect on the TIOAx output pin caused by an external event. The external event source is selected by EEVT[1:0] of the mode register. See the bit description table in "ASWTRG: TIOA Software Trigger" where xxx = EEVT.

## • ASWTRG: TIOA Software Trigger

These two bits determine the effect on the TIOAx output pin caused by a software trigger.

See the following table where xxx = SWTRG.

Аххх		Waveform Pin
0	0	None
0	1	Set
1	0	Clear
1	1	Toggle

## • BCPB: TIOB Compare B

These two bits determine the effect on the TIOBx output pin caused by an equal comparison between the counter and the Compare Register B value. These bits are active only if TIOBx is not an input (see "EEVT[1:0]: External Event" of the "GPT Mode Register in Waveform Mode" on page 364). See the bit description table in "BSWTRG: TIOB Software Trigger" where xxx = CPB.

Note: If several events that control TIOBx output (set, clear or toggle) arrive at the same time, only one has an action according to the following priority:

- 1. BSWTRG (highest priority)
- 2. BEEVT
- 3. BCPC
- 4. BCPB

## BCPC: TIOB Compare C

These two bits determine the effect on the TIOBx output pin caused by an equal comparison between the counter and the Compare register C value. These bits are active only if TIOBx is not an input (see "EEVT[1:0]: External Event" of the "GPT Mode Register in Waveform Mode" on page 364). See the bit description table in "BSWTRG: TIOB Software Trigger" where xxx = CPC.

#### • BEEVT: TIOB External Event

These two bits determine the effect on the TIOBx output pin caused by an external event. The external event source is selected by EEVT[1:0] of the mode register. These bits are active only if TIOBx is not an input (see "EEVT[1:0]: External Event" of the "GPT Mode Register in Waveform Mode" on page 364). See the bit description table in "BSWTRG: TIOB Software Trigger" where xxx = EEVT.

# BSWTRG: TIOB Software Trigger

These two bits determine the effect on the TIOBx output pin caused by a software trigger. These bits are active only if TIOBx is not an input (see "EEVT[1:0]: External Event" of the "GPT Mode Register in Waveform Mode" on page 364). See following table with xxx = SWTRG:

B	xxx	Waveform Pin
0	0	None
0	1	Set
1	0	Clear
1	1	Toggle





# 27.43 GPT Status Register in Waveform Mode

Name: Access: Base Address:	GPT_SF Read-or 0x70						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	TCLKS	TIOAS	TIOBS
15	14	13	12	11	10	9	8
-	_	-	-	—	MTIOB	MTIOA	CLKSTA
7	6	5	4	3	2	1	0
ETRGS	_	_	CPCS	CPBS	CPAS	_	COVFS

Note: This register is a "read-active" register; thus, reading it can affect the state of some bits. When reading GPT\_SR register, following bits are cleared if set: COVFS, CPAS, CPBS, CPCS, ETRGS, TIOBS, TIOAS and TCLKS. When debugging, to avoid this behavior, users should use ghost registers (see "Ghost Registers" on page 9).

# COVFS: Counter Overflow Status

This bit is set when a counter overflow is detected. An overflow occurs when the counter reaches its maximal value 0xFFFF (2<sup>16</sup>-1) and passes to 0x0000.

- 0: No overflow detected
- 1: Overflow detected since last read of GPTX\_SR

# • CPAS: Compare Register A Status

This bit is set when the counter reaches the register A value.

- 0: Compare A condition has not occurred since last read of GPTX\_SR
- 1: Compare A condition has occurred since last read of GPTX\_SR

# • CPBS: Compare Register B Status

This bit is set when the counter reaches the register B value.

- 0: Compare B condition has not occurred since last read of GPTX\_SR
- 1: Compare B condition has occurred since last read of GPTX\_SR

# • CPCS: Compare Register C Status

This bit is set when the counter reaches the register C value.

- 0: Compare C condition has not occurred since last read of GPTX\_SR
- 1: Compare C condition has occurred since last read of GPTX\_SR

# • ETRGS: External Trigger Status

This bit is set when an external trigger is detected. An external trigger occurs with a valid edge (the edge polarity is set by EEVTEDG[1:0] of the mode register) on the valid trigger pin (set by EEVT[1:0] of the mode register if ENETRG, bit 12 of the Mode Register, is high).

0: External trigger not detected

# 368 AT91SAM7A2

1: External trigger detected since last read of GPTX\_SR

## • CLKSTA: Clock Status

- 0: Clock disabled
- 1: Clock enabled

#### • MTIOA: TIOA Mirror

This bit reflects the TIOAx pin value.

Its reset value is undefined because the operating mode after hardware reset is capture mode.

#### • MTIOB: TIOB Mirror

This bit reflects the TIOBx pin value.

Its reset value is undefined because the operating mode after hardware reset is capture mode.

#### • TIOBS: TIOB Status

0: At least one input change has been detected on the pin TIOBx since the register was last read.

1: No input change has been detected on the TIOBx pin since the register was last read.

#### • TIOAS: TIOA Status

0: At least one input change has been detected on the TIOAx pin since the register was last read.

1: No input change has been detected on the TIOAx pin since the register was last read.

# • TCLKS: TCLK Status

0: At least one input change has been detected on the TCLKx pin since the register was last read.

1: No input change has been detected on the TCLKx pin since the register was last read.

Note: X: Channel number





# 27.44 GPT Interrupt Enable Register in Waveform Mode

Name:GPT\_IERAccess:Write-onlyBase Address:0x74

# 27.45 GPT Interrupt Disable Register in Waveform Mode

Name: Access: Base Address:	GPT_ID Write-or 0x78						
31	30	29	28	27	26	25	24
_	-	-	-	-	_	-	-
23	22	21	20	19	18	17	16
_	-	_	-	-	TCLKS	TIOAS	TIOBS
15	14	13	12	11	10	9	8
-	-	—	-	-	-	-	-
7	6	5	4	3	2	1	0
ETRGS	_	—	CPCS	CPBS	CPAS	-	COVFS

# • COVFS: Counter Overflow Status

0: COVFS interrupt is disabled.

1: COVFS interrupt is enabled.

# CPAS: Compare Register A Status

- 0: CPAS interrupt is disabled.
- 1: CPAS interrupt is enabled.

# • CPBS: Compare Register B Status

- 0: CPBS interrupt is disabled.
- 1: CPBS interrupt is enabled.

#### • CPCS: Compare Register C Status

- 0: CPCS interrupt is disabled.
- 1: CPCS interrupt is enabled.

# • ETRGS: External Trigger Status

- 0: ETRGS interrupt is disabled.
- 1: ETRGS interrupt is enabled.

#### • TIOBS: TIOB Status

0: TIOBS interrupt is disabled.

1: TIOBS interrupt is enabled.

• TIOAS: TIOA Status

# 370 **AT91SAM7A2**

0: TIOAS interrupt is disabled.

1: TIOAS interrupt is enabled.

## • TCLKS: TCLK Status

- 0: TCLKS interrupt is disabled.
- 1: TCLKS interrupt is enabled.





# 27.46 GPT Interrupt Mask Register in Waveform Mode

Name: Access: Base Address:	GPT_IN Read-or 0x7C						
31	30	29	28	27	26	25	24
-	_	—	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	—	-	-	TCLKS	TIOAS	TIOBS
15	14	13	12	11	10	9	8
-	_	-	-	-	-	-	-
7	6	5	4	3	2	1	0
ETRGS	_	—	CPCS	CPBS	CPAS	—	COVFS

## COVFS: Counter Overflow Status

0: COVFS interrupt is disabled.

1: COVFS interrupt is enabled.

# • CPAS: Compare Register A Status

0: CPAS interrupt is disabled.

1: CPAS interrupt is enabled.

## CPBS: Compare Register B Status

0: CPBS interrupt is disabled.

1: CPBS interrupt is enabled.

## • CPCS: Compare Register C Status

0: CPCS interrupt is disabled.

1: CPCS interrupt is enabled.

#### • ETRGS: External Trigger Status

0: ETRGS interrupt is disabled.

1: ETRGS interrupt is enabled.

#### • TIOBS: TIOB Status

0: TIOBS interrupt is disabled.

1: TIOBS interrupt is enabled.

#### • TIOAS: TIOA Status

0: TIOAS interrupt is disabled.

1: TIOAS interrupt is enabled.

# • TCLKS: TCLK Status

0: TCLKS interrupt is disabled.

# 372 **AT91SAM7A2**

1: TCLKS interrupt is enabled.

# 27.47 GPT Counter Value in Waveform Mode

Name: Access: Base Address:	GPT_C` Read-or 0x80						
31	30	29	28	27	26	25	24
_	_	—	-	-	-	_	-
23	22	21	20	19	18	17	16
-	_	—	-	-	-	_	-
15	14	13	12	11	10	9	8
			CV	[15:8]			
7	6	5	4	3	2	1	0
			CV	[7:0]			

# • CV[15:0]: Counter Value

These 16 bits contain the counter value in real time.

The maximal counter value is  $0xFFFF 2^{16} - 1 = 65535$ .

When a trigger occurs, the counter will be reset to 0x0000 at the next valid counter clock edge.

# 27.48 GPT Register A in Waveform Mode

27.40 01 11	legister A li	Waveloini	Mode				
Name:	GPT_R						
Access:	Read/W	/rite					
Base Address	: 0x84						
31	30	29	28	27	26	25	24
_	_	_	-	_	_	_	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	_	-
15	14	13	12	11	10	9	8
			RA	[15:8]			
7	6	5	4	3	2	1	0
			RA	[7:0]			

# • RA[15:0]: Register A Value

When the counter reaches this value, three events occur:

- The CPAS flag is set in the status register.
- If enabled, CPAS interrupt is generated.
- TIOAx pin can be set, clear, toggle or unchanged following bits ACPA[1:0] of the mode register.





# 27.49 GPT Register B in Waveform Mode

Name:	GPT_R	3					
Access:	Read/W	rite					
Base Address:	0x88						
31	30	29	28	27	26	25	24
-	_	_	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	_	-	-	-	_	-
15	14	13	12	11	10	9	8
			RB[	15:8]			
7	6	5	4	3	2	1	0
			RB	[7:0]			

# • RB[15:0]: Register B Value

When the counter reaches this value, three events occur:

- The CPBS flag is set in the status register.
- If enabled, CPBS interrupt is generated.
- TIOBx pin can be set, clear, toggle or unchanged following bits BCPB[1:0] of the mode register.

These bits are active only if TIOB is not an input (see "EEVT[1:0]: External Event" of the "GPT Mode Register in Waveform Mode" on page 364).

	g.e.e. e						
Name: Access: Base Address:	GPT_R0 Read/W 0x8C						
31	30	29	28	27	26	25	24
_	—	_	-	-	_	—	-
23	22	21	20	19	18	17	16
_	-	—	-	_	_	_	-
15	14	13	12	11	10	9	8
			RC[	15:8]			
7	6	5	4	3	2	1	0
			RC	[7:0]			

# 27.50 GPT Register C in Waveform Mode

## • RC[15:0]: Register C Value

When the counter reaches this value, seven events can occur:

- The CPCS flag is set in the status register.
- If enabled, CPCS interrupt is generated.
- If bit CPCTRG (bit [14] of the GPTX\_MR) is high, the counter is reset and restarts at 0x0000 at the next valid counter clock edge.
- The counter clock can be disabled according to CPCDIS (bit [7] of the mode register).
- The counter can be stopped according to CPCSTOP (bit [6] of the mode register).
- TIOAx pin can be set, clear, toggle or unchanged following bits ACPC[1:0] of the mode register.
- TIOBx pin can be set, clear, toggle or unchanged following bits BCPC[1:0] of the mode register.



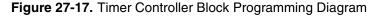


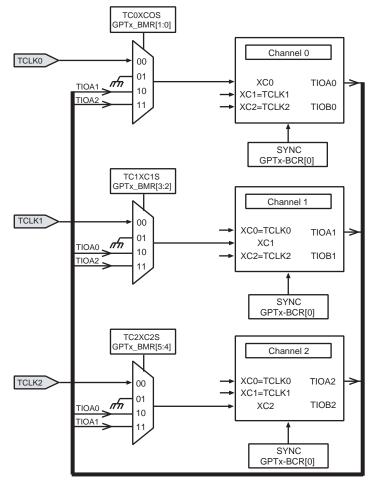
# 27.51 Timer Controller Block Programming

This module controls the entire timer controller with its three channels.

It has two functions:

- The block control register provides the means to synchronize the three timer channels. It can generate a software trigger on all three channels at exactly the same time.
- The block mode register provides the means to daisy chain two or three channels. Thus the user can improve counter capacity.





#### 27.51.1 External Clock Generation for Channel 1 Using Channel 0

With GPT0\_RA and GPT0\_RC, Channel 0 generates a pulse width modulation output (PWM) in waveform mode that is used as an external clock by Channel 1.

Note that if RC is not used, this generates a 32-bit counter.

Each time the counter 0 passes 0xFFFF (overflow condition), this increments the counter by 1.

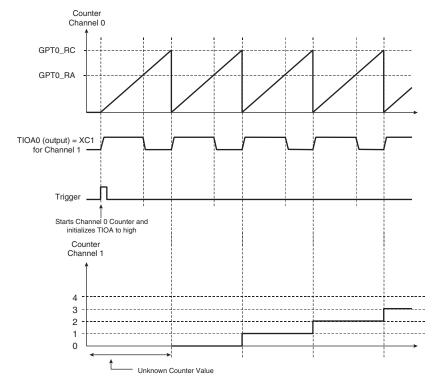


Figure 27-18. Timer Controller Block Programming Application





# 27.52 GPT Block Control Register

Name: Access: Base Address:	GPT_B0 Write-or 0x00						
31	30	29	28	27	26	25	24
-	_	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	—	-	-	-	—	-	-
7	6	5	4	3	2	1	0
-	_	_	-	-	-	TCSYNC	SWRST

## • SWRST: Software Reset

This bit generates a software reset on the three timer channels simultaneously.

0: No effect.

1: Generates a software reset.

## • TCSYNC: Synchronization Bit

This bit generates a software trigger on the three channels of the general-purpose timer simultaneously.

A software trigger resets and starts the counter at the next valid counter clock edge.

0: No effect.

1: Resets and starts all three timer channel counters simultaneously.

# 27.53 GPT Block Mode Register

Name: Access: Base Address:	GPT_BI Read/W						
31	30	29	28	27	26	25	24
-	—	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	_	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	_	_	-	-	—	-	-
7	6	5	4	3	2	1	0
-	_	_	-	_	-		

# TC0XC0S[1:0]: TCLK0 XC0 Selection

These bits select the external clock XC0 source for the channel 0.

TC0XC0S[1:0]		Selected Signal
0	0	TCLK0
0	1	none
1	0	TIOA1
1	1	TIOA2

# TC1XC1S[1:0]: TCLK1 XC1 Selection

These bits select the external clock XC1 source for the channel 1.

TC1XC1S[1:0]		Selected Signal
0	0	TCLK1
0	1	none
1	0	TIOA0
1	1	TIOA2

# TC2XC2S[1:0]: TCLK2 XC2 Selection

These bits select the external clock XC2 source for the channel 2.

TC2XC2S[1:0]		Selected Signal
0	0	TCLK2
0	1	none
1	0	TIOA0
1	1	TIOA1





# 28. Electrical Characteristics

Applicable over recommended operating temperature and voltage range.

# 28.1 3V Core and I/O Characteristics

## Table 28-1.3V Power Supply

Symbol	Parameter	Min	Тур	Max	Units
V <sub>VDD3V</sub>	Supply voltage	3.0	3.3	3.6	V
V <sub>IH</sub>	High-level input voltage	0.7 x V <sub>VDDCORE</sub>		V <sub>VDDCORE</sub> + 0.3	V
V <sub>IL</sub>	Low-level input voltage	-0.3		0.3 x V <sub>DDCORE</sub>	V
V <sub>IHST</sub>	High-level input voltage (Schmitt trigger)	1.625		1.825	V
V <sub>ILST</sub>	Low-level input voltage (Schmitt trigger)	1.075		1.225	V
V <sub>HYS</sub>	Hysteresis input voltage (Schmitt trigger)	0.400		0.750	V
V <sub>OH</sub>	High-level output voltage (drive = 0.3 mA)	V <sub>VDDCORE</sub> - 0.1			V
V <sub>OL</sub>	Low-level output voltage (drive = 0.3 mA)			V <sub>SS</sub> + 0.1	V
I <sub>LEAK</sub>	Input leakage current		90		nA
O <sub>LEAK</sub>	Output leakage current		100		nA
FANIN	Pad capacitance		6		pF
I <sub>PD</sub>	Internal pull-down current	99		429	mA
I <sub>PU</sub>	Internal pull-up current	130		352	mA
ESD	Electrostatic Discharge (Human body model)				V

Note: 1. In CMOS, the behavior of a cell is independent of its load, as loads are purely capacitive.

# 28.2 5V I/O Characteristics

# Table 28-2.5V Power Supply

Symbol	Parameter	Min	Тур	Max	Units
$V_{VDD5V}$	Supply voltage	V <sub>VDD3V</sub>		5.5	V
V <sub>IH</sub>	High-level input voltage	2.0		V <sub>VDD5V</sub> + 0.3	V
V <sub>IL</sub>	Low-level input voltage	-0.3		0.8	V
V <sub>IHST</sub>	High-level input voltage (Schmitt trigger)	1.675		1.725	V
V <sub>ILST</sub>	Low-level input voltage (Schmitt trigger)	1.025		1.125	V
V <sub>HYS</sub>	Hysteresis input voltage (Schmitt trigger)	0.550		0.700	V
V <sub>OH</sub>	High-level output voltage (drive = pin output current)	V <sub>VDD5V</sub> - 0.4			V
V <sub>OL</sub>	Low-level output voltage (drive = pin output current)			0.4	V
I <sub>LEAK</sub>	Input leakage current		90		nA
O <sub>LEAK</sub>	Output leakage current		100		nA
FANIN	Pad capacitance		6		pF

Notes: 1. In CMOS, the behavior of a cell is independent of its load, as loads are purely capacitive.

# 28.3 Analog Characteristics

# Table 28-3.Analog Power Supply

Symbol	Parameter	Min	Тур	Max	Units
V <sub>VDDANA</sub>	Supply voltage	3.0		3.6	V
V <sub>ana</sub>	Analog input voltage	V <sub>SS</sub>		V <sub>VDDANA</sub>	V
V <sub>REFP</sub>	Positive analog voltage reference	2.4		V <sub>VDDANA</sub>	V
I <sub>LEAK</sub>	Input leakage current	-100		+100	nA
FANIN	Pad capacitance	4		8	pF
ESD	Electrostatic Discharge (Human body model)	1000			V

# 28.4 ADC Characteristics

Worst cases of VDDana and process, unless otherwise noted.

#### Table 28-4.ADC Characteristics

Code	Parameter	Condition	Min	Тур	Max	Unit
	Resolution			10		bit
VDDana	Supply voltage		3.0	3.3	3.6	V
∆VDDana	Supply ripple	rms value, 10 kHz to 10 MHz			30	mV
IDDana on	Current consumption (not taking into account Vref current)	onad = 1	40	150	400	μA
IDDana stdby	Standby current consumption	onad = 0			10	μA
VREFP	Positive reference voltage range		2.4	3.0	3.6 30 400 10 Vdd 24 700 90 1.5 1 10 9 9 1.5 1.5	V
	VREF input resistance	@ 25°C	12	18	24	kΩ
	Conversion time	1 for sample, 10 for conversion	11			clk
	Clock frequency (1/Tclk)	For specified duty cycle range			700	kHz
	Input capacitance <sup>(1)</sup>	Switched during sampling process	0	80	90	pF
	Input capacitance (including sample and hold)	Pad selected		107		pF
	Integral Non-Linearity		-1.5	-0.5/+0.7	1.5	lsb
	Differential Non-Linearity	3.0 ≤Vdd ≤3.6V, 3.0 ≤Vrefp ≤Vdd,	-0.999	-0.3/+0.6	1	lsb
	Offset	Clk ≤250 kHz, -40° C ≤Temp ≤105° C Best fit straight line	-10	3	10	mV
	Gain error		-9	-1	9	mV
	Integral Non-Linearity		-1.5	-0.5/+0.7	1.5	lsb
	Differential Non-Linearity	3.0 ≤Vdd ≤3.6V, 3.0 ≤Vrefp ≤Vdd,	-0.999	-0.3/+0.6	1.5	lsb
	Offset	Clk ≤500 kHz, -40° C ≤Temp ≤105° C Best fit straight line	-10	3	10	mV
	Gain error		-9	-1	9	mV





# Table 28-4. ADC Characteristics (Continued)

Code	Parameter	Condition	Min	Тур	Max	Unit
	Integral Non-Linearity		-3		3	lsb
	Differential Non-Linearity	3.0 ≤Vdd ≤3.6V, 2.4 ≤Vrefp ≤Vdd,	-0.999		1.5	lsb
	Offset	Clk ≤700 kHz, -40° C ≤Temp ≤105° C Best fit straight line	-10		10	mV
	Gain error		-9		9	mV
	Startup time		0.7	10	13	mW

Note: 1. Capacitor on pin does not take into account any pad and/or connection to pad capacitance.

# 29. Packaging Information

# 29.1 Thermal Data

The heat transfer between the top surface of the die and the surrounding ambient air can be characterized by the following equation:

$$T_J - T_A = P \times \Theta_{JA}$$

where:

P = Device operating power (in W)

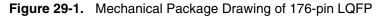
 $T_J$  = Temperature of a junction on the device (in °C)

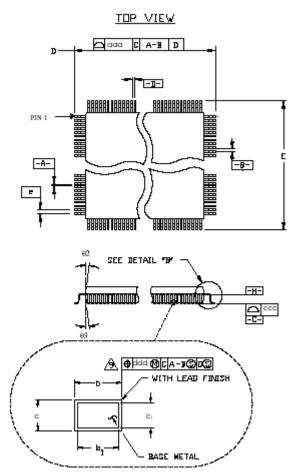
 $T_A$  = Temperature of the surrounding ambient air (in °C)

 $\theta_{JA}$  = Package thermal resistance i.e. between junction and ambient air (in ° C/W)

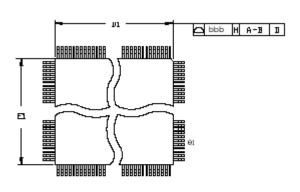
The package thermal resistance  $\theta_{JA}$  for the 176-pin LQFP package is 21° C/W.

# 29.2 Package Drawing





BOTTOM VIEW



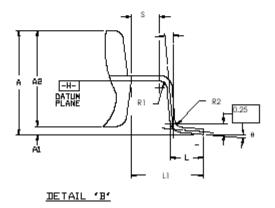






Table 29-1. Package Dimensions in mm

Symbol	Min	Nom	Мах
С	0.09		0.20
с1	0.09		0.16
L	0.45	0.6	0.75
L1		1.00 REF	
R2	0.08		0.2
R1	0.08		
S	0.2		
q	0°	3.5°	<b>7</b> °
θ1	0°		
θ2	11°	12°	13°
θ3	11°	12°	13°
А			1.6
A1	0.05		0.15
A2	1.35	1.4	1.45
Tolerar	nces of Form and Position		
aaa		0.2	
bbb		0.2	

# Table 29-2. Lead Count Dimensions (mm)

Pin	D/E	D1/E1		b			b1		е		
Count	BSC	BSC	Min	Nom	Max	Min	Nom	Max	BSC	ccc	ddd
176	26.0	24.0	0.17	0.20	0.27	0.17	0.20	0.23	0.50	0.10	0.08

Table 29-3.
 Device and 176-lead LQFP Package Maximum Weight

1900 mg
---------

# **30. Soldering Profile**

Table 30-1 below gives the recommended soldering profile from J-STD-20.

	Convection or IR/Convection	VPR
Average Ramp-up Rate (183°C to Peak)	3° C/sec. max	10° C/sec.
Preheat Temperature 125° C ±25° C	120 sec. max	
Temperature Maintained Above 183° C	60 sec. to 150 sec.	
Time within 5°C of Actual Peak Temperature	10 sec. to 20 sec.	60 sec.
Peak Temperature Range	220 +5/-0° C or 235 +5/-0° C	215 to 219° C or 235 +5/-0° C
Ramp-down Rate	6° C/sec.	10° C/sec.
Time 25° C to Peak Temperature	6 min. max	

# Table 30-1.Soldering Profile

Small packages may be subject to higher temperatures if they are reflowed in boards with larger components. In this case, small packages may have to withstand temperatures of up to 235°C, not 220°C (IR reflow).

Recommended package reflow conditions depend on package thickness and volume. See Table 30-2 below.

Table 30-2. Recommended Package Reflow Conditions	1)(2)(3)
---	----------

Parameter	Temperature
Convection	235 +5/-0° C
VPR	215 to 219° C
IR/Convection	235 +5/-0° C

Notes: 1. The packages are qualified by Atmel by using IR reflow conditions, not convection or VPR.

- 2. By default, the package level 1 is qualified at 220°C (unless 235°C is stipulated).
- 3. The body temperature is the most important parameter but other profile parameters such as total exposure time to hot temperature or heating rate may also influence component reliability.

A maximum of three reflow passes is allowed per component.





# **31. Environmental Specifications**

# 31.1 Operational Temperature

The operational temperature range of the AT91SAM7A2 embedded system is -40°C to +85°C.

# 31.2 Storage Temperature

The storage temperature range of the AT91SAM7A2 embedded system chip is -50°C to +150°C.

# 32. Revision History

# Table 32-1.

Document Ref.	Comments	Change Request Ref.
6021A	First issue.	
	In "Clocks", in Section 6.1 "Overview" on page 18 and in "Clock Manager (CM)", Section 12.1 "Overview" on page 61, updated oscillator frequency range.	2931
	In Section 6.3 "Phase Locked Loop" on page 18, removed text on characteristics, added recommended value for the RC filter, updated min value for VCO in Table 6-2, "PLL Characteristics," on page 19.	2930, 2937
	Updated tables in Section 11.5.5 "Timings" on page 46 and added Figure 11-32 "Synchronous Read and Write Access Waveform," on page 53.	2940
	Added information on stopping PDC transfers in Section 16.5 "Transfer Counter" on page 97.	2932
	Added information on checksum and LIN protocol revision support to Section 19.7 "LIN Protocol" on page 148.	2939
6021B	Added text to SYSCAL bit descriptions in Section 21.16 "ST Channel 0 Prescalar Register" on page 208 and Section 21.18 "ST Channel 1 Prescalar Register" on page 210.	2927
	Added information in COUNT bit description in Section 21.20 "ST Current Counter Value 0 Register" on page 212 and Section 21.21 "ST Current Counter Value 1 Register" on page 212.	2929
	Added text in Section 26.11.6 "PHASE_SEG1, PHASE_SEG2" on page 291.	2934
	Added paragraph to Section 26.11.11 "Resynchronization Jump Width" on page 292.	2935
	Updated CHANEN bit description in Section 26.37 "CAN Channel Control Register" on page 313.	2936
	Added Section 26.15 "Using the CAN Module" on page 294.	2928
	Added notes to PHSEG2 and SJW bit descriptions in Section 26.22 "CAN Mode Register" on page 302.	2933
	Updated values in Table 28-4, "ADC Characteristics," on page 381.	2938





6021B-ATARM-02-Mar-07

# **Table of Contents**

	Features	1
1	Description	1
2	Pin Configuration	3
3	Signal Description	5
4	Block Diagram	8
5	Product Overview	9
	5.1 Register Considerations	9
	5.2 Power Consumption	
	5.3 Reset	13
	5.4 Electrical Characteristics	14
6	Clocks	.18
	6.1 Overview	18
	6.2 Crystals	18
	6.3 Phase Locked Loop	18
	6.4 Clock Timings	20
	6.5 Internal Oscillator Characteristics	21
7	Memory Map	.22
	7.1 Reboot Mode	22
	7.2 Remap Mode	23
	7.3 External Memory	23
	7.4 Peripheral Memory	24
8	Power Management Block	.25
9	PIO Controller Block	.27
10	Multiplexed I/O Lines	.28
	10.1 Output Selection	28
	10.2 I/O Levels	28
	10.3 Interrupts	28
	10.4 User Interface	29
	10.5 Multi-driver (Open Drain)	29
	10.6 MPIO Block Diagram	30
11	Advanced Memory Controller (AMC)	.31





	11.1 Overview	31
	11.2 Boot on NCS0	31
	11.3 External Memory Mapping	31
	11.4 External Memory Device Connection	32
	11.5 External Bus Interface Timings	36
	11.6 Advanced Memory Controller (AMC) Memory Map	54
	11.7 AMC Chip Select Register 0	55
	11.8 AMC Chip Select Register	57
	11.9 AMC Remap Control Register	59
	11.10 AMC Memory Control Register	59
12	Clock Manager (CM)	61
	12.1 Overview	61
	12.2 Clock Manager (CM) Memory Map	64
	12.3 CM Clock Enable Register	65
	12.4 CM Clock Disable Register	65
	12.5 CM Clock Status Register	66
	12.6 CM PLL Stabilization Timer Register	68
	12.7 CM PLL Divider Register	69
	12.8 CM Oscillator Stabilization Timer Register	70
	12.9 CM Master Clock Divider Register	71
13	Special Function Mode (SFM)	72
	13.1 Overview	72
	13.2 Chip Identification	72
	13.3 Reset status	72
	13.4 Special Function Mode (SFM) Memory Map	72
	13.5 SFM Chip ID Register	73
	13.6 SFM Reset Status Register	74
14	Watchdog (WD)	75
	14.1 Overview	75
	14.2 Architecture	75
	14.3 Block Diagram	76
	14.4 Example	76
	14.5 Watchdog (WD) Memory Map	78
	14.6 WD Control Register	78
	14.7 WD Mode Register	79

# " AT91SAM7A2

	14.8 WD Overflow Mode Register	80
	14.9 WD Clear Status Register	81
	14.10 WD Status Register	82
	14.11 WD Interrupt Enable Register	83
	14.12 WD Interrupt Disable Register	83
	14.13 WD Interrupt Mask Register	84
	14.14 WD Pending Window Register	85
15	Watch Timer (WT)	86
	15.1 Overview	86
	15.2 Example	86
	15.3 Watch Timer (WT) Memory Map	87
	15.4 WT Control Register	88
	15.5 WT Mode Register	89
	15.6 WT Clear Status Register	90
	15.7 WT Status Register	91
	15.8 WT Interrupt Enable Register	92
	15.9 WT Interrupt Disable Register	92
	15.10 WT Interrupt Mask Register	93
	15.11 WT Seconds Register	94
	15.12 WT Alarm Register	94
16	Peripheral Data Controller (PDC)	95
	16.1 Overview	95
	16.2 Block Diagram	95
	16.3 PDC Transfers	97
	16.4 Memory Pointers	97
	16.5 Transfer Counter	97
	16.6 PDC Configuration	
	16.7 PDC Transfer Example	99
	16.8 Peripheral Data Controller (PDC) Memory Map	
	16.9 PDC CH0CH9 Peripheral Register Address	
	16.10 PDC CH0CH9 Control Register	103
	16.11 PDC CH0CH9 Memory Pointer Register	104
	16.12 PDC CH0CH9 Transfer Register	105
17	Generic Interrupt Controller (GIC)	106
	17.1 Overview	





	17.2 Interrupt Handling	108
	17.3 Standard Interrupt Sequence	
	17.4 Fast Interrupt Sequence	
	17.5 Spurious Interrupt Sequence	
	17.6 Generic Interrupt Controller (GIC) Memory Map	
	17.7 GIC Source Mode Register	
	17.8 GIC Source Vector Register	
	17.9 GIC Interrupt Vector Register	
	17.10 GIC FIQ Vector Register	
	17.11 GIC Interrupt Status Register	
	17.12 GIC Interrupt Pending Register	
	17.13 GIC Interrupt Mask Register	
	17.14 GIC Core Interrupt Status Register	
	17.15 GIC Interrupt Enable Command Register	
	17.16 GIC Interrupt Disable Command Register	
	17.17 GIC Interrupt Clear Command Register	
	17.18 GIC Interrupt Set Command Register	120
	17.19 GIC End of Interrupt Command Register	120
	17.20 GIC Spurious Vector Register	121
18		
18	10-bit Analog to Digital Converter (ADC)	122
18	10-bit Analog to Digital Converter (ADC)	<b>122</b> 122
18	<b>10-bit Analog to Digital Converter (ADC)</b> 18.1 Overview 18.2 Block Diagram	<b>122</b> 122 124
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> </ul>	<b>122</b> 122 124 127
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> </ul>	<b>122</b> 122 124 127 127
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> <li>18.5 10-bit Analog to Digital Converter (ADC) Memory Map</li> </ul>	<b>122</b> 122 124 127 127 129
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> </ul>	<b>122</b> 122 124 127 127 129 130
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> <li>18.5 10-bit Analog to Digital Converter (ADC) Memory Map</li> <li>18.6 ADC Enable Clock Register</li> </ul>	<b>122</b> 122 124 127 127 129 130 130
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> <li>18.5 10-bit Analog to Digital Converter (ADC) Memory Map</li> <li>18.6 ADC Enable Clock Register</li> <li>18.7 ADC Disable Clock Register</li> </ul>	
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> <li>18.5 10-bit Analog to Digital Converter (ADC) Memory Map</li> <li>18.6 ADC Enable Clock Register</li> <li>18.7 ADC Disable Clock Register</li> <li>18.8 ADC Power Management Status Register</li> </ul>	<b>122</b> 122 124 127 127 127 129 130 130 131 132
18	<ul> <li>10-bit Analog to Digital Converter (ADC)</li> <li>18.1 Overview</li> <li>18.2 Block Diagram</li> <li>18.3 Power Management</li> <li>18.4 Example of Use</li> <li>18.5 10-bit Analog to Digital Converter (ADC) Memory Map</li> <li>18.6 ADC Enable Clock Register</li> <li>18.7 ADC Disable Clock Register</li> <li>18.8 ADC Power Management Status Register</li> <li>18.9 ADC Control Register</li> </ul>	<b>122</b> 122 124 127 127 127 129 130 130 131 132 133
18	<b>10-bit Analog to Digital Converter (ADC)</b> 18.1 Overview         18.2 Block Diagram         18.3 Power Management         18.4 Example of Use         18.5 10-bit Analog to Digital Converter (ADC) Memory Map         18.6 ADC Enable Clock Register         18.7 ADC Disable Clock Register         18.8 ADC Power Management Status Register         18.9 ADC Control Register         18.10 ADC Mode Register	<b>122</b> 124 124 127 127 127 127 130 130 131 132 133 135
18	<b>10-bit Analog to Digital Converter (ADC)</b> 18.1 Overview         18.2 Block Diagram         18.3 Power Management         18.4 Example of Use         18.5 10-bit Analog to Digital Converter (ADC) Memory Map         18.6 ADC Enable Clock Register         18.7 ADC Disable Clock Register         18.8 ADC Power Management Status Register         18.9 ADC Control Register         18.10 ADC Mode Register         18.11 DC Conversion Mode Register	<b>122</b> 122 124 127 127 127 129 130 130 131 132 133 135 136
18	<b>10-bit Analog to Digital Converter (ADC)</b> 18.1 Overview         18.2 Block Diagram         18.3 Power Management         18.4 Example of Use         18.5 10-bit Analog to Digital Converter (ADC) Memory Map         18.6 ADC Enable Clock Register         18.7 ADC Disable Clock Register         18.8 ADC Power Management Status Register         18.9 ADC Control Register         18.10 ADC Mode Register         18.11 DC Conversion Mode Register         18.12 ADC Clear Status Register	<b>122</b> 122 124 127 127 127 127 129 130 130 131 132 135 136 137
18	<b>10-bit Analog to Digital Converter (ADC)</b> 18.1 Overview         18.2 Block Diagram         18.3 Power Management         18.4 Example of Use         18.5 10-bit Analog to Digital Converter (ADC) Memory Map         18.6 ADC Enable Clock Register         18.7 ADC Disable Clock Register         18.8 ADC Power Management Status Register         18.10 ADC Control Register         18.11 DC Conversion Mode Register         18.12 ADC Clear Status Register         18.13 ADC Status Register	<b>122</b> 124 124 127 127 127 127 130 130 131 132 133 135 136 137 139

# iv AT91SAM7A2

18.17 ADC Convert Data Register	141
Universal Synchronous/ Asynchronous Receiver/Transn (USART) 142	nitter
19.1 Overview	142
19.2 Block Diagram	142
19.3 Baud Rate Generator	143
19.4 Receivers	143
19.5 Transmitter	146
19.6 Break Condition	147
19.7 LIN Protocol	148
19.8 Line Configuration in LIN Mode	148
19.9 Message Characteristics	149
19.10 Smart Card Protocol (ISO7816-3)	149
19.11 Line Configuration in Smart Card Mode	151
19.12 PIO Controller	151
19.13 Power Management	151
19.14 Example	151
19.15 USART Memory Map	153
19.16 USART PIO Enable Register	155
19.17 USART PIO Disable Register	155
19.18 USART PIO Status Register	156
19.19 USART PIO Output Enable Register	157
19.20 USART PIO Output Disable Register	157
19.21 USART PIO Output Status Register	158
19.22 USART PIO Set Output Data Register	159
19.23 USART PIO Clear Output Data Register	159
19.24 USART PIO Output Data Status Register	160
19.25 USART PIO Pin Data Status Register	161
19.26 USART PIO Multi Drive Enable Register	162
19.27 USART PIO Multi Drive Disable Register	162
19.28 USART PIO Multi Drive Status Register	163
19.29 USART Enable Clock Register	164
19.30 USART Disable Clock Register	164
19.31 USART Power Management Status Register	165
19.32 USART Control Register	166
19.33 USART Mode Register	168





	19.34 USART Clear Status Register	171
	19.35 USART Status Register	172
	19.36 USART Interrupt Enable Register	175
	19.37 USART Interrupt Disable Register	175
	19.38 USART Interrupt Mask Register	176
	19.39 USART Receiver Holding Register	178
	19.40 USART Transmit Holding Register	178
	19.41 USART Baud Rate Generator Register	179
	19.42 USART Receiver Time Out Register	180
	19.43 USART Transmit Time Guard Register	181
	19.44 USART LIN Identifier Register	181
	19.45 USART Data Field Write 0 Register	182
	19.46 USART Data Field Write 1 Register	182
	19.47 USART Data Field Read 0 Register	183
	19.48 USART Data Field Read 1 Register	183
	19.49 USART Sync Break Length Register	184
20	Capture (CAPT)	185
	20.1 Overview	105
		185
	20.1 Overview	
		185
	20.2 Example of Use	185 186
	20.2 Example of Use	185 186 186
	20.2 Example of Use 20.3 Capture Limits 20.4 Power Management	185 186 186 187
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> </ul>	185 186 186 187 188
	<ul> <li>20.2 Example of Use</li></ul>	185 186 186 187 188 188
	<ul> <li>20.2 Example of Use</li></ul>	185 186 186 187 188 188 189
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> </ul>	185 186 186 187 188 188 189 190
	<ul> <li>20.2 Example of Use</li></ul>	185 186 186 187 188 188 189 190 191
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> </ul>	185 186 186 187 188 188 189 190 191 192
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> <li>20.11 CAPTURE Clear Status Register</li> </ul>	185 186 186 187 188 188 189 190 191 192 193
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> <li>20.11 CAPTURE Clear Status Register</li> <li>20.12 CAPTURE Status Register</li> </ul>	185 186 186 187 188 188 189 190 191 191 192 193 194
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> <li>20.11 CAPTURE Clear Status Register</li> <li>20.12 CAPTURE Status Register</li> <li>20.13 CAPTURE Interrupt Enable Register</li> </ul>	185 186 186 187 188 188 189 190 191 191 192 193 194 194
	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Power Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> <li>20.11 CAPTURE Clear Status Register</li> <li>20.12 CAPTURE Status Register</li> <li>20.13 CAPTURE Interrupt Enable Register</li> <li>20.14 CAPTURE Interrupt Disable Register</li> </ul>	185 186 186 187 188 188 189 190 191 191 193 194 194 195
21	<ul> <li>20.2 Example of Use</li> <li>20.3 Capture Limits</li> <li>20.4 Power Management</li> <li>20.5 Capture (CAPT) Memory Map</li> <li>20.6 CAPTURE Enable Clock Register</li> <li>20.7 CAPTURE Disable Clock Register</li> <li>20.8 CAPTURE Dower Management Status Register</li> <li>20.9 CAPTURE Control Register</li> <li>20.10 CAPTURE Mode Register</li> <li>20.11 CAPTURE Clear Status Register</li> <li>20.12 CAPTURE Status Register</li> <li>20.13 CAPTURE Interrupt Enable Register</li> <li>20.14 CAPTURE Interrupt Mask Register</li> </ul>	185 186 186 187 188 188 189 190 191 191 192 194 195 196

vi AT91SAM7A2

	21.2 Block Diagram	.198
	21.3 Peripheral Structure	. 198
	21.4 Power Management	. 198
	21.5 Example of Use	. 198
	21.6 Simple Timer (ST) Memory Map	.200
	21.7 ST Enable Clock Register	.201
	21.8 ST Disable Clock Register	.201
	21.9 ST Power Management Status Register	.202
	21.10 ST Control Register	.203
	21.11 ST Clear Status Register	.204
	21.12 ST Status Register	.205
	21.13 ST Interrupt Enable Register	.206
	21.14 ST Interrupt Disable Register	.206
	21.15 ST Interrupt Mask Register	.207
	21.16 ST Channel 0 Prescalar Register	.208
	21.17 ST Channel 0 Counter Register	.209
	21.18 ST Channel 1 Prescalar Register	.210
	21.19 ST Channel 1 Counter Register	.211
	21.20 ST Current Counter Value 0 Register	.212
	<ul><li>21.20 ST Current Counter Value 0 Register</li><li>21.21 ST Current Counter Value 1 Register</li></ul>	
F	-	.212
F	21.21 ST Current Counter Value 1 Register	.212 <b>213</b>
F	21.21 ST Current Counter Value 1 Register Pulse Width Modulator (PWM)	.212 <b>213</b> .213
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li> <li>Pulse Width Modulator (PWM)</li> <li>22.1 Overview</li> </ul>	.212 <b>213</b> .213 .213
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li> <li>Pulse Width Modulator (PWM)</li> <li>22.1 Overview</li> <li>22.2 Block Diagram</li> </ul>	.212 <b>213</b> .213 .213 .213
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li> <li>Pulse Width Modulator (PWM)</li> <li>22.1 Overview</li></ul>	.212 <b>213</b> .213 .213 .213 .213
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li></ul>	.212 <b>213</b> .213 .213 .213 .213 .213 .214
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li></ul>	.212 <b>213</b> .213 .213 .213 .213 .213 .214 .214
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li></ul>	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li></ul>	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215 .216
F	21.21 ST Current Counter Value 1 Register         Pulse Width Modulator (PWM)         22.1 Overview         22.2 Block Diagram         22.3 Pin Description         22.4 PWM Parameters         22.5 Power Management         22.6 Example of Use         22.7 Pulse Width Modulator (PWM) Memory Map         22.8 PWM Enable Clock Register	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215 .216 .216
F	21.21 ST Current Counter Value 1 Register         Pulse Width Modulator (PWM)         22.1 Overview         22.2 Block Diagram         22.3 Pin Description         22.4 PWM Parameters         22.5 Power Management         22.6 Example of Use         22.7 Pulse Width Modulator (PWM) Memory Map         22.8 PWM Enable Clock Register         22.9 PWM Disable Clock Register	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215 .216 .216 .217
F	<ul> <li>21.21 ST Current Counter Value 1 Register</li> <li>Pulse Width Modulator (PWM)</li> <li>22.1 Overview</li> <li>22.2 Block Diagram</li> <li>22.3 Pin Description</li> <li>22.4 PWM Parameters</li> <li>22.5 Power Management</li> <li>22.6 Example of Use</li> <li>22.7 Pulse Width Modulator (PWM) Memory Map</li> <li>22.8 PWM Enable Clock Register</li> <li>22.9 PWM Disable Clock Register</li> <li>22.10 PWM Power Management Status Register</li> </ul>	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215 .216 .216 .217 .218
F	21.21 ST Current Counter Value 1 Register         Pulse Width Modulator (PWM)         22.1 Overview         22.2 Block Diagram         22.3 Pin Description         22.4 PWM Parameters         22.5 Power Management         22.6 Example of Use         22.7 Pulse Width Modulator (PWM) Memory Map         22.8 PWM Enable Clock Register         22.9 PWM Disable Clock Register         22.10 PWM Power Management Status Register         22.11 PWM Control Register	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .215 .216 .216 .217 .218 .219
F	21.21 ST Current Counter Value 1 Register         Pulse Width Modulator (PWM)         22.1 Overview         22.2 Block Diagram         22.3 Pin Description         22.4 PWM Parameters         22.5 Power Management         22.6 Example of Use         22.7 Pulse Width Modulator (PWM) Memory Map         22.8 PWM Enable Clock Register         22.9 PWM Disable Clock Register         22.10 PWM Power Management Status Register         22.11 PWM Control Register         22.12 PWM Mode Register	.212 <b>213</b> .213 .213 .213 .213 .213 .214 .214 .215 .216 .216 .217 .218 .219 .220
F	21.21 ST Current Counter Value 1 Register         Pulse Width Modulator (PWM)         22.1 Overview         22.2 Block Diagram         22.3 Pin Description         22.4 PWM Parameters         22.5 Power Management         22.6 Example of Use         22.7 Pulse Width Modulator (PWM) Memory Map         22.8 PWM Enable Clock Register         22.9 PWM Disable Clock Register         22.10 PWM Power Management Status Register         22.11 PWM Control Register         22.12 PWM Mode Register         22.13 PWM Clear Status Register	.212 <b>213</b> .213 .213 .213 .213 .214 .214 .214 .215 .216 .216 .217 .218 .219 .220 .221





	22.16 PWM Interrupt Disable Register	222
	22.17 PWM Interrupt Mask Register	223
	22.18 PWM Delay Register x [x = 03]	224
	22.19 PWM Pulse Register x [x = 03]	225
23	Serial Peripheral Interface (SPI)	226
	23.1 Overview	226
	23.2 Master Mode	226
	23.3 Slave Mode	230
	23.4 PIO Controller	230
	23.5 Power Management	230
	23.6 Serial Peripheral Interface (SPI) Memory Map	231
	23.7 SPI PIO Enable Register	232
	23.8 SPI PIO Disable Register	232
	23.9 SPI PIO Status Register	233
	23.10 SPI PIO Output Enable Register	234
	23.11 SPI PIO Output Disable Register	234
	23.12 SPI PIO Output Status Register	235
	23.13 SPI PIO Set Output Data Register	236
	23.14 SPI PIO Clear Output Data Register	236
	23.15 SPI PIO Output Data Status Register	237
	23.16 SPI PIO Pin Data Status Register	238
	23.17 SPI PIO Multi-driver Enable Register	239
	23.18 SPI PIO Multi-driver Disable Register	239
	23.19 SPI PIO Multi-driver Status Register	240
	23.20 SPI Enable Clock Register	241
	23.21 SPI Disable Clock Register	241
	23.22 SPI Power Management Status Register	242
	23.23 SPI Control Register	243
	23.24 SPI Mode Register	244
	23.25 SPI Status Register	246
	23.26 SPI Interrupt Enable Register	248
	23.27 SPI Interrupt Disable Register	248
	23.28 SPI Interrupt Mask Register	250
	23.29 SPI Receive Data Register	251
	23.30 SPI Transmit Data Register	252
	23.31 SPI Chip Select Register 03	253

# viii AT91SAM7A2

	23.32 Timing Diagrams	255
24	Unified Parallel I/O Controller (UPIO)	257
	24.1 Overview	257
	24.2 Output Selection	257
	24.3 I/O Levels	257
	24.4 Interrupts	257
	24.5 User Interface	257
	24.6 Multi-Driver (Open Drain)	257
	24.7 Block Diagram	258
	24.8 Special Multiplexed PIO	258
	24.9 Power Management	259
	24.10 Unified Parallel I/O Controller (UPIO) Memory Map	
	24.11 UPIO Output Enable Register	261
	24.12 UPIO Output Disable Register	261
	24.13 UPIO Output Status Register	261
	24.14 UPIO Set Output Data Register	
	24.15 UPIO Clear Output Data Register	
	24.16 UPIO Output Data Status Register	
	24.17 UPIO Pin Data Status Register	
	24.18 UPIO Multi-driver Enable Register	
	24.19 UPIO Multi-driver Disable Register	264
	24.20 UPIO Multi-driver Status Register	264
	24.21 UPIO Enable Clock Register	
	24.22 UPIO Disable Clock Register	265
	24.23 UPIO Power Management Status Register	265
	24.24 UPIO Control Register	
	24.25 UPIO Mode Register	
	24.26 UPIO Status Register	
	24.27 UPIO Interrupt Enable Register	
	24.28 UPIO Interrupt Disable Register	
	24.29 UPIO Interrupt Mask Register	
25	Power Management Controller (PMC)	269
	25.1 Overview	269
	25.2 Power Management Controller (PMC) Memory Map	270
	25.3 PMC Enable Clock Register	270





26

25.4 PMC Disable Clock	Register	271
25.5 PMC Power Manag	ement Status Register	272
Controller Area Netwo	ork (CAN)	273
26.1 Overview		273
26.2 Basic Concepts		
26.3 Channel Overview		
26.4 Message Transfer .		
26.5 Message Filtering .		
26.6 Message Validation	۱	
26.7 Coding		
26.8 Error Handling		
26.9 Fault Confinement		
26.10 Oscillator Tolerand	е	
26.11 Bit Timing Require	ments	
26.12 Reception Mode .		
26.13 Time Stamp		
26.14 Power Manageme	nt	
26.15 Using the CAN Mo	dule	
26.16 Example of Use		
26.17 Controller Area Ne	etwork (CAN) Memory Map	296
26.18 CAN Enable Clock	Register	
26.19 CAN Disable Clock	k Register	
26.20 CAN Power Manag	gement Status Register	
26.21 CAN Control Regis	ster	
26.22 CAN Mode Registe	er	
26.23 CAN Clear Status	Register	
26.24 CAN Status Regis	ter	
26.25 CAN Interrupt Ena	ble Register	
26.26 CAN Interrupt Disa	able Register	
26.27 CAN Interrupt Mas	sk Register	
26.28 CAN Clear Interru	pt Source Status Register	
26.29 CAN Interrupt Sou	rce Status Register	
26.30 CAN Source Interr	rupt Enable Register	
26.31 CAN Source Interr	rupt Disable Register	
26.32 CAN Source Interr	rupt Mask Register	
26.33 CAN Channel Data	a Register A	

# × AT91SAM7A2

26.34 CAN Channel Data Register B	311
26.35 CAN Channel Mask Register	311
26.36 CAN Channel Identifier Register	312
26.37 CAN Channel Control Register	313
26.38 CAN Channel Stamp Register	314
26.39 CAN Channel Clear Status Register	315
26.40 CAN Channel Status Register	317
26.41 CAN Channel Interrupt Enable Register	319
26.42 CAN Channel Interrupt Disable Register	319
General-purpose Timer (GPT)	
27.1 Overview	320
27.2 Block Diagram	320
27.3 Pin Description	321
27.4 Clock Sources	
27.5 16-bit Counter	
27.6 16-bit Registers	324
27.7 External Edge Detection	
27.8 Interrupts	325
27.9 PIO Controller	325
27.10 Power Management	326
27.11 Example Of Use	326
27.12 General-purpose Timer (GPT) Memory Map	327
27.13 GPT PIO Enable Register	329
27.14 GPT PIO Disable Register	329
27.15 GPT PIO Status Register	
27.16 GPT PIO Output Enable Register	331
27.17 GPT PIO Output Disable Register	331
27.18 GPT PIO Output Status Register	
27.19 GPT PIO Set Output Data Register	
27.20 GPT PIO Clear Output Data Register	
27.21 GPT PIO Output Data Status Register	334
27.22 GPT PIO Pin Data Status Register	335
27.23 GPT PIO Multi-driver Enable Register	336
27.24 GPT PIO Multi-driver Disable Register	336
27.25 GPT PIO Multi-driver Status Register	337
27.26 GPT Enable Clock Register	





	27.27	GPT Disable Clock Register	
	27.28	GPT Power Management Status Register	
	27.29	General-purpose Timer in Capture Mode	
	27.30	GPT Control Register in Capture Mode	
	27.31	GPT Mode Register in Capture Mode	
	27.32	GPT Status Register in Capture Mode	
	27.33	GPT Interrupt Enable Register in Capture Mode	
	27.34	GPT Interrupt Disable Register in Capture Mode	
	27.35	GPT Interrupt Mask Register in Capture Mode	
	27.36	GPT Counter Value in Capture Mode	
	27.37	GPT Register A in Capture Mode	
	27.38	GPT Register B in Capture Mode	
	27.39	GPT Register C in Capture Mode	
	27.40	General-purpose Timer in Waveform Mode	
	27.41	GPT Control Register in Waveform Mode	
	27.42	GPT Mode Register in Waveform Mode	
	27.43	GPT Status Register in Waveform Mode	
	27.44	GPT Interrupt Enable Register in Waveform Mode	
	27.45	GPT Interrupt Disable Register in Waveform Mode	
	27.46	GPT Interrupt Mask Register in Waveform Mode	
	27.47	GPT Counter Value in Waveform Mode	
	27.48	GPT Register A in Waveform Mode	
	27.49	GPT Register B in Waveform Mode	
	27.50	GPT Register C in Waveform Mode	
	27.51	Timer Controller Block Programming	
	27.52	GPT Block Control Register	
	27.53	GPT Block Mode Register	
28	Electri	cal Characteristics	
	28.1 3	3V Core and I/O Characteristics	
	28.2 5	5V I/O Characteristics	
	28.3	Analog Characteristics	
	28.4	ADC Characteristics	
29	Packad	ging Information	383
-	-	Thermal Data	
		Package Drawing	
		<b>U U</b>	

# xii AT91SAM7A2

30	Soldering Profile	
31	Environmental Specifications	
	31.1 Operational Temperature	
	31.2 Storage Temperature	
32	Revision History	
	Table of Contents	i





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